## LA1787M

## Monolithic Linear IC

## Single-Chip Tuner IC for Car Radios

## Overview

The LA1787M integrates all six blocks required in a car radio tuner on a single chip.

## Functions

- FM front end • FM IF • Noise canceller • Multiple
- AM up-conversion
- FM/AM switch
- MRC


## Features

- Improved noise reduction methods
— The FM front end provides excellent 3-signal characteristics equivalent to those of the LA1193M.
- Superlative listenability due to improved medium and weak field noise canceller characteristics.
- Improved separation characteristics
- Anti-birdie filter
- Improved AM and FM thermal characteristics
— Excellent FM signal meter linearity
- Modified N.C. circuit for improved noise rejection
- Improved AM adjacent channel interference characteristics ( $\Delta 40 \mathrm{kHz}$ )
- Double conversion AM tuner (up conversion) Reduces the number of external components required as compared to earlier double conversion tuners, in particular, no crystal is required (when used in conjunction with the LC72144).
- Sample-to-sample variation reduction circuit built into the FM IF circuit.
(Fixed resistors are used for the SD, keyed AGC, mute on adjustment, ATT, SNC, and HCC functions.)
- Improved FM separation temperature characteristics
- The LA1787 inherits the block arrangement of the LA1780M and supports pin-compatible designs.


## Package Dimensions

unit : mm (typ)
QIP64E(14X14)



## Specifications

Maximum Ratings at $\mathbf{T a}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :--- | :---: | :---: |
| Maximum supply voltage | $\mathrm{V}_{\mathrm{CC}} 1 \mathrm{max}$ | Pins 6, 40, and 61 | 9 | V |
|  | $\mathrm{~V}_{\mathrm{CC}} 2 \mathrm{max}$ | Pins $7,45,54,59$, and 60 | V |  |
| Allowable power dissipation | $\mathrm{Pd} \max$ | $\mathrm{Ta} \leq 55^{\circ} \mathrm{C}$ | mW |  |
| Operating temperature | Topr |  | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |

Operating Conditions at $\mathbf{T a}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :--- | :---: | :---: |
| Recommended supply voltage | $\mathrm{V}_{\mathrm{CC}}$ | Pins $6,7,40,45,54,59,60$, and 61 | 8 | V |
|  | $\mathrm{~V}_{\mathrm{CC}} \mathrm{ST}$ IND | Pin 26 | V | V |
| Operating supply voltage range | $\mathrm{V}_{\mathrm{CC}}$ op |  | 7.5 to 9.0 | V |

Operating Characteristics at $\mathrm{Ta}=\mathbf{2 5}^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=\mathbf{8 . 0} \mathrm{V}$, in the specified test cricuit for the FM IF input

| Parameter | Symbol | Conditions | Ratings |  |  | unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | min | typ | max |  |
| [FM Characteristics] At the FM IF input |  |  |  |  |  |  |
| Current drain | ICco-FM | No input, I 40 + I 45 + $\mathrm{I} 54+\mathrm{I} 59+\mathrm{I} 60$ + I 61 | 60 | 94 | 110 | mA |
| Demodulation output | $\mathrm{V}_{\mathrm{O}}$-FM | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}, 100 \% \mathrm{mod}$, The pin 15 output | 205 | 310 | 415 | mVrms |
| Pin 31 demodulation output | $\mathrm{V}_{\mathrm{O}}$-FM31 | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}, 100 \% \mathrm{mod}$, The pin 31 output | 190 | 295 | 380 | mVrms |
| Channel balance | CB | The ratio between pins 15 and 16 at $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}$ | -1 | 0 | +1 | dB |
| Total harmonic distortion | THD-FM mono | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}, 100 \%$ mod, pin 15 |  | 0.3 | 1 | \% |
| Signal-to-noise ratio: IF | S/N-FM IF | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}, 100 \%$ mod, pin 15 | 75 | 82 |  | dB |
| AM suppression ratio: IF | AMR IF | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 30 \% \mathrm{AM}$, pin 15 | 55 | 68 |  | dB |
| Muting attenuation | Att-1 | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}$. The pin 15 attenuation when V 33 goes from 0 to 2 V | 5 | 10 | 15 | dB |
|  | Att-2 | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}$. The pin 15 attenuation when V 33 goes from 0 to $2 \mathrm{~V}^{* 1}$ | 15 | 20 | 25 | dB |
|  | Att-3 | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 1 \mathrm{kHz}$. The pin 15 attenuation when V 33 goes from 0 to 2 V *2 | 28 | 33 | 38 | dB |
| Separation | Separation | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, \mathrm{~L}+\mathrm{R}=90 \%$, pilot $=10 \%$. The pin 15 output ratio | 30 | 40 |  | dB |
| Stereo on level | ST-ON | The pilot modulation such that $\mathrm{V} 26<0.5 \mathrm{~V}$ | 1.2 | 2.4 | 4.4 | \% |
| Stereo off level | ST-OFF | The pilot modulation such that $\mathrm{V} 26>3.5 \mathrm{~V}$ | 0.6 | 1.6 |  | \% |
| Main total harmonic distortion | THD-Main L | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, \mathrm{~L}+\mathrm{R}=90 \%$, pilot $=10 \%$. The pin 15 signal |  | 0.3 | 1.2 | \% |
| Pilot cancellation | PCAN | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu \text {, pilot }=10 \% \text {. }$ <br> The pin 15 signal/the pilot level leakage. DIN audio | 20 | 30 |  | dB |
| SNC output attenuation | AttSNC | $\begin{aligned} & 10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, \mathrm{~L}-\mathrm{R}=90 \% \text {, pilot }=10 \% . \\ & \mathrm{V} 28=3 \mathrm{~V} \rightarrow 0.6 \mathrm{~V} \text {, pin } 15 \end{aligned}$ | 1 | 5 | 9 | dB |
| HCC output attenuation | AttHCC-1 | $\begin{aligned} & 10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 10 \mathrm{kHz}, \mathrm{~L}+\mathrm{R}=90 \% \text {, pilot }=10 \% . \\ & \mathrm{V} 29=3 \mathrm{~V} \rightarrow 0.6 \mathrm{~V} \text {, pin } 15 \end{aligned}$ | 0.5 | 4.5 | 8.5 | dB |
|  | AttHCC-2 | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu, 10 \mathrm{kHz}, \mathrm{L}+\mathrm{R}=90 \%$, pilot $=10 \%$. V29 $=3 \mathrm{~V} \rightarrow 0.1 \mathrm{~V}$, pin 15 | 6 | 10 | 14 | dB |
| Input limiting voltage | Vi-lim | $100 \mathrm{~dB} \mu, 10.7 \mathrm{MHz}, 30 \%$ modulation. The IF input such that the input reference output goes down by 3 dB | 33 | 40 | 47 | dB $\mu$ |
| Muting sensitivity | Vi-mute | The IF input level (unmodulated) when $\mathrm{V} 33=2 \mathrm{~V}$ | 27 | 35 | 43 | dB $\mu$ |
| SD sensitivity | SD-sen1 FM | The IF input level (unmodulated) (over 100 mV rms) such that the IF counter buffer output goes on | 54 | 62 | 70 | dB $\mu$ |
|  | SD-sen2 FM |  | 54 | 62 | 70 | dB $\mu$ |
| IF counter buffer output | $\mathrm{V}_{\text {IFBUFF-FM }}$ | $10.7 \mathrm{MHz}, 100 \mathrm{~dB} \mu$, unmodulated. The pin 23 output | 130 | 200 | 270 | mVrms |
| Signal meter output | $V_{S M} \mathrm{FM}-1$ | No input. The pin 24 DC output, unmodulated | 0.0 | 0.1 | 0.3 | V |
|  | $\mathrm{V}_{\text {SM }}$ FM-2 | $50 \mathrm{~dB} \mu$. The pin 24 DC output, unmodulated | 0.4 | 1.0 | 1.5 | V |
|  | $\mathrm{V}_{\text {SM }}$ FM-3 | $70 \mathrm{~dB} \mu$. The pin 24 DC output, unmodulated | 2.0 | 2.7 | 3.5 | V |
|  | $\mathrm{V}_{\text {SM }}$ FM-4 | $100 \mathrm{~dB} \mu$. The pin 24 DC output, unmodulated | 4.7 | 5.5 | 6.2 | V |
| Muting bandwidth | BW-mute | $100 \mathrm{~dB} \mu$. The bandwidth when V33 $=2 \mathrm{~V}$, unmodulated | 150 | 220 | 290 | kHz |
| Mute drive output | $\mathrm{V}_{\text {MUTE-100 }}$ | $100 \mathrm{~dB} \mu, 0 \mathrm{~dB} \mu$. The pin 33 DC output, unmodulated | 0.00 | 0.03 | 0.20 | V |

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| Parameter | Symbol | Conditions | Ratings |  |  | unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | min | typ | max |  |
| [FM FE Mixer Input |  |  |  |  |  |  |
| N-AGC on input | $\mathrm{V}_{\mathrm{N}}$-AGC | 83 MHz , unmodulated. <br> The input such that the pin 2 voltage is 2.0 V or below | 81 | 88 | 95 | dB $\mu$ |
| W-AGC on input | $\mathrm{V}_{\mathrm{w}} \mathrm{AGC}$ | 83 MHz , unmodulated. The input such that the pin 2 voltage is 2.0 V or below. (When the keyed AGC is set to 4.0 V .) | 104 | 110 | 116 | dB $\mu$ |
| Conversion gain | A.V | $83 \mathrm{MHz}, 80 \mathrm{~dB} \mu$, unmodulated. The FE CF output | 19 | 30 | 48 | mVrms |
| Oscillator buffer output | V ${ }_{\text {OScbuFFFm }}$ | No input | 85 | 110 | 165 | mVrms |
| [NC Block] NC input (pin 30) |  |  |  |  |  |  |
| Gate time | $\tau$ GATE1 | $\mathrm{f}=1 \mathrm{kHz}$, for a $1-\mu \mathrm{s}, 100-\mathrm{mV}$ p-o pulse |  | 55 |  | $\mu \mathrm{s}$ |
| Noise sensitivity | SN | The level of a $1=k H z, 1-\mu \mathrm{s}$ pulse input that starts noise canceller operation. Measured at pin 30. |  | 40 |  | mVp-o |
| $N C$ effect | SN-NC | The pulse rejection effect provided by the noise canceller. For a repeated $1-\mu \mathrm{s}$ wide pulse, frequency $=10 \mathrm{kHz}$, 150 mV p-o. The ratio of the FM mode pin 15 output referenced to the AM mode pin 15 output (effective value) | 5 |  |  |  |
| [Multipath Rejection Circuit] MRC input (pin 27) |  |  |  |  |  |  |
| MRC output | VMRC | $\mathrm{V} 24=5 \mathrm{~V}$ | 2.2 | 2.3 | 2.4 | V |
| MRC operating level | MRC-ON | The pin 32 input level at $f=70 \mathrm{kHz}$ such that pin 24 goes to 5 V and pin 27 goes to 2 V | 10 | 15 | 20 | mVrms |
| [AM Characteristics] AM ANT input |  |  |  |  |  |  |
| Practical sensitivity | S/N-30 | $1 \mathrm{MHz}, 30 \mathrm{~dB} \mu, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 30 \%$ modulation, pin 15 | 20 |  |  | dB |
| Detector output | $\mathrm{V}_{\mathrm{O}}$-AM | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 30 \%$ modulation, pin 15 | 130 | 195 | 270 | mVrms |
| Pin 31 detector output | $\mathrm{V}_{\mathrm{O}}$-AM31 | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 30 \%$ modulation, pin 31 | 110 | 175 | 230 | mVms |
| AGC F.O.M. | $\mathrm{V}_{\text {AGC-FOM }}$ | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu$, referenced to the output, the input amplitude such that the output falls by 10 dB . Pin 15 | 51 | 56 | 61 | dB |
| Signal-to-noise ratio | S/N-AM | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 30 \%$ modulation | 47 | 52 |  | dB |
| Total harmonic distortion | THD-AM | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu, \mathrm{f}_{\mathrm{m}}=1 \mathrm{kHz}, 80 \%$ modulation |  | 0.3 | 1 | \% |
| Signal meter output | $\mathrm{V}_{\text {SM }} \mathrm{AM}-1$ | No input | 0.0 | 0.2 | 0.5 | V |
|  | $\mathrm{V}_{\text {SM }} \mathrm{AM}-2$ | $1 \mathrm{MHz}, 130 \mathrm{~dB} \mu$, unmodulated | 4.8 | 6 | 7.3 | V |
| Oscillator buffer output | Voscbuff am1 | No input, the pin 15 output | 185 | 230 |  | mVrms |
| Wide band AGC sensitivity | W-AGCsen1 | 1.4 MHz , the input when $\mathrm{V} 46=0.7 \mathrm{~V}$ | 92 | 98 | 104 | dB $\mu$ |
|  | W-AGCsen2 | 1.4 MHz , the input when $\mathrm{V} 46=0.7 \mathrm{~V}$ (seek mode) | 83 | 89 | 95 | dB $\mu$ |
| SD sensitivity | SD-sen1 AM | 1 MHz , the ANT input level such that the IF counter output turns on. | 24 | 30 | 36 | dB $\mu$ |
|  | SD-sen2 AM | 1 MHz , the ANT input level such that the SD pin goes to the on state. | 24 | 30 | 36 | dB $\mu$ |
| IF buffer output | VIFBUFF-AM | $1 \mathrm{MHz}, 74 \mathrm{~dB} \mu$, unmodulated. The pin 23 output | 200 | 290 |  | mVrms |

Note: These measurements must be made using the either the IC-51-0644-824 or KS8277 IC socket (manufactured by Yamaichi Electronics).

* 1. When the resistor between pin 58 and ground is $200 \mathrm{k} \Omega$.
* 2. When the resistor between pin 58 and ground is $30 \mathrm{k} \Omega$.


## Function List

FM Front End (Equivalent to the Sanyo LA1193)

- Double input type double balanced mixer
- Pin diode drive AGC output
- MOSFET second gate drive AGC output
- Keyed AGC adjustment pin
- Differential IF amplifier
- Wide band AGC sensitivity setting pin, and narrow band AGC sensitivity setting pin
- Local oscillator

FM IF

- IF limiter amplifier
- S-meter output (also used for AM) 6-stage pickup
- Multipath detection pin (shared FM signal meter)
- Quadrature detection
- AF preamplifier
- AGC output
- Band muting
- Weak input muting
- Soft muting adjustment pin
- Muting attenuation adjustment pin
- IF counter buffer output (also used for AM)
- SD (IF counter buffer on level) adjustment pin
- SD output (active high) (also used for AM)

Noise Canceller

- High-pass filter (first order)
- Delay circuit based low-pass filter (fourth order)
- Noise AGC
- Pilot signal compensation circuit
- Noise sensitivity setting pin
- Function for disabling the noise canceller in AM mode

Multiplex Functions

- Adjustment-free VCO circuit
- Level follower type pilot canceller circuit
- HCC (high cut control)
- Automatic stereo/mono switching
- VCO oscillation stop function (AM mode)
- Forced monaural
- SNC (stereo noise controller)
- Stereo display pin
- Anti-birdie filter

AM

- Double balanced mixer (1st, 2nd)
- IF amplifier
- Detection
- RF AGC (narrow/wide)
- Pin diode drive pin
- IF AGC
- Signal meter output (also used for FM)
- Local oscillator circuits (first and second)
- Local oscillator buffer output
- IF counter buffer output (also used by the FM IF)
- SD (IF counter buffer on level) adjustment pin
- SD output (active high) (also used for AM)
- Wide AGC
- Detection output frequency characteristics adjustment pin (low cut, high deemphasis)
- AM stereo buffer

MRC (multipath noise rejection circuit)

AM/FM switching output (linked to the FM $\mathrm{V}_{\mathrm{CC}}$ )

## LA1787M

## Operating Characteristics and Symbols Used in the Test Circuit Diagrams

Switches (SW)
Switch on $=1, S W$ off $=0$
There are two switches that use signal transfer.

- SW2: switches between the mixer input and the IF input.
- SW4: switches between noise canceler input and IF output + noise canceler input.

Types of SG used

| PG1 (AC1) | Used for noise canceler testing. A pulse generator and an AF oscillator are required. |
| :--- | :--- |
| AC2 | Used for FM front end testing. Outputs an 83 MHz signal. |
| AC3 | Used for FM IF, noise canceler, and MPX testing. Outputs a 10.7 MHz signal. Stereo modulation must be possible. |
| AC4 | Used for AM testing. Outputs 1 MHz and 1.4 MHz signals. |
| AC5 | Used with the MRC. Can also be used for AF and OSC. |

Power supply

| $\mathrm{V}_{\mathrm{CC}}$ | 8 V |  |  |
| :--- | :--- | :--- | :--- |
| $\mathrm{~V}_{\mathrm{CC}} 1$ | 5 V |  | SD, stereo, seek/stop |
| $\mathrm{V}_{\mathrm{CC}} 2$ | $0.1 \mathrm{~V} / 0.7 \mathrm{~V} / 2 \mathrm{~V} / 4 \mathrm{~V}$ | These levels <br> must be variable. | Keyed AGC, Mute ATT |
| $\mathrm{V}_{\mathrm{CC}} 3$ | $0.1 \mathrm{~V} / 0.6 \mathrm{~V} / 2 \mathrm{~V}$ |  |  |

- Switches

|  | Parameter | ON | OFF |
| :---: | :--- | :---: | :---: |
| SW1 | AM/FM switching. The FE VCc is supplied to pin 62. | FM | AM |
| SW2 | FM IF switching. Pin 51/FE output | FE IF OUT (A) | AC3 (B) |
| SW3 | For conversion gain testing | Conversion gain measurement (A) | Other/purposes |
| SW4 | For switching between noise canceler input and IF output + noise canceler. | AC1 (A) | Other/purposes |
| SW5 | High-speed SD | High-speed SD | Other/purposes |
| SW6 | SEEK/STOP (IF BUFF ON/OFF) | STOP | Seek (IF buffer output) |
| SW7 | MUTE ATT 200 k $\Omega$ | MUTE $200 \mathrm{k} \Omega$ | OFF |
| SW8 | MUTE ATT 30 k $\Omega$ | MUTE $30 \mathrm{k} \Omega$ | OFF |
| SW9 | For pilot cancellation testing | When pilot cancellation is used | When pilot cancellation is not used |
| SW10 | Mute off (pin 33) | MUTE OFF |  |

- Trimmers (variable resistors)

| VR1 | Separation adjustment |
| :--- | :--- |
| VR2 | Pilot cancellation adjustment |

Test Points

- DC voltages

| VD1 | FM RF AGC voltage | Pin 2 |
| :--- | :--- | :--- |
| VD2 | AM/FM SD, AM Tweet, FM stereo indicator | Pin 26 |
| VD3 | AM/FM S-meter | Pin 24 |
| VD4 | MRC output | Pin 27 |
| VD5 | Mute drive output | Pin 33 |
| VD6 | AM antenna damping voltage | Pin 46 |
| VD7 | N.C. Gate time | Pin 8 |

- AC voltages

| VA1 | AM/FM OSC Buff | Pin 4 |
| :--- | :--- | :--- |
| VA2 | First IF output | Pin $53 \rightarrow$ CF $\rightarrow$ pin 51 load level $(10.7 \mathrm{MHz})$ |
| VA3 | IF counter buffer | Pin $23(10.7 \mathrm{MHz} / 450 \mathrm{kHz})$ |
| VA4 | MPX OUT Left ch | Pin $15(\mathrm{AF})$ |
| VA5 | MPX OUT Right ch | Pin $16(\mathrm{AF})$ |

Pin Descriptions

| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| 1 | Antenna damping drive | An antenna damping current flows when the RF AGC voltage (pin 2) reaches $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{D}}$. |  |
| 2 | RF AGC | Used to control the FET second gate. |  |
| 3 | F.E.GND |  |  |
| 4 | OSC | Oscillator connection |  |
| 7 | AM OSC | AM first oscillator <br> This circuit can oscillator up to the SW band. <br> An ALC circuit is included. |  |

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| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 8 \\ & 9 \end{aligned}$ | Noise AGC sensitivity AGC adjustment | After setting up the medium field (about $50 \mathrm{~dB} \mu$ ) sensitivity with the noise sensitivity setting pin (pin 8), set the weak field (about 20 to $30 \mathrm{~dB} \mu$ ) sensitivity with the AGC adjustment pin (pin 9) |  |
| $\begin{aligned} & 11 \\ & 12 \end{aligned}$ | Memory circuit connection | Recording circuit used during noise canceller operation. |  |
| 13 | Pilot input | Pin 13 is the PLL circuit input pin. |  |
| 14 | N.C, MPX, MRC, GND | Ground for the N.C., MPX, and MRC circuits. |  |

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| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 15 \\ & 16 \end{aligned}$ | MPX output (left) MPX output (right) | Deemphasis <br> $50 \mu \mathrm{~s}: 0.015 \mu \mathrm{~F}$ <br> $75 \mu \mathrm{~s}: 0.022 \mu \mathrm{~F}$ |  |
| 17 | Pilot canceller signal output | Adjustment is required since the pilot signal level varies with the sample-to-sample variations in the IF output level and other parameters. |  |
| 18 | Pilot canceller signal output | Pin 18 is the output pin for the pilot canceller signal. |  |

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Pin No. | Function |
| :---: |
| Separation |
| adjustment pin |
| PHASE COMP. |
| PHASE COMP. |
| VCO |

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| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| 27 | MRC control voltage time constant | The MRC detector time constant is determined by a $100 \Omega$ resistor and C2 when discharging and by the $2-\mu \mathrm{A}$ current and C 2 when charging. |  |
| 28 | SNC control input | The sub-output is controlled by a 0 to 1-V input. | A13572 |
| 29 | HCC control input | The high band frequency output is controlled by a 0 to $1-\mathrm{V}$ input. <br> It can also be controlled by the MRC output. <br> Use a resistor of at least $100 \mathrm{k} \Omega$ when controlling with the pin 32 FM S-meter signal. |  |

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| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 34 \\ & 35 \\ & 36 \\ & 37 \end{aligned}$ | AGC <br> QD output <br> QD input <br> $\mathrm{V}_{\text {REF }}$ | -The resistor $\mathrm{R}_{1}$ determines the width of the band muting function. Increasing the value of $\mathrm{R}_{1}$ narrows the band. <br> Reducing the value of $R_{1}$ widens the band. <br> -Null voltage When tuned, the voltage between pins 34 and $37, \mathrm{~V}_{34-37}$, will be 0 V . The band muting function turns on when $\left\|V_{34-37}\right\| \geq 0.7 \mathrm{~V}$. $V_{37}=4.9 \mathrm{~V}$ |  |
| 38 | FM SD ADJ | A 130- $\mu \mathrm{A}$ current flows from pin 38 and, in conjunction with the external resistance R, determines the comparison voltage. |  |
| 39 | Keyed AGC AM stereo buffer | The keyed AGC operates when the voltage created by dividing the pin 24 S-meter output voltage by the 6.4 and $3.6 \mathrm{k} \Omega$ resistors becomes lower than the voltage determined by the resistor between pin 39 and ground. <br> This pin also is used as the AM stereo IF buffer pin. |  |

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Pin No. | Function |
| :--- |
| HCC capacitor |
| Pilot detector |
| AM L.C. pin |
| The HCC frequency characteristics |
| are determined by the external |
| capacitor connected at this pin. |
| Inserting a 1-M 2 |
| pin resistor between |
| to mono mode. |

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| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| 44 | IF AGC | G1; Used for time constant switching during seeks. <br> - Reception $\tau=2.2 \mu \mathrm{~F} \times 300 \mathrm{k} \Omega$ <br> - Seek $\tau=2.2 \mu \mathrm{~F} \times 10 \Omega$ <br> The external capacitors are connected to $\mathrm{V}_{\mathrm{Cc}}$. <br> This is because the IF amplifier operates referenced to $\mathrm{V}_{\mathrm{CC}}$. |  |
| 45 | IF output | The IF amplifier load |  |
| 46 | AM antenna damping drive output Wide band AGC input | $\mathrm{I} 46=6 \mathrm{~mA}$ (maximum) <br> This is the antenna damping current. |  |

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Pin No. \begin{tabular}{l}
Function <br>
FM muting on level <br>
adjustment

 IF input 

Modify the value of the external <br>
resistor to adjust the muting on <br>
lever
\end{tabular}

| Pin No. | Function | Description | Equivalent circuit |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 53 \\ & 56 \end{aligned}$ | IF amplifier output IF amplifier input | - Input and output pin or the first IF amplifier <br> - Inverting amplifier $\mathrm{V} 56=2 \mathrm{~V}$ <br> Input impedance: $\mathrm{R}_{\mathrm{IN}}=330 \Omega$ $\mathrm{V} 53=5.3 \mathrm{~V}$ <br> Output impedance <br> $R_{\text {OUT }}=330 \Omega$ |  |
| $\begin{aligned} & 54 \\ & 49 \end{aligned}$ | Mixer output: $130 \mu \mathrm{~A}$ Mixer input | The mixer coil connected to the pin 54 mixer output must be wired to $\mathrm{V}_{\mathrm{CC}}$ (pin 40). <br> The pin 49 mixer input impedance is $330 \Omega$ |  |
| 55 58 | W-AGC IN <br> AM SD ADJ <br> N-AGC IN Muting attenuation adjustment pin | Pins 55 and 58 include built-in DC cut capacitors. <br> The AGC on level is determined by the values of the capacitors C1 and C2. <br> Pin 55 functions as the SD sensitivity adjustment pin in AM mode. <br> The output current 155 is $50 \mu \mathrm{~A}$, and V55 varies depending on the value of the external resistor. The SD function operates by comparing V55 with the S-meter voltage. |  |

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Block Diagram


## AC Characteristics Test Circuit



Test Conditions

| Parameter | Symbol | Switch states |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | SW1 | SW2 | SW3 | SW4 | SW5 | SW6 | SW7 | SW8 | SW9 | SW10 |
| Current drain | $\mathrm{ICCO}^{\text {-FM }}$ | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Demodulation output | $\mathrm{V}_{\mathrm{O}}$-FM | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Pin 31 demodulation output | $\mathrm{V}_{\mathrm{O}}$-FM31 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Channel balance | CB | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Total harmonic distortion (FM) | THD-FMmono | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Signal-to-noise ratio: IF | S/N-FM IF | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| AM suppression ratio: IF | AMR IF | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Muting attenuation | Att-1 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
|  | Att-2 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
|  | Att-3 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Separation | Separation | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Stereo on level | ST-ON | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Stereo off level | ST-OFF | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Main total harmonic distortion | THD-Main L | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Pilot cancellation | PCAN | ON | b | OFF | b | - | ON | OFF | OFF | OFF/ON | - |
| SNC output attenuation | AttSNC | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| HCC output attenuation 1 | AttHCC-1 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| HCC output attenuation 2 | AttHCC-2 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Input limiting voltage | Vi-lim | ON | b | OFF | b | - | ON | OFF | OFF | ON | ON |
| Muting sensitivity | Vi-mute | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| SD sensitivity 1 | SD-sen1 FM | ON | b | OFF | b | OFF | OFF | OFF | OFF | ON | - |
| SD sensitivity 2 | SD-sen2 FM | ON | b | OFF | b | ON | OFF | OFF | OFF | ON | - |
| IF counter buffer output | $\mathrm{V}_{\text {IFBUFF-FM }}$ | ON | b | OFF | b | OFF | OFF | OFF | OFF | ON | - |
| Signal meter output (FM) | $V_{S M} \mathrm{FM}-1$ | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
|  | $V_{\text {SM }}$ FM-2 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
|  | $V_{\text {SM }}$ FM-3 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
|  | $V_{\text {SM }}$ FM-4 | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Muting bandwidth | BW-mute | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| Mute drive output | $\mathrm{V}_{\text {MUTE-100 }}$ | ON | b | OFF | b | - | ON | OFF | OFF | ON | - |
| N-AGC on input | $\mathrm{V}_{\text {NAGC }}$ | ON | a | ON | b | - | ON | OFF | OFF | - | - |
| W-AGC on input | $V_{\text {WAGC }}$ | ON | a | ON | b | - | ON | OFF | OFF | - | - |
| Conversion gain | A.V | ON | a | ON | b | - | ON | OFF | OFF | - | - |
| Oscillator buffer output | Voscbufffm | ON | a | ON | b | - | ON | OFF | OFF | - | - |
| Gate time 1 | $\tau$ GATE1 | ON | - | OFF | a | - | ON | OFF | OFF | - | - |
| Noise sensitivity | SN | ON | - | OFF | a | - | ON | OFF | OFF | - | - |
| NC effect | SN-NC | ON/OFF | - | OFF | a | - | ON | OFF | OFF | - | - |
| MRC output | $\mathrm{V}_{\text {MRC }}$ | ON | - | OFF | b | - | ON | OFF | OFF | - | - |
| MRC operating level | MRC-ON | ON | - | OFF | b | - | ON | OFF | OFF | - | - |
| Practical sensitivity | S/N-30 | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Detection output | $\mathrm{V}_{\mathrm{O}}$-AM | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Pin 31 detection output | $\mathrm{V}_{\mathrm{O}}$-AM31 | OFF | - | OFF | b | ON | ON | - | - | - | - |
| AGC F.O.M. | $\mathrm{V}_{\text {AGC-FOM }}$ | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Signal-to-noise ratio | S/N-AM | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Total harmonic distortion (AM) | THD-AM | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Signal meter output (AM) | $\mathrm{V}_{\text {SM }} \mathrm{AM}-1$ | OFF | - | OFF | b | ON | ON | - | - | - | - |
|  | $\mathrm{V}_{\text {SM }} \mathrm{AM}-2$ | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Oscillator buffer output | $V_{\text {OSCBUFF AM-1 }}$ | OFF | - | OFF | b | ON | ON | - | - | - | - |
| Wide band AGC sensitivity | W-AGCsen 1 | OFF | - | OFF | b | ON | ON | - | - | - | - |
|  | W-AGCsen 2 | OFF | - | OFF | b | ON | ON | - | - | - | - |
| SD sensitivity | SD-sen1 AM | OFF | - | OFF | b | OFF | OFF | - | - | - | - |
|  | SD-sen2 AM | OFF | - | OFF | b | OFF | OFF | - | - | - | - |
| IF buffer output | $\mathrm{V}_{\text {IFBUFF-AM }}$ | OFF | - | OFF | b | OFF | OFF | - | - | - | - |

## LA1787M

## Usage Notes

## 1. Notes on $\mathrm{V}_{\mathrm{CC}}$ and Ground

| Pin 40 | V CC for the FM IF, AM, NC, MPX, and MRC blocks |
| :---: | :--- |
| Pin 25 | Ground for the FM IF and AM blocks |
| Pin 14 | Ground for the NC, MPX, and MRC blocks |
| Pin 61 | V $_{\text {CC }}$ for the FM front end, AM first mixer, and first oscillator blocks |
| * Pin 6 | V ${ }_{\text {CC }}$ for the FM front end and AGC blocks, and the AM/FM switching pin |
| Pin 3 | Ground for the FM front end, first mixer, and first oscillator blocks |

*: When applying the $\mathrm{V}_{\mathrm{CC}}$ voltage to pin 6 , that voltage must not exceed the pin 40 and pin $61 \mathrm{~V}_{\mathrm{CC}}$ voltages.
(This condition must be checked carefully when first applying the pin 6 voltage.)

## 2. Notes on AM Coil Connection

The $\mathrm{V}_{\mathrm{CC}}$ used for the first oscillator coil connected to pin 7 must be at the same potential as pin 61 .
Connect to the IFT connected with pin 45 , and to the MIX coil connected with pin $54 . \mathrm{V}_{\mathrm{CC}}$ must be at the same potential as $\operatorname{pin} 40$.

## 3. AM/FM Switching

Pin 6 is also used as the FM front end and RF AGC $V_{C C}$


| Pin 6 voltage | Mode |
| :---: | :---: |
| 8 | FM |
| OPEN | AM |

## LA1787M Overview

## 1. Notes on the LA1781M, LA1784M, and LA1787M

The LA1784M is a version of the LA1781M that uses an external oscillator circuit, and has the same characteristics as the LA1781M.
The LA1787M is a version of the LA1784M that features improved characteristics.


## LA1787M

## 2. Modified circuits

The following characteristics have been improved over those of the The LA1784M.

- The AM adjacent channel interference characteristics ( $\Delta 40 \mathrm{kHz}$ ) have been improved.
- The AM S-meter curve slope has been increased.
- The FM separation temperature characteristics have been improved.
- The stereo indicator sensitivity has been improved.
- The FM oscillator circuit has been omitted.
(1) AM interference characteristics improvement

The second signal interference and suppression have been improved for adjacent channels ( $\pm 40 \mathrm{kHz}$ ) by increasing the AM second mixer input dynamic range.
(2) The AM S-meter curve slope has been increased.

The slope of the AM S-Meter curve has been increased from that of the LA1781M and LA1784M.

(3) FM separation temperature characteristics improvement

The temperature characteristics have been improved, the amount of change in the separation due to drift when at power on has been stabilized. This makes it easier to adjust the separation.



## LA1787M

(4) Stereo indicator sensitivity improvement

The stereo indicator sensitivity (on/off) is equivalent to that of the LA1780M

|  | Stereo on level | Stereo off level |
| :---: | :---: | :---: |
| LA1781M/1784M | $4.1 \%$ | $3.1 \%$ |
| LA1787M/1780M | $2.6 \%$ | $1.6 \%$ |
| (Typical value) |  |  |

*: The pilot level such that the stereo indicator goes on or off for a 10.7 MHz unmodulated IF input.
(5) FM oscillator circuit removed

The internal FM oscillator circuit provided in the LA1781M has been removed. The FM oscillator level can be adjusted by constructing an external circuit block.
*: However, this requires 4 more external parts than the LA1781M: 1 transistor and 3 resistors/capacitors.


LA1787M/1784M FM OSC

## 3. Gain distribution

The table below shows the gain distribution of the LA1780M, LA1784M, and LA1787M. (These are measured values.) Compared to the LA1784M, the total gain is lower.

|  | 1st MIX (10.7) | 1st IF (10.7) | 2nd MIX (450) | 2nd IF (450) |
| :---: | :---: | :---: | :---: | :---: |
| LA1780M | 10 dB | 3.3 dB | 3.2 dB | 69 dB |
| LA1784M | 7.5 dB | 13 dB | 7 dB | 66 dB |
| LA1787M | 7.5 dB | 3.5 dB | 8.6 dB | 67 dB |

[^0]4. Changes to applications

Component values that change from LA1781M/LA1784M applications
(Since the total AM gain has changed in the LA1787M)

- AM SD adjustment resistor (pin 55): Because Vsm is higher.
- AM level adjustment resistor (pin 31): Since the post-detection audio amplifier gain is higher than in the LA1781M and LA1784M, the output level is also higher. This resistor must be changed to match the set value.
- AM mixer coil (pin 54), IFT coil (pin 45) damp resistor: Since the IF block gain is increased, the mixer (pin 54) and IFT (pin 45) coil damping must be adjusted.
- Separation adjustment resistor (pin 19): Since an internal $4 \mathrm{k} \Omega$ resistor has been added to the pin 19 input circuit to improve the separation temperature characteristics, the value of the external resistor must be reduced from that used with the LA1780M, LA1781M, and LA1784M. (See the following page.)



## Functions

## 1. Notes on the FM Front End

Notes on interference rejection characteristics

- Intermodulation characteristics

The LA1787M applies two high-band AGC functions to prevent IM (the generation of intermodulation). These are the narrow AGC (pin 58: mixer input detection type) and the wide AGC (for the pin 55 input), and this results in the antenna frequency characteristics shown in figure 2. The levels at which the AGC functions turn on are determined by the capacitors attached at pins 55 and 58.


Fig. 2

## LA1787M

- Notes on second-channel attenuation suppression

Keyed AGC (3D AGC) is a technique for achieving good characteristics for both intermodulation and secondchannel attenuation at the same time. When the desired signal is faint or nonexistent, the high-band AGC level will be essentially 0 , and as a result automatic tuning may malfunction and blocking oscillation may occur in the presence of strong interfering stations. Keyed AGC helps resolve these problems.
This 3D AGC technique uses information that has the following three frequency characteristics and is a unique Sanyo-developed system for determining the high-band AGC level.

RF and ANT circuit information: Mixer input AGC
Mixer circuit information: Mixer output AGC
CF selectivity information: S-meter output

- 3D AGC Features

| Feature | Merit |
| :--- | :--- |
| Only the narrow AGC sensitivity (operation at $\Delta \mathrm{f}<1.5 \mathrm{MHz}$ ) is <br> controlled by the field strength of the desired station. | • Effective in resolving second-channel attenuation problems. |
| The narrow AGC sensitivity is controlled by a voltage $\left(\mathrm{V}_{23}\right)$ that is <br> under 0.5 V. | • Allows effective resolution of second-channel attenuation problems without <br> degrading three-signal characteristics. |
| The wide AGC can operate even when $\mathrm{V}_{23}=0$ (when the desired <br> station is not present). | • Seek operations may stop incorrectly due to the occurrence of <br> intermodulation. <br> - It is possible to prevent the occurrence of intermodulation in the RF tuning <br> circuit and antenna in the presence of strong interfering stations, and <br> blocking oscillation due to AGC operation can be prevented. |
| The narrow and wide AGC sensitivities can be set independently. <br> (See figure 3 and 4.) | - Settings can be optimized for the field conditions. |
| The system has two AGC systems: narrow and wide AGC. <br> (See figure 5.) | • Since the narrow AGC operates for the desired station and adjacent <br> stations, the wide AGC sensitivity can be lowered and AGC malfunction <br> due to local oscillator signal can be prevented. |




## LA1787M

3D AGC Sensitivity Characteristics


A12075
Fig. 6

3D AGC Sensitivity $-\Delta f, V_{23}$ characteristics

- The wide AGC sensitivity is determined by the antenna and RF circuit selectivity, regardless of $\mathrm{V}_{23}$.
- The narrow AGC sensitivity is determined by the following.

The total selectivity of the antenna, RF circuit, and mixer when $V_{23} \geq 0.5 \mathrm{~V}$
The above selectivity and $\mathrm{V}_{23}$ when $\mathrm{V}_{23}<0.5 \mathrm{~V}$

- The improvement in the second-channel attenuation corresponds to the area occupied by the narrow AGC in the total AGC sensitivity area.
Figure 8 on the next page shows the actual operation of the circuit.

$\mathrm{f}_{\mathrm{D}}=98.1 \mathrm{MHz}$ Second-channel pad


Fig. 7

Notes on 3D AGC (Keyed AGC)


Fig. 8

- The antenna damping current from the pin due to the pin diode flows when the V 2 pin reaches the $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{BE}}$ level.
- The narrow AGC operates as follows.

When pin V39 > pin V24: The narrow AGC turns off.
When pin V39 < pin V24: The narrow AGC turns on.

## LA1787M

- The LA1787M includes two AGC circuits in its front end block.
- Antenna input limiter using a pin diode.
- FET second gate control

The AGC input pin is pin 59, and the AGC circuit turns on when a signal of about 30 mVrms is input.

AGC activation
The pin diode drive circuit turns on when $\mathrm{V}_{\mathrm{CC}}-\mathrm{V} 2$ is greater than or equal to about 1 V , and input limitation is applied to the antenna circuit. In application circuits, there will be an attenuation of about 30 to 40 dB . Next, when an adequate current flows in the antenna attenuator pin diode, the inductance falls, the FET second gate voltage drops, the FET gm falls, and the AGC operates. The recommended FET is the Sanyo 3SK263, which is an enhancement-type MOSFET. Therefore, full AGC is applied when the voltage, $\mathrm{V}_{\mathrm{G} 2-\mathrm{S}}$, between the second gate and the source is 0 . Note that if a depletion-type MOSFET is used, AGC will not be applied unless $\mathrm{V}_{\mathrm{G} 2-\mathrm{S}}$ is less than 0 .


- Mixer

The mixer circuit in this IC is a double-balanced mixer with both balanced input and balanced output.
Input circuit type
Emitter input
Input impedance: $25 \Omega$
Due to optimized device geometry, emitter current, the bias, this IC achieves the following performance.

Mixer input usable sensitivity: $15 \mathrm{~dB} \mu$
Mixer input IMQS: $90.5 \mathrm{~dB} \mu$
(For an oscillator level of 200 mVrms )

* The mixer input IMQS is defined as:
$\mathrm{fr}=98.8 \mathrm{MHz}$, no input
fu1 $=98.8 \mathrm{MHz}, 1 \mathrm{kHz}, 30 \%$ modulation -
fu2 $=99.6 \mathrm{MHz}$, no modulation

The interference 1 and 2 input levels such that generated intermodulation output signal-to-noise ratio becomes 30 dB when an interference signal with the same level as the mixer input is input, and distortion occurs in the mixer.


Fig. 10

## - Oscillator

Figure 11 shows the type of oscillator circuit used in this IC. It includes both an oscillator and an oscillator buffer.


Fig. 11

- Figure 12 shows the type of FM first IF amplifier used in this IC. It is a differential single-stage amplifier.


A12079
Fig. 12
Specifications
Input impedance: $330 \Omega$ Output impedance: $330 \Omega$
Gain: 20 dB

## LA1787M

2. FM IF

- Notes on the FM SD and SD adjustment

The figure below presents an overview of the FM SD and the IF count buffer.


Figure 14 shows the relationship between the FM SD, the IF count buffer output, the S-meter, and the muting drive output.


Fig. 14

## LA1787M

- Transient response characteristics during automatic tuning

The transient characteristics for SD and IF count buffer on/off operation are determined by the time constants of the RC circuits attached to the following pins.
(1) Muting time constant: pin 33
(2) S-meter time constant: pin 24
(3) AFC time constant: pin 34

There are two points that require consideration when using fast tuning.
(1) The SD time constant due to the S-meter time constant

Since the current I24 (pin 24) varies with the field strength, the time constant also changes. There is no hysteresis in the comparator.
If a smaller value is used for C 24 , you must select a value for C such that the AGC does not become unstable when the pin 24 voltage is used for keyed AGC.


Fig. 15
(2) The SD time constant due to the pin 33 muting voltage time constant

The changes in volume due to field fluctuation during weak field reception can be made smoother by setting the attack and release times during soft muting operation.



## LA1787M

However, when testing this stop sensitivity, note that when checking the waveform on the IF count buffer output (pin 23), there are cases, such as that shown below, where current in the test system may be seen as flowing to ground and cause oscillation that causes the IF count buffer output to go to the output state.


The 10.7 MHz feeds back through ground.
Fig. 18
A12081

- FM Muting control pin (pin 47) (R47: $30 \mathrm{k} \Omega$ variable resistor)

The -3 dB limiting sensitivity can be adjusted with R47.


- FM muting attenuation adjustment (pin 58)

The muting attenuation can be switched between the three levels of $-20,-30$, and -40 dB by the resistor inserted between pin 58 and ground. (Note that the exact values depend on the total tuner gain.)
The noise convergence with no input is determined by the pin 58 voltage.


Fig. 20

| R58 | Mute ATT |
| :---: | :---: |
| Open | -20 dB |
| $200 \mathrm{k} \Omega$ | -30 dB |
| $30 \mathrm{k} \Omega$ | -40 dB |

The attenuation can be set by making R33 smaller as listed in the table above.


- FM muting off function

Forcing this pin to the ground level turns muting off.


Fig. 24

## LA1787M

- Hall detection

The Hall detection function detects the level of the pin 36 quadrature input signal and then applies peak detection to that result. The result is output from pin 33. This circuit has three effects.
(1) It assures that muting will be applied for weak inputs with an antenna input of under $5 \mathrm{~dB} \mu$. The amount of attenuation is referenced to an antenna input of $60 \mathrm{~dB} \mu, \mathrm{fm}=1 \mathrm{kHz}$, and a 22.5 kHz dev output, and is variable from 10 dB to 40 dB when there is no input. Thus one feature of this circuit is that the weak input noise attenuation and the -3 dB limiting sensitivity for over $5 \mathrm{~dB} \mu$ inputs can be set independently.

(2) When the pin 36 quadrature input is a saturated input, the pin 36 noise level ( Va ) is detected and a peak-hold function is applied to pin $33(\mathrm{Vb})$ for locations rapid field strength variations and severe multipath occurs for fields that result in an antenna input level of over $5 \mathrm{~dB} \mu$.


Fig. 26
(3) Unique features

One unique feature of the LA1784M is that if there are adjacent stations such that $f_{1}=98.1 \mathrm{MHz}$ and $f_{2}=$ 97.9 MHz , a search operation will not stop at 98.0 MHz . Since $\mathrm{V}_{\mathrm{AFC}}=0 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{SM}}=3.6 \mathrm{~V}$ at 98.0 MHz in the situations shown in figure 27 and 28, even though Hall detection would normally not operate and SD would be high, in this IC the Hall detection circuit will operate, $\mathrm{V}_{\text {Mute }}$ will be set to 1.2 V (over 0.7 V ) and the SD signal will go low, thus preventing incorrect stopping of the search.



- Notes on the quadrature input level

When a strong field is being received the quadrature signal input (pin 36) requires a 200 mV rms input, and the detection transformer and the damping resistor between pins 36 and 37 must be designed.
(We recommend the Sumida SA-208 transformer and a $10 \mathrm{k} \Omega$ resistor between pins 36 and 37.)
When the pin 36 input level falls below 160 mV rms , the Hall detection circuit operates and the pin 33 mute drive output voltage increases. Therefore, when pin 36 input is from 160 to under 200 mV rms during strong field reception, the muting circuit may or may not operate due to sample-to-sample variations between individual ICs. Furthermore, the SD function may not operate, and the audio output level may be reduced. Incorrect operation due to sample-to-sample variations and temperature characteristics can be prevented by keeping the pin 36 voltage at 200 mVrms or higher.



## LA1787M

| $\mathrm{R}_{36-37}$ | Detector output <br> MPX OUT <br> Vo | Pin 36 AC level |
| :---: | :---: | :---: |
| Open | 330 mVrms | 235 mVrms |
| $10 \mathrm{k} \Omega$ | 280 mVrms | 200 mVrms |

- Band Muting Adjustment Procedure

The muting bandwidth can be modified as shown in figure 31 with the resistor $\mathrm{R}_{\mathrm{BW}}$ between pin 34 and 37 .

3. AM

- AM AGC system

The LA1787M RF AGC circuit takes its input from three sources: the WIDE AGC pin (pin 46), the MIDDLE AGC pin (pin 49) and NARROW AGC. There is also an IF AGC circuit.



The wide band AGC circuit in this IC has the frequency characteristics shown above. The pin 46 input frequency characteristics are identical to those of the RF amplifier gate. This AGC circuit serves to prevent distortion at the FET input when a strong signal is applied to the antenna circuit. The level at which the AGC circuit turns on can be adjusted to an arbitrary level with the wide band AGC adjustment resistor. A delayed AGC on level can be handled by reducing the value of the adjustment resistor.


## LA1787M

- Notes on AM SD (pin 26) and the SD adjustment pin

SD and the IF buffer are operated by comparing the S-meter level (V24) and the 5 V reference voltage as shown in figure 36 .


Fig. 36
A12085

Figure 37 shows the relationship between the AM SD, the IF count buffer, and the S-meter.


Fig. 37

## LA1787M

- AM high band cut and detector output level adjustment methods

The pin 31 AM and FM tuner output has an impedance of $10 \mathrm{k} \Omega$ in AM mode and a few tens of Ohms in FM mode. Therefore, R31 is used to lower the AM detector output level and C31 determines the AM high band frequency characteristics.


Fig. 39

- AM stereo system pins


Fig. 40

- AM low band cut adjustment method

The AM low band frequency characteristics can be adjusted with C 42 , which is inserted between pin 42 and $\mathrm{V}_{\mathrm{CC}}$. Since the detector is designed with $\mathrm{V}_{\mathrm{CC}}$ as the reference, C 42 must be connected to $\mathrm{V}_{\mathrm{CC}}$.


Fig. 41

4. Noise Canceler Block

- The noise canceler input (pin 30) has an input impedance of about $50 \mathrm{k} \Omega$. Check the low band frequency characteristics carefully when determining the value of the coupling capacitor used. Note that $f_{C}$ will be about 3 Hz when a $1 \mu \mathrm{~F}$ capacitor is used in the application.
- Pins 8 and 9 are used to set the noise detector sensitivity and the noise AGC. It is advisable to first set the noise sensitivity for a medium field (an antenna input of about $50 \mathrm{~dB} \mu$ ) with pin 8 (the noise sensitivity setting pin), and then set the AGC level for a weak field ( 20 to $30 \mathrm{~dB} \mu$ ) with pin 9 (the AGC adjustment pin). If the noise sensitivity is increased, the AGC will become more effective but, inversely, the weak field sensitivity will be reduced.

Noise canceler 10 kHz overmodulation malfunction may be a problem. In particular, when an overmodulated signal is input, the noise canceler may, in rare cases, malfunction. This is due to the fact that the IF detector output has a waveform of the type shown in figure 43 due to the bands of the IF ceramic filters as shown below. (Here, the antenna input is $60 \mathrm{~dB} \mu$, the ceramic filters are $150 \mathrm{kHz} \times 1$ and $180 \mathrm{kHz} \times 2, \mathrm{f}=10 \mathrm{kHz}, 180 \mathrm{kHz} \mathrm{dev}$.) The noise canceler reacts to the spikes (whiskers) generated due to this overmodulation, which results in distortion to the audio output. (The spike components due to overmodulation occur due to the bands of the ceramic filters in the tuner.) The following describes a method for resolving this problem. This incorrect operation due to overmodulation is prevented by removing the spike components due to this overmodulation with a low-pass filter consisting of a $1 \mathrm{k} \Omega$ resistor and a 2200 pF capacitor shown in figure 44 . However, note that the FM separation characteristics in the high band and the AM frequency characteristics will change.


Fig. 44

## LA1787M

5. Multiplexer Block

- HCC (high cut control) frequency characteristics (pin 41)

When the HCC function operates, the frequency characteristics of the output signal are determined by the capacitance of the external capacitor connected to pin 41.


Fig. 45

$$
\mathrm{f}_{\mathrm{C}}=\frac{1}{2 \pi \times \mathrm{C} \times 20 \mathrm{k} \Omega}[\mathrm{~Hz}]
$$



Fig. 46

A12091


- Pilot canceler adjustment (pins 17 and 18)


Fig. 48

The pilot canceler signal waveform (pin 19) is a 19 kHz signal that contains no third harmonic as shown in figure 48. Since this signal has the same phase as the pilot signal, no capacitor is required between pin 18 and ground. Since it has no third harmonic component, excellent pilot cancellation can be acquired in both the left and right channels by adjusting with a variable resistor.

## LA1787M

- Separation adjustment (pin 19)


Fig. 49

The separation is adjusted by modifying the input level to the subdecoder with the variable resistor connected to pin 19. Since only the sub-modulation level is changed by changing the variable resistor setting, the monaural (main) output level is not changed. Furthermore, degradation of high band separation in the decoder can be avoided if the impedance of the external capacitor (C) in the subchannel frequency band ( 23 to 53 kHz ) is made sufficiently smaller than the variable resistor.
6. MRC Circuit


Fig. 50
(1) When there is no AC noise on pin 32

$$
\mathrm{V}_{24}=\mathrm{V}_{27}-\mathrm{V}_{\uparrow} \mathrm{BE}
$$

$$
\mathrm{Q}_{\mathrm{MRC}}
$$

V 27 is about 2.5 V when the antenna input is 60 dB or higher.
(2) Since the MRC noise amplifier gain is fixed, the MRC circuit is adjusted by reducing the AC input level.

(3) The MRC attack and release are determined by C27 on pin 27.

Attack: $7 \mu \mathrm{~A} \cdot \mathrm{C} 27 \rightarrow 2 \mu \mathrm{~A} \cdot \mathrm{C} 27$
Release: $500 \Omega \cdot \mathrm{C} 27 \rightarrow 100 \Omega$

Notes on the Noise Canceler
The noise canceler characteristics have been improved by implementing the circuit that determines the gate time in logic. Since the time constant in earlier noise cancelers was determined by an RC circuit such as that shown in figure 52, the rise time shown in figure 53 was influenced by the values of the resistor and capacitor used. As a result the noise exclusion efficiency was reduced by this delay in the rise time. In the LA1787M, this rise time was shortened by implementing the circuit that determines the gate time in logic, allowing it to reliably exclude noise.


A11771
Fig. 52


A11772

## LA1787M

Gain Distribution (FM)
This section investigates the gain in each block in the LA1787M when the Sanyo recommended circuits are used.
(Test conditions)
Ambient temperature: $26^{\circ} \mathrm{C}$
Antenna and mixer input frequency: 98.1 MHz
First and second IF input frequency: 10.7 MHz
The input levels when $\mathrm{V}_{\mathrm{SM}}=2 \mathrm{~V}$ will be as follows.
ANT IN: $19 \mathrm{~dB} \mu$
MIX IN: $30 \mathrm{~dB} \mu$
1st IF IN: $42 \mathrm{~dB} \mu$
2nd IF IN: $60 \mathrm{~dB} \mu$
When the gains for each block are determined according to the above, the results are as follows.
RF GAIN: 11 dB
MIX GAIN: 12 dB
1st IF GAIN: 18 dB


Fig. 54

## LA1787M

(AM)
This section investigates the gain in each block in the LA1787M when the Sanyo recommended circuits are used.
(Test conditions)
Ambient temperature: $26^{\circ} \mathrm{C}$
Antenna and mixer input frequency: 1 MHz
First and second mixer input frequency: 10.7 MHz
Second IF input frequency: 450 kHz

The gains at each stage will be as follows.
RF Gain (ANT IN-pin62): 17 dB
1st MIX Gain (pin62-pin56): 8 dB
1st IF Gain (pin55-pin53): 15 dB


Fig. 55

Input Circuits for Each Stage
[FM]

- Mixer input


- First IF input

- IF input


A11777
[AM]

- First mixer input


$$
\mathrm{fr}=\mathrm{RF}
$$

A11778

- Second mixer input

$\mathrm{fr}=10.71 \mathrm{MHz}$ (f2nd osc +0.45 MHz )
A11779
- Del input



## LA1787M

## Sample AM tuner Circuit with the LC72144 Used Together



A11782

|  |  | AM 1st IF | Step | FM IF |
| :---: | :---: | :---: | :---: | :---: |
| 1 | fosc 10.25 MHz | 10.7 MHz | $10 \mathrm{kHz}, 11 \mathrm{kHz}$ | 10.7 MHz |
| 2 | fosc 10.35 MHz | 10.8 MHz | $9 \mathrm{kHz}, 10 \mathrm{kHz}$ | 10.8 MHz |



## Crystal Oscillator Element

Kinseki, Ltd.
Frequency: 10.26 MHz
CL: 20 pF
Model No.: HC-49/U-S

## Coil Specifications

Sumida Electronics, Ltd.
[AM Block]
AM FILTEER (SA-1051)


AM IF1 (SA-264)


AM loading (SA-1062)


AM RF amplifier (RC875-222J)

[FM Block]
FM RF (SA-1060)


FM OSC (SA-1052)


FM DET (SA-208)


AM OSC (SA-359)


AM IF2 (SA-1063)


AM ANT IN (SA-1048)


FM ANT (SA-1061)


FM MIX (SA-266)


The Toko Electric Corporation
[AM Block]
AM FILTEER (A2861BIS-15327)


AM IF1 (7PSGTC-5001A)


AM loading (269ANS-0720Z)


AM RF amplifier (187LY-222)


FM RF (V666SNS-208AQ)


FM OSC (V666SNS-205APZ)


FM DET (DM600DEAS-8407GLF)


AM OSC (V666SNS-214BY)


AM IF2 (7PSGTC-5002Y)


AM ANT IN (385BNS-027Z)


FM ANT (V666SNS-209BS)


FM MIX (371DH-1108FYH)


## Coil Specifications

Sagami Elec Co., Ltd.
[AM Block]
AM FILTEER (000021055)


AM IF1 (000021057)


AM loading (000021061)


AM RF amplifier (000021063)

[FM Block]
FM RF (000021064)


FM OSC (000021066)


FM DET (010021075)


AM OSC (000021056)


AM IF2 (000021059)


AM ANT IN (000021062)


FM ANT (000021065)


FM MIX (000021067)






FM Gain Distribution (2)



First IF I/O Characteristics




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[^0]:    First mixer : No circuit changes from the LA1784M.
    First IF amplifier : Equivalent to the LA1780M circuit. (The gain is lower than that in the LA1781M and LA1784M.)
    Second mixer : The mixer circuit has been modified to improve adjacent channel suppression and interference.
    Second IF amplifier : Equivalent to the LA1780M circuit.

