

FEATURES

Stable over time and temperature

0.5% initial accuracy

1% accuracy over the full temperature range

Compatible with Type II or Type III compensation networks

Reference voltage: 1.225 V

Compatible with DOSA

Low power operation: <7 mA total

Wide voltage supply range

V_{DD1} : 3 V to 20 V

V_{DD2} : 3 V to 20 V

Bandwidth: 400 kHz

Isolation voltage: 2.5 kV rms

Safety and regulatory approvals (pending)

UL recognition: 2500 V rms for 1 minute per UL 1577

CSA Component Acceptance Notice #5A

VDE certificate of conformity

DIN V VDE V 0884-10 (VDE V 0884-10):2006-12

$V_{IORM} = 560$ V peak

Wide temperature range

−40°C to +125°C ambient operation

150°C maximum junction temperature

APPLICATIONS

Linear power supplies

Inverters

Uninterruptible Power Supply (UPS)

DOSA-compatible modules

Voltage monitors

GENERAL DESCRIPTION

The ADuM3190¹ is an isolated error amplifier based on Analog Devices, Inc., iCoupler® technology. The ADuM3190 is ideal for linear feedback power supplies. The primary side controllers of the ADuM3190 enable improvements in transient response, power density, and stability as compared to commonly used optocoupler and shunt regulator solutions.

Unlike optocoupler-based solutions, which have an uncertain current transfer ratio over lifetime and at high temperatures, the ADuM3190 transfer function does not change over its lifetime and it is stable over a wide temperature range of −40°C to +125°C.

Included in the ADuM3190 is a wideband operational amplifier for a variety of commonly used power supply loop compensation techniques. The ADuM3190 is fast enough to allow a feedback loop to react to fast transient conditions and overcurrent conditions. Also included is a high accuracy 1.225 V reference to compare with the supply output setpoint.

The ADuM3190 is packaged in a small 16-lead QSOP package for a 2.5 kV rms isolation voltage rating.

FUNCTIONAL BLOCK DIAGRAM

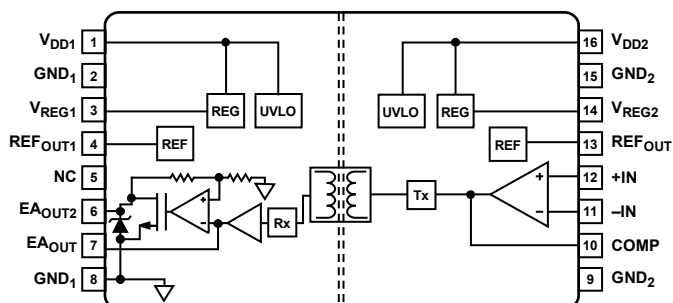


Figure 1.

¹ Protected by U.S. Patents 5,952,849, 6,873,065 and 7,075,329. Other patents pending.

Rev. 0

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TABLE OF CONTENTS

Features	1	Pin Configuration and Function Descriptions.....	7
Applications.....	1	Typical Performance Characteristics	8
General Description	1	Test Circuits.....	11
Functional Block Diagram	1	Applications Information	12
Revision History	2	Theory of Operation	12
Specifications.....	3	Accuracy Circuit Operation.....	12
Package Characteristics	4	Application Block Diagram	12
Regulatory Information.....	4	Setting the Output Voltage.....	13
Insulation and Safety Related Specifications	4	DOSA Module Application.....	14
Recommended Operating Conditions	4	DC Correctness and Magnetic Field Immunity	14
DIN V VDE V 0884-10 (VDE V 0884-10) Insulation		Insulation Lifetime	15
Characteristics	5	Packaging and Ordering Information	16
Absolute Maximum Ratings	6	Outline Dimensions.....	16
ESD Caution.....	6	Ordering Guide	16

REVISION HISTORY

2/13—Revision 0: Initial Version

SPECIFICATIONS

$V_{DD1} = V_{DD2} = 3\text{ V to }20\text{ V}$ for $T_A = T_{MIN}$ to T_{MAX} . All typical specifications are at $T_A = 25^\circ\text{C}$ and $V_{DD1} = V_{DD2} = 5\text{ V}$, unless otherwise noted.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
ACCURACY	$(1.225\text{ V} - EA_{OUT})/1.225\text{ V} \times 100\%$; see Figure 18				
Initial Error	$T_A = 25^\circ\text{C}$		0.25	0.5	%
Total Error	$T_A = T_{MIN}$ to T_{MAX}		0.5	1	%
OP AMP					
Offset Error		−5	±2.5	+5	mV
Open-Loop Gain		66	80		dB
Input Common-Mode Range		0.35		1.5	V
Gain Bandwidth Product			10		MHz
Common-Mode Rejection			72		dB
Input Capacitance			2		pF
Output Voltage Range	COMP pin	0.2		2.7	V
Input Bias Current			0.01		μA
REFERENCE					
Output Voltage	At 25°C , 0 mA to 1 mA load, $C_{REFOUT} = 15\text{ pF}$	1.215	1.225	1.235	V
	−40°C to +125°C, 0 mA to 1 mA load, $C_{REFOUT} = 15\text{ pF}$	1.213	1.225	1.237	V
Output Current	$C_{REFOUT} = 15\text{ pF}$	2.0			mA
UVLO					
Positive Going Threshold			2.8	2.96	V
Negative Going Threshold		2.4	2.6		V
EA_{OUT} Impedance	$V_{DD2} < \text{UVLO threshold}$ or $V_{DD1} < \text{UVLO threshold}$		High-Z		Ω
OUTPUT CHARACTERISTICS					
Output Gain	From COMP to EA_{OUT} , dc, 0.3 V to 2.4 V; and from COMP to EA_{OUT2} , dc, 0.4 V to 5.0 V	0.9	1.0	1.1	V/V
	From COMP to EA_{OUT} , 0.3 V to 2.4 V, ±3 mA; and from COMP to EA_{OUT2} , 0.4 V to 5.0 V, ±1 mA	2.34	2.6	2.86	V/V
Output −3 dB Bandwidth					
A and S Grades		100	200		kHz
B and T Grades		250	400		kHz
Output Voltage					
EA_{OUT}	±3 mA output				
Low				0.4	V
High		2.4	2.5		V
EA_{OUT2}	±1 mA output				
Low	$V_{DD1} = 4.5\text{ V to }5.5\text{ V}$		0.3	0.6	V
High	$V_{DD1} = 4.5\text{ V to }5.5\text{ V}$	4.8	4.9		V
Low	$V_{DD1} = 10\text{ V to }20\text{ V}$		0.3	0.6	V
High	$V_{DD1} = 10\text{ V to }20\text{ V}$	5.0	5.4		V
Noise	See Figure 18				
EA_{OUT}			1.7		mV rms
EA_{OUT2}			4.8		mV rms
POWER SUPPLY					
Operating Range, Side 1	V_{DD1}	3.0		20	V
Operating Range, Side 2	V_{DD2}	3.0		20	V
Power Supply Rejection	DC, $V_{DD1} = V_{DD2} = 3\text{ V to }20\text{ V}$	60			dB
Supply Current					
I_{DD1}	See Figure 18		1.4	2.0	mA
I_{DD2}	See Figure 18		2.9	5.0	mA

PACKAGE CHARACTERISTICS

Table 2.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
RESISTANCE Input-to-Output ¹	R _{I-O}		10 ¹³		Ω	
CAPACITANCE Input-to-Output ¹	C _{I-O}		2.2		pF	f = 1 MHz
Input Capacitance ²	C _I		4.0		pF	
IC JUNCTION-TO-AMBIENT THERMAL RESISTANCE 16-Lead QSOP	θ _{JA}		76		°C/W	Thermocouple located at center of package underside

¹ The device is considered a 2-terminal device; Pin 1 through Pin 8 are shorted together, and Pin 9 through Pin 16 are shorted together.

² Input capacitance is from any input data pin to ground.

REGULATORY INFORMATION

The ADuM3190 is pending approval by the organizations listed in Table 3. See Table 8 and the Insulation Lifetime section for recommended maximum working voltages for specific cross-isolation waveforms and insulation levels.

Table 3.

UL (Pending)	CSA (Pending)	VDE (Pending)
Recognized Under 1577 Component Recognition Program ¹ Single Protection, 2500 V rms Isolation Voltage, 16-Lead QSOP File E214400	Approved under CSA Component Acceptance Notice #5A Basic insulation per CSA 60950-1-03 and IEC 60950-1, 400 V rms (565 V peak) maximum working voltage File 205078	Certified according to DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12 ² Reinforced insulation, 560 V peak File 2471900-4880-0001

¹ In accordance with UL 1577, each ADuM3190 is proof tested by applying an insulation test voltage ≥ 3000 V rms for 1 sec (current leakage detection limit = 5 μA).

² In accordance with DIN V VDE V 0884-10, each ADuM3190 is proof tested by applying an insulation test voltage ≥ 1050 V peak for 1 sec (partial discharge detection limit = 5 pC). The asterisk (*) marked on the component designates DIN V VDE V 0884-10 approval.

INSULATION AND SAFETY RELATED SPECIFICATIONS

Table 4.

Parameter	Symbol	Value	Unit	Test Conditions/Comments
Rated Dielectric Insulation Voltage		2500	V rms	1-minute duration
Minimum External Air Gap (Clearance)	L(I01)	3.8 min	mm	Measured from input terminals to output terminals, shortest distance through air
Minimum External Tracking (Creepage)	L(I02)	3.1 min	mm	Measured from input terminals to output terminals, shortest distance path along body
Minimum Internal Gap (Internal Clearance)		0.017 min	mm	Insulation distance through insulation
Tracking Resistance (Comparative Tracking Index)	CTI	>400	V	DIN IEC 112/VDE 0303, Part 1
Isolation Group		II		Material Group DIN VDE 0110, 1/89, Table 1

RECOMMENDED OPERATING CONDITIONS

Table 5.

Parameter	Symbol	Min	Max	Unit
OPERATING TEMPERATURE BY MODEL ADuM3190A/ADuM3190B ADuM3190S/ADuM3190T	T _A	-40 -40	+85 +125	°C °C
SUPPLY VOLTAGES ¹	V _{DD1} , V _{DD2}	3.0	20	V
INPUT SIGNAL RISE AND FALL TIMES	t _R , t _F		1.0	ms

¹ All voltages are relative to their respective grounds.

DIN V VDE V 0884-10 (VDE V 0884-10) INSULATION CHARACTERISTICS

These isolators are suitable for reinforced isolation only within the safety limit data. Maintenance of the safety data is ensured by protective circuits. The asterisk (*) marking branded on the package denotes DIN V VDE V 0884-10 approval for a 560 V peak working voltage.

Table 6.

Description	Test Conditions/Comments	Symbol	Characteristic	Unit
Installation Classification per DIN VDE 0110 For Rated Mains Voltage ≤ 150 V rms For Rated Mains Voltage ≤ 300 V rms For Rated Mains Voltage ≤ 400 V rms			I to IV I to III I to II	
Climatic Classification			40/105/21	
Pollution Degree per DIN VDE 0110, Table 1			2	
Maximum Working Insulation Voltage		V_{IORM}	560	V peak
Input-to-Output Test Voltage, Method B1	$V_{IORM} \times 1.875 = V_{pd(m)}$, 100% production test, $t_{ini} = t_m = 1$ sec, partial discharge < 5 pC	$V_{pd(m)}$	1050	V peak
Input-to-Output Test Voltage, Method A After Environmental Tests Subgroup 1	$V_{IORM} \times 1.5 = V_{pd(m)}$, $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge < 5 pC	$V_{pd(m)}$	840	V peak
After Input and/or Safety Test Subgroup 2 and Subgroup 3	$V_{IORM} \times 1.2 = V_{pd(m)}$, $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge < 5 pC	$V_{pd(m)}$	672	V peak
Highest Allowable Overvoltage		V_{IOTM}	3500	V peak
Surge Isolation Voltage		V_{IOSM}	4000	V peak
Safety Limiting Values	$V_{PEAK} = 10$ kV; 1.2 μ s rise time; 50 μ s, 50% fall time Maximum value allowed in the event of a failure (see Figure 2)			
Case Temperature		T_S	150	$^{\circ}\text{C}$
Safety Total Dissipated Power		P_S	1.64	W
Insulation Resistance at T_S	$V_{IO} = 500$ V	R_S	$>10^9$	Ω

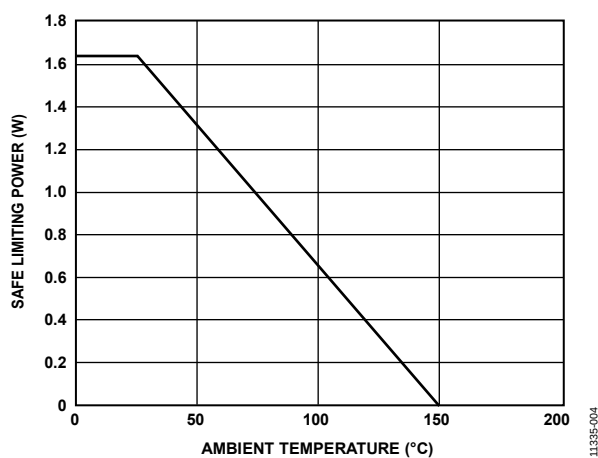


Figure 2. Thermal Derating Curve, Dependence of Safety Limiting Values on Case Temperature, per DIN V VDE V 0884-10

ABSOLUTE MAXIMUM RATINGS

$T_A = 25^\circ\text{C}$, unless otherwise noted.

Table 7.

Parameter	Rating
Storage Temperature (T_{ST}) Range	-65°C to $+150^\circ\text{C}$
Ambient Operating Temperature (T_A) Range	-40°C to $+125^\circ\text{C}$
Junction Temperature	-40°C to $+150^\circ\text{C}$
Supply Voltages	
V_{DD1}, V_{DD2}^1	-0.5 V to $+24\text{ V}$
V_{REG1}, V_{REG2}^1	-0.5 V to $+3.6\text{ V}$
Input Voltages ($+IN, -IN$)	-0.5 V to $+3.6\text{ V}$
Output Voltages	
$REF_{OUT}, COMP, REF_{OUT1}, EA_{OUT}$	-0.5 V to $+3.6\text{ V}$
EA_{OUT2}	-0.5 V to $+5.5\text{ V}$
Output Current per Output Pin	-11 mA to $+11\text{ mA}$
Common-Mode Transients ²	$-100\text{ kV}/\mu\text{s}$ to $+100\text{ kV}/\mu\text{s}$

¹ All voltages are relative to their respective grounds.

² Refers to common-mode transients across the insulation barrier. Common-mode transients exceeding the absolute maximum ratings may cause latch-up or permanent damage.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 8. Maximum Continuous Working Voltage¹

Parameter	Max	Unit	Constraint
WAVEFORM			
AC Voltage			
Bipolar	560	V peak	50-year minimum lifetime
Unipolar	1131	V peak	50-year minimum lifetime
DC Voltage	1131	V peak	50-year minimum lifetime

¹ Refers to continuous voltage magnitude imposed across the isolation barrier. See the Insulation Lifetime section for more details.

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

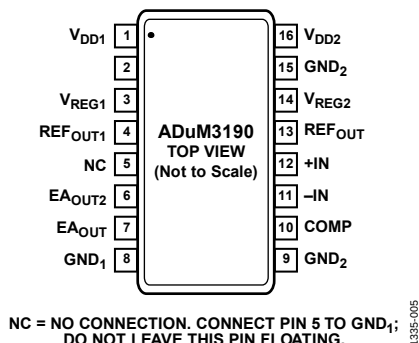


Figure 3. Pin Configuration

Table 9. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	V _{DD1}	Supply Voltage for Side 1 (3.0 V to 20 V). Connect a 1 μ F capacitor between V _{DD1} and GND ₁ .
2	GND ₁	Ground Reference for Side 1.
3	V _{REG1}	Internal Supply Voltage for Side 1. Connect a 1 μ F capacitor between V _{REG1} and GND ₁ .
4	REF _{OUT1}	Reference Output Voltage for Side 1. The maximum capacitance for this pin (C _{REFOUT1}) must not exceed 15 pF.
5	NC	No Connection. Connect Pin 5 to GND ₁ ; do not leave this pin floating.
6	EA _{OUT2}	Isolated Output Voltage 2, Open-Drain Output. Connect a pull-up resistor between EA _{OUT2} and V _{DD1} for current up to 1 mA.
7	EA _{OUT}	Isolated Output Voltage.
8	GND ₁	Ground Reference for Side 1.
9	GND ₂	Ground Reference for Side 2.
10	COMP	Output of the Op Amp. A loop compensation network can be connected between the COMP pin and the -IN pin.
11	-IN	Inverting Op Amp Input. Pin 11 is the connection for the power supply setpoint and compensation network.
12	+IN	Noninverting Op Amp Input. Pin 12 can be used as a reference input.
13	REF _{OUT}	Reference Output Voltage for Side 2. The maximum capacitance for this pin (C _{REFOUT}) must not exceed 15 pF.
14	V _{REG2}	Internal Supply Voltage for Side 2. Connect a 1 μ F capacitor between V _{REG2} and GND ₂ .
15	GND ₂	Ground Reference for Side 2.
16	V _{DD2}	Supply Voltage for Side 2 (3.0 V to 20 V). Connect a 1 μ F capacitor between V _{DD2} and GND ₂ .

TYPICAL PERFORMANCE CHARACTERISTICS

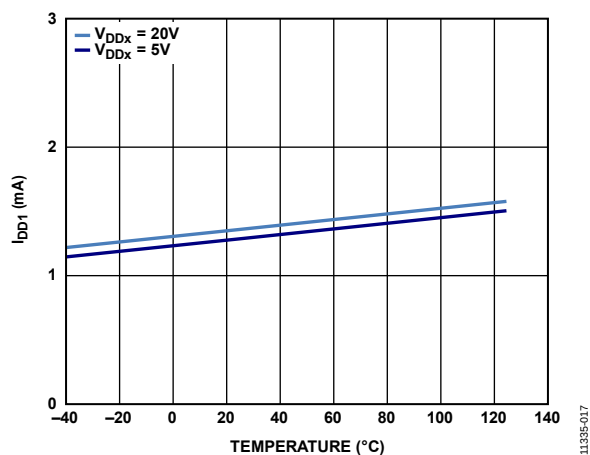
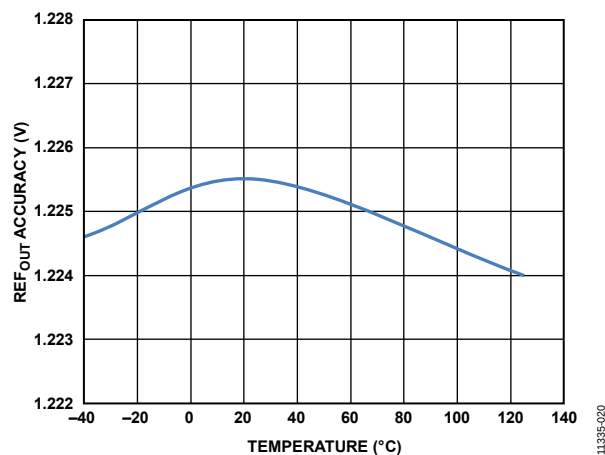
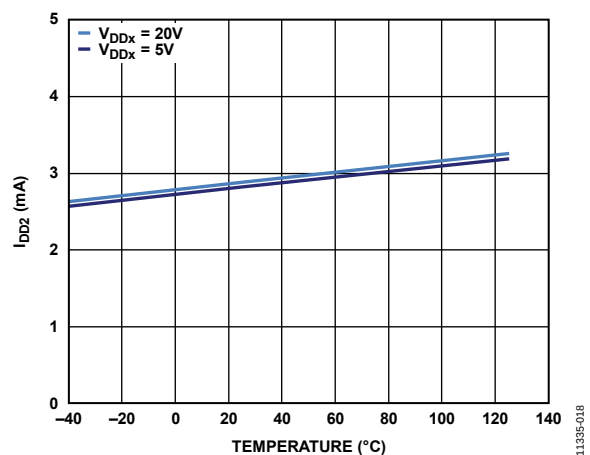
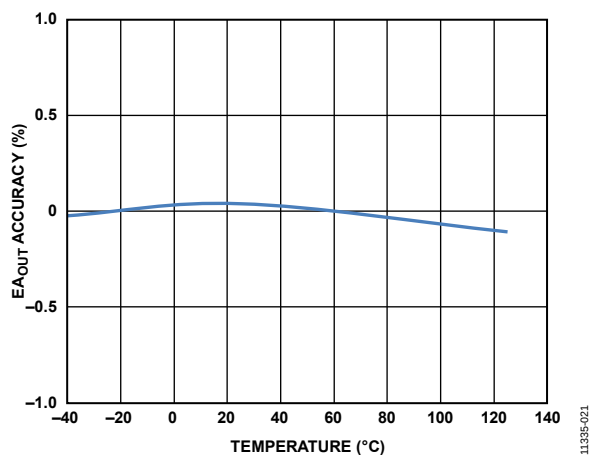
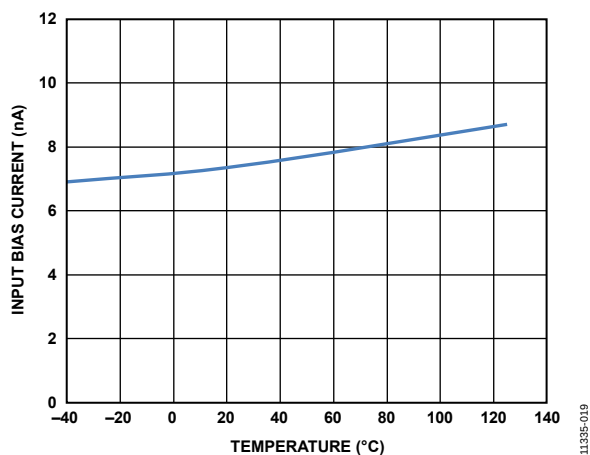
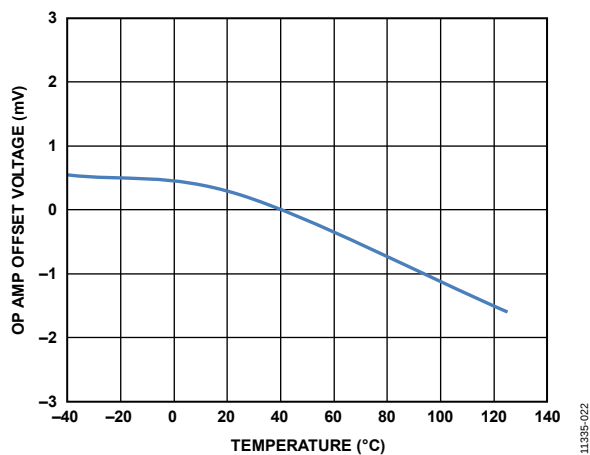
Figure 4. Typical I_{DD1} Supply Current vs. TemperatureFigure 7. REF_{OUT} Accuracy vs. TemperatureFigure 5. Typical I_{DD2} Supply Current vs. TemperatureFigure 8. EA_{OUT} Accuracy vs. TemperatureFigure 6. $+IN$, $-IN$ Input Bias Current vs. Temperature

Figure 9. Op Amp Offset Voltage vs. Temperature

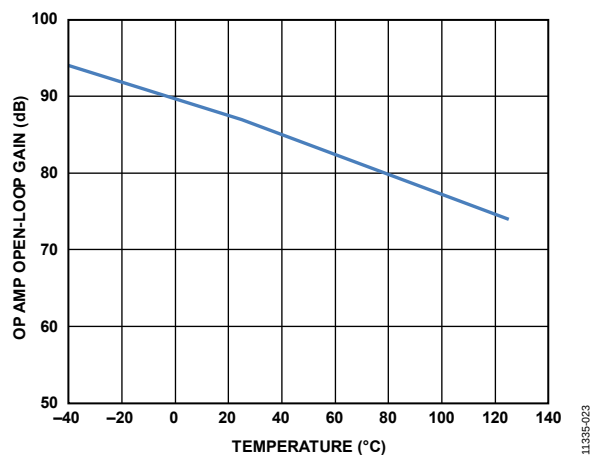


Figure 10. Op Amp Open-Loop Gain vs. Temperature

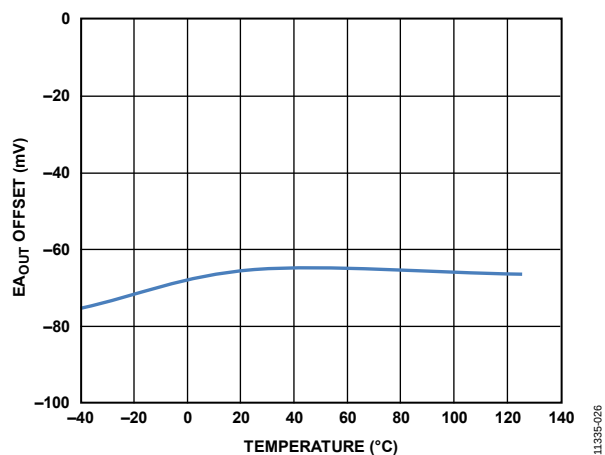


Figure 13. EAOUT Offset vs. Temperature

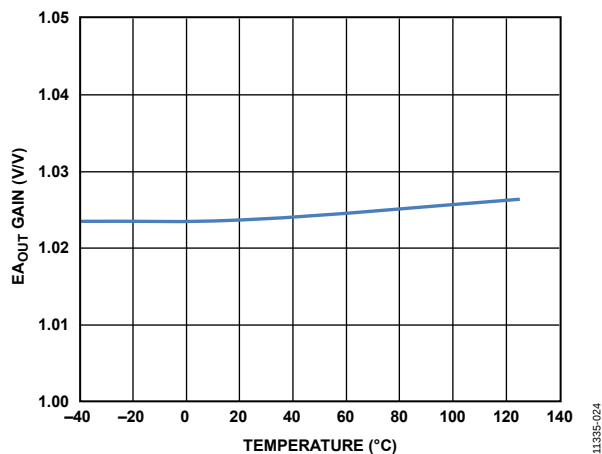


Figure 11. EAOUT Gain vs. Temperature

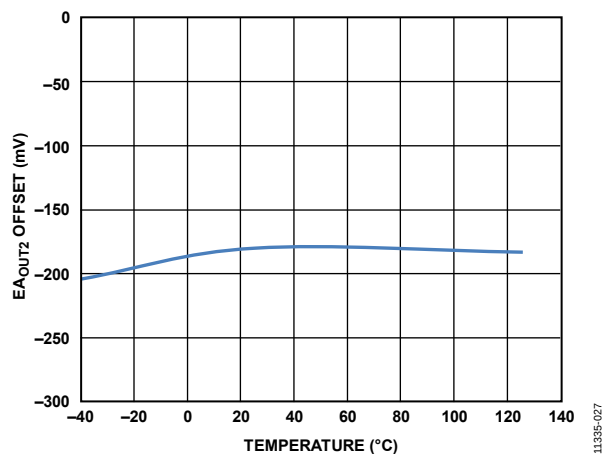


Figure 14. EAOUT2 Offset vs. Temperature

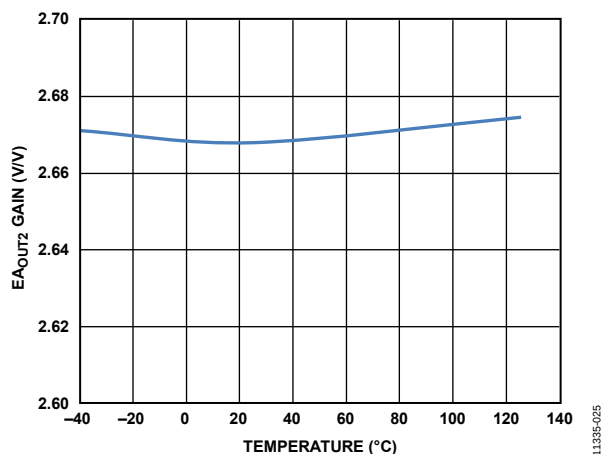


Figure 12. EAOUT2 Gain vs. Temperature

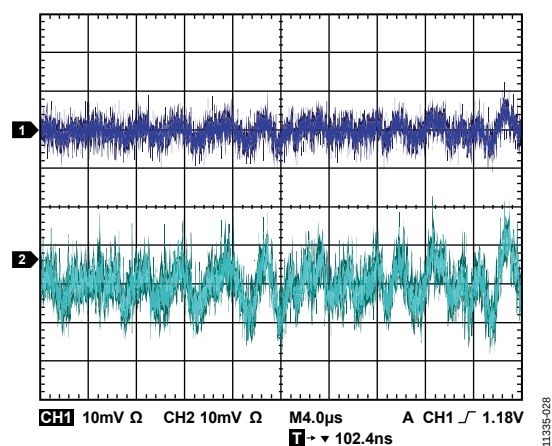


Figure 15. Output Noise with Test Circuit 1, Channel 1 = EAOUT (10 mV/div), Channel 2 = EAOUT2 (10 mV/div)

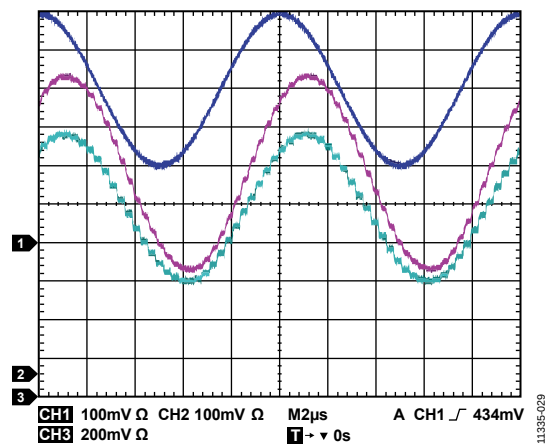


Figure 16. Output 100 kHz Signal with Test Circuit 3, Channel 1 = +IN, Channel 2 = EA_{OUT1}, Channel 3 = EA_{OUT2}

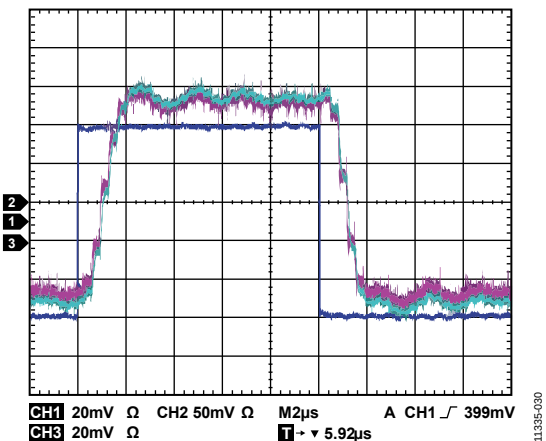


Figure 17. Output Square Wave Response with Test Circuit 3, Channel 1 = +IN, Channel 2 = EA_{OUT1}, Channel 3 = EA_{OUT2}

TEST CIRCUITS

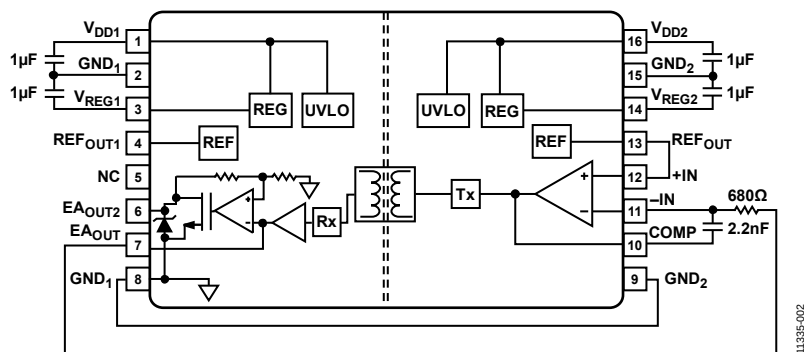


Figure 18. Test Circuit 1

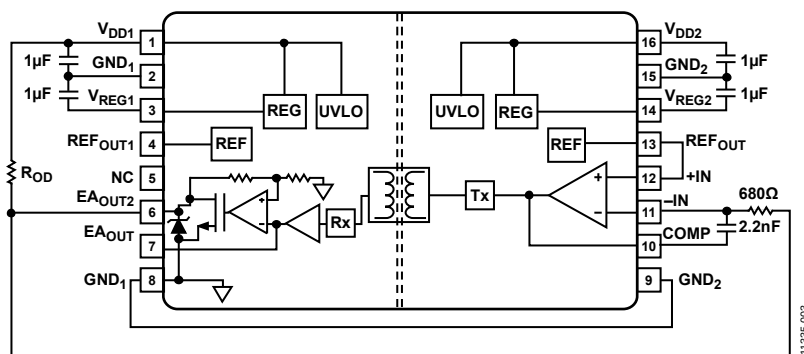


Figure 19. Test Circuit 2

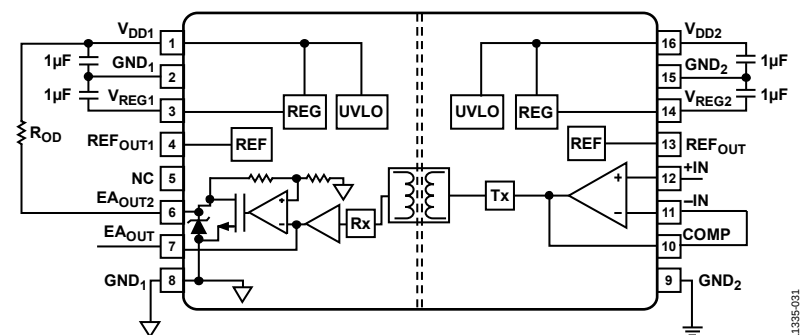


Figure 20. Test Circuit 3

APPLICATIONS INFORMATION

THEORY OF OPERATION

In the test circuits of the ADuM3190 (see Figure 18 through Figure 20), external supply voltages from 3 V to 20 V are provided to the V_{DD1} and V_{DD2} pins, and internal regulators provide 3.0 V to operate the internal circuits of each side of the ADuM3190. An internal precision 1.225 V reference provides the reference for the $\pm 1\%$ accuracy of the isolated error amplifier. UVLO circuits monitor the V_{DDx} supplies to turn on the internal circuits when the 2.8 V rising threshold is met and to turn off the error amplifier outputs to a high impedance state when V_{DDx} falls below 2.6 V.

The op amp on the right side of the device has a noninverting +IN pin and an inverting -IN pin available for connecting a feedback voltage in an isolated dc-to-dc converter output, usually through a voltage divider. The COMP pin is the op amp output, which can be used to attach resistor and capacitor components in a compensation network. The COMP pin internally drives the Tx transmitter block, which converts the op amp output voltage into an encoded output that is used to drive the digital isolator transformer.

On the left side of the ADuM3190, the transformer output PWM signal is decoded by the Rx block, which converts the signal into a voltage that drives an amplifier block; the amplifier block produces the error amplifier output available at the EA_{OUT} pin. The EA_{OUT} pin can deliver ± 3 mA and has a voltage level between 0.4 V and 2.4 V, which is typically used to drive the input of a PWM controller in a dc-to-dc circuit.

For applications that need more output voltage to drive their controllers, Figure 19 illustrates the use of the EA_{OUT2} pin output which delivers up to ± 1 mA with an output voltage of 0.6 V to 4.8 V for an output that has a pull-up resistor to a 5 V supply. If the EA_{OUT2} pull-up resistor connects to a 10 V to 20 V supply, the output is specified to a minimum of 5.0 V to allow use with a PWM controller requiring a minimum input operation of 5 V.

ACCURACY CIRCUIT OPERATION

See Figure 18 and Figure 19 for stability of the accuracy circuits. The op amp on the right side of the ADuM3190, from the -IN pin to the COMP pin, has a unity-gain bandwidth (UGBW) of 10 MHz. Figure 21, Bode Plot 1, shows a dashed line for the op amp alone and its 10 MHz pole.

Figure 21 also shows the linear isolator alone (the blocks from the op amp output to the ADuM3190 output, labeled as the linear isolator), which introduces a pole at approximately 400 kHz. This total Bode plot of the op amp and linear isolator shows that the phase shift is approximately -180° from the -IN pin to the EA_{OUT} pin before the crossover frequency. Because a -180° phase shift can make the system unstable, adding an integrator configuration, as shown in the test circuits in Figure 18 and Figure 19, consisting of a 2.2 nF capacitor and a 680 Ω resistor, helps to make the system stable. In Figure 22, Bode Plot 2 with an integrator configuration added, the system crosses over 0 dB at

approximately 100 kHz, but the circuit is more stable with a phase shift of approximately -120° , which yields a stable 60° phase margin.

This circuit is used for accuracy tests only, not for real-world applications, because it has a 680 Ω resistor across the isolation barrier to close the loop for the error amplifier; this resistor causes leakage current to flow across the isolation barrier. For this test circuit only, GND₁ must be connected to GND₂ to create a return for the leakage current created by the 680 Ω resistor connection.

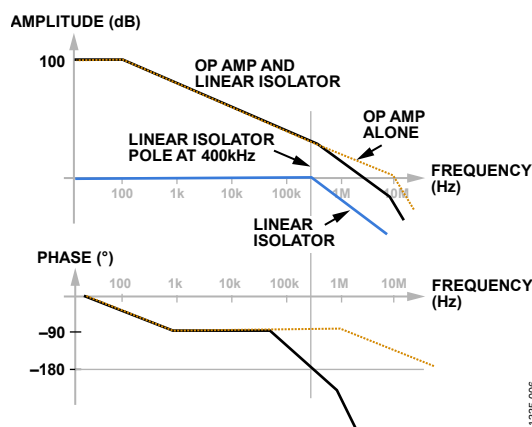


Figure 21. Bode Plot 1

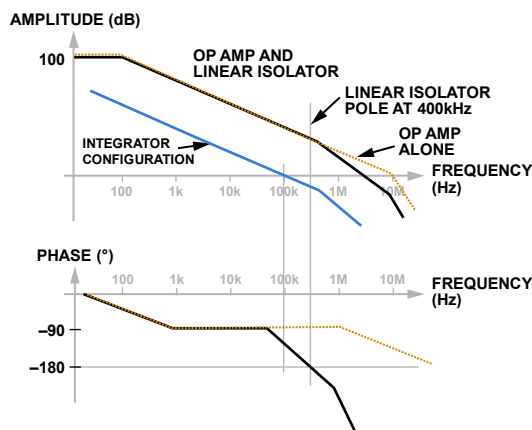


Figure 22. Bode Plot 2

APPLICATION BLOCK DIAGRAM

Figure 23 shows a typical application, an isolated error amplifier in primary side control, for the ADuM3190. The op amp of the ADuM3190 is used as the error amplifier for the feedback of the output voltage, V_{OUT} , using a resistor divider to the -IN pin of the op amp. This configuration inverts the output signal at the COMP pin when compared to the +IN pin, which is connected to the internal 1.225 V reference. For example, when the output voltage, V_{OUT} , falls due to a load step, the divider voltage at the -IN pin falls below the +IN reference voltage, causing the COMP pin output signal to go high. The COMP output of the op amp is encoded and then decoded back by the digital isolator transformer block to a signal that drives the output of the ADuM3190

high. The output of the ADuM3190 drives the COMP pin of the PWM controller, which is designed to reset the PWM latch output to low only when its COMP pin is low. A high at the COMP pin has the effect of making the latching PWM comparator produce a PWM duty cycle output. This PWM duty cycle output drives the power stage to increase the V_{OUT} voltage until it returns to regulation.

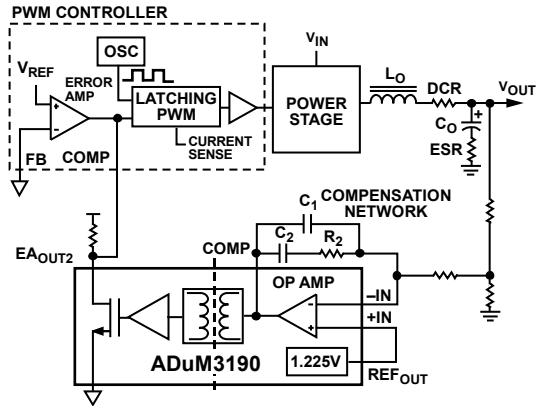


Figure 23. Application Block Diagram 1

The power stage output is filtered by output capacitance and, in some applications, by an inductor. Various elements contribute to the gain and phase of the control loop and the resulting stability. The Output Filter L and Output Filter C components create a double pole; the op amp has a pole at 10 MHz (see Figure 21), and the linear isolator has a pole at 400 kHz (see Figure 21 and Figure 22).

The output capacitor and its ESR can add a zero at a frequency that is dependent on the component type and values. With the ADuM3190 providing the error amplifier, a compensation network is provided from the -IN pin to the COMP pin to compensate the control loop for stability. The compensation network values depend on both the application and the components that are selected; information about the component network values is provided in the data sheet of the selected PWM controller.

The ADuM3190 has two different error amplifier outputs, EAOUT and EAOUT2. The EAOUT output, which can drive ± 3 mA, has a guaranteed maximum high output voltage of at least 2.4 V, which may not be enough to drive the COMP pin of some PWM controllers. The EAOUT2 pin can drive ± 1 mA and has an output range that guarantees 5.0 V for a V_{DD1} voltage range of 10 V to 20 V, which works well with the COMP pin of many PWM controllers.

In an application where the 5 V minimum output of the EAOUT2 pin is not enough to drive the COMP pin of some PWM controllers, for example, controllers that operate with 6 V or more of COMP pin voltage, use EAOUT2 to drive the FB pin of the error amplifier of the PWM controller (see Figure 24). The typical V_{REF} voltage level of PWM controllers is approximately 1.25 V or 2.5 V, setting the reference level for the FB pin. In Figure 24, the

EAOUT2 output is used for a PWM controller with a 2.5 V reference.

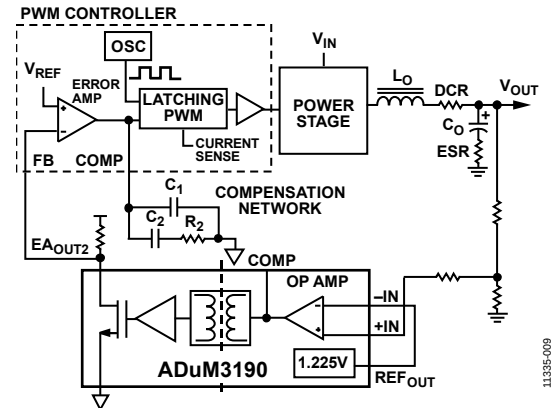


Figure 24. Application Block Diagram 2

As shown in Figure 24, the ADuM3190 op amp has the feedback voltage from the V_{OUT} output divider is connected to the +IN pin, and the +1.225 V reference voltage connected to the -IN pin. This configuration produces a low going ADuM3190 COMP pin when the V_{OUT} voltage drops from a load step. The EAOUT2 pin follows the COMP pin, going low and connecting to the PWM controller FB pin. The error amplifier of the PWM controller has a reference (V_{REF}) at the noninverting input, which, when the FB pin is low, makes the COMP pin of the error amplifier output go high. A high at the COMP pin causes the latching PWM comparator to produce a PWM duty cycle output. This PWM duty cycle output drives the power stage to increase the V_{OUT} voltage until it returns to regulation.

The application block diagrams (see Figure 23 and Figure 24) show two different ways to use the ADuM3190 to provide isolated feedback in the control loop of an isolated dc-to-dc converter. In both figures, the loop is closed at approximately the 1.225 V reference voltage, providing $\pm 1\%$ accuracy over temperature. The ADuM3190 op amp has a high gain bandwidth of 10 MHz to allow the dc-to-dc converter to operate at high switching speeds, enabling smaller values for the Output Filter L and Output Filter C components.

The 400 kHz bandwidth of the ADuM3190 error amplifier output offers faster loop response for better transient response than the typical shunt regulator and optocoupler solutions, which typically have bandwidths of only 25 kHz to 50 kHz maximum.

SETTING THE OUTPUT VOLTAGE

The output voltage in the application circuit can be set with two resistors in a voltage divider, as shown in Figure 25.

The output voltage is determined by the following equation, where $V_{REF} = 1.225$ V:

$$V_{OUT} = V_{REF} \times (R_1 + R_2)/R_2 \quad (1)$$

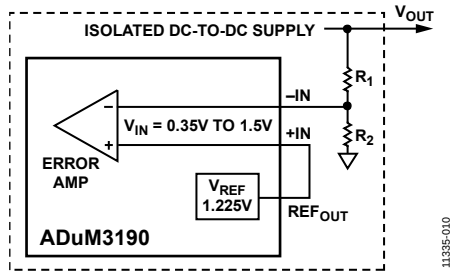


Figure 25. Setting the Output Voltage

DOSA MODULE APPLICATION

Figure 26 is a block diagram of a Distributed-power Open Standards Alliance (DOSA) circuit using the ADuM3190. The block diagram shows how to use the ADuM3190 1.225 V reference and the error amp in a DOSA standard power supply module circuit to produce output voltage settings using a combination of resistors.

The ADuM3190 1.225 V reference is specified for $\pm 1\%$ over the -40°C to $+125^\circ\text{C}$ temperature range. See Table 10 to select the resistor values to set the output voltage of the module. Two different ranges of V_{OUT} can be implemented, $V_{\text{OUT}} > 1.5\text{ V}$ or $V_{\text{OUT}} < 1.5\text{ V}$, depending on the required module.

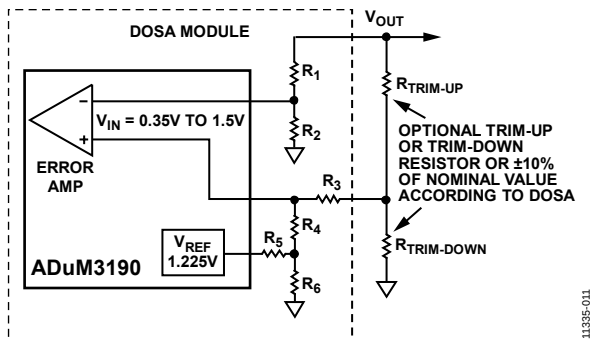


Figure 26. DOSA Module

Table 10. Resistor Values for DOSA Module

Module Nominal Output	R3	R4	R5	R6
$V_{\text{OUT}} > 1.5\text{ V}$	1 k Ω	1 k Ω	0 Ω	Open
$V_{\text{OUT}} < 1.5\text{ V}$	1 k Ω	0 Ω	2.05 k Ω	1.96 k Ω
$V_{\text{OUT}} > 1.5\text{ V}$	5.11 k Ω	5.11 k Ω	0 Ω	Open
$V_{\text{OUT}} < 1.5\text{ V}$	5.11 k Ω	0 Ω	10.5 k Ω	10.0 k Ω

DC CORRECTNESS AND MAGNETIC FIELD IMMUNITY

Positive and negative logic transitions at the isolator input cause narrow ($\sim 1\text{ ns}$) pulses to be sent to the decoder via the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. In the absence of logic transitions of more than $1\text{ }\mu\text{s}$ at the input, a periodic set of refresh pulses indicative of the correct input state are sent to ensure dc correctness at the output.

If the decoder receives no internal pulses for more than approximately $3\text{ }\mu\text{s}$, the input side is assumed to be unpowered or nonfunctional, in which case, the isolator output is forced to

a default high impedance state by the watchdog timer circuit. In addition, the outputs are in a default high impedance state while the power is increasing before the UVLO threshold is crossed.

The ADuM3190 is immune to external magnetic fields. The limitation on the ADuM3190 magnetic field immunity is set by the condition whereby induced voltage in the transformer receiving coil is sufficiently large to either falsely set or reset the decoder. The following analysis defines the conditions under which this can occur. The 3 V operating condition of the ADuM3190 is examined because it represents the most susceptible mode of operation. The pulses at the transformer output have an amplitude that is greater than 1.0 V . The decoder has a sensing threshold at approximately 0.5 V , therefore establishing a 0.5 V margin within which induced voltages are tolerated. The voltage induced across the receiving coil is given by

$$V = (-d\beta/dt) \sum \pi r_n^2, n = 1, 2, \dots, N$$

where:

β is the magnetic flux density (gauss).

r_n is the radius of the n th turn in the receiving coil (cm).

N is the number of turns in the receiving coil.

Given the geometry of the receiving coil in the ADuM3190 and an imposed requirement that the induced voltage be, at most, 50% of the 0.5 V margin at the decoder, a maximum allowable magnetic field is calculated, as shown in Figure 27.

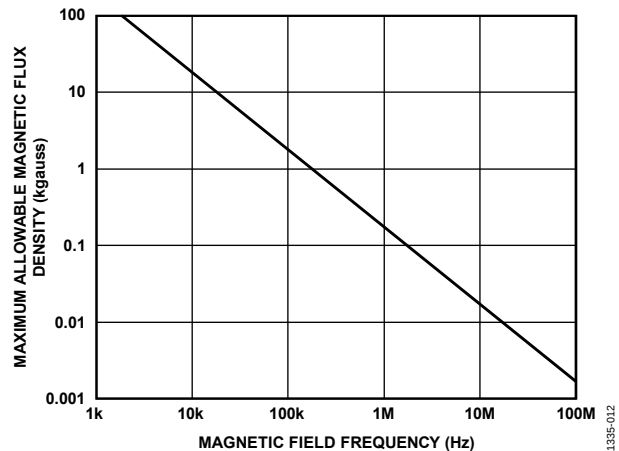


Figure 27. Maximum Allowable External Magnetic Flux Density

For example, at a magnetic field frequency of 1 MHz , the maximum allowable magnetic field of 0.02 kgauss induces a voltage of 0.25 V at the receiving coil. This is approximately 50% of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event were to occur during a transmitted pulse (and had the worst-case polarity), the received pulse is reduced from $>1.0\text{ V}$ to 0.75 V , still well above the 0.5 V sensing threshold of the decoder.

The preceding magnetic flux density values correspond to specific current magnitudes at given distances away from the ADuM3190 transformers. Figure 28 shows these allowable current magnitudes as a function of frequency for selected distances. As shown Figure 28, the ADuM3190 is immune and can

be affected only by extremely large currents operating at a high frequency very close to the component. For the 1 MHz example, a 0.7 kA current must be placed 5 mm away from the ADuM3190 to affect the operation of the device.

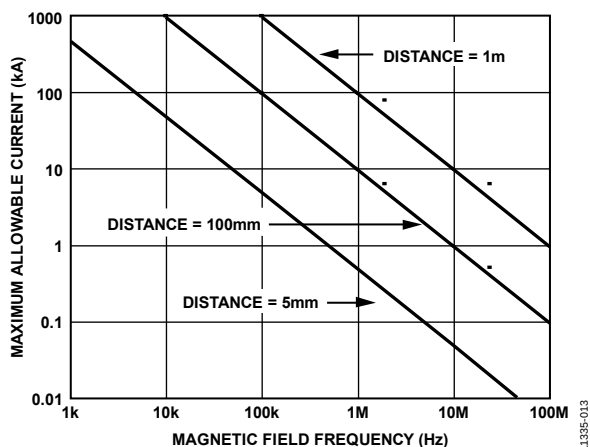


Figure 28. Maximum Allowable Current for Various Current-to-ADuM3190 Spacings

INSULATION LIFETIME

All insulation structures eventually break down when subjected to voltage stress over a sufficiently long period. The rate of insulation degradation is dependent on the characteristics of the voltage waveform applied across the insulation. In addition to the testing performed by the regulatory agencies, Analog Devices carries out an extensive set of evaluations to determine the lifetime of the insulation structure within the ADuM3190.

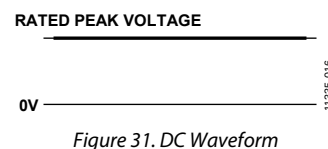
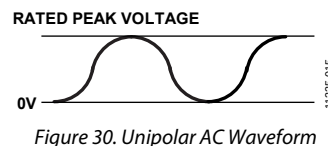
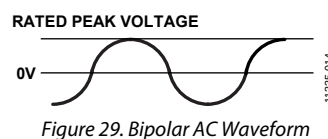
Analog Devices performs accelerated life testing using voltage levels higher than the rated continuous working voltage. Acceleration factors for several operating conditions are determined. These factors allow calculation of the time to failure at the actual working voltage.

The values shown in Table 8 summarize the peak voltage for 50 years of service life for a bipolar ac operating condition. In many cases, the approved working voltage is higher than the 50-year service life voltage. Operation at these high working voltages can lead to shortened insulation life in some cases.

The ADuM3190 insulation lifetime depends on the voltage waveform type imposed across the isolation barrier. The iCoupler insulation structure degrades at different rates depending on whether the waveform is bipolar ac, unipolar ac, or dc. Figure 29, Figure 30, and Figure 31 illustrate these different isolation voltage waveforms.

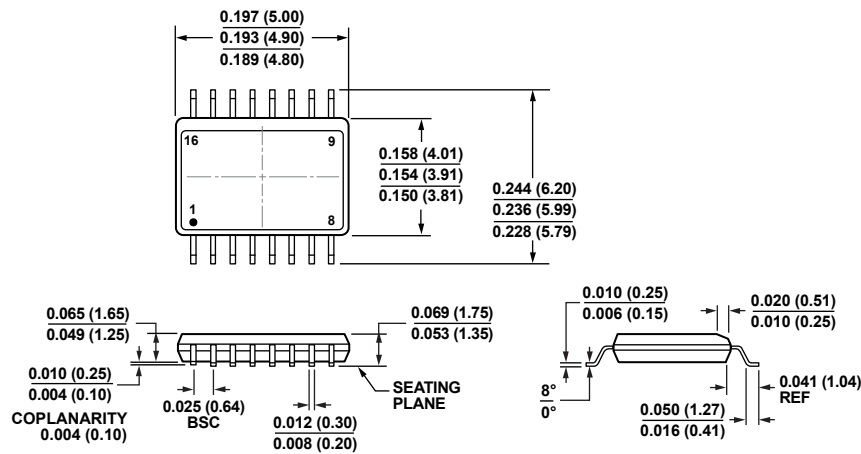
A bipolar ac voltage environment is the worst case for the iCoupler products yet meets the 50-year operating lifetime recommended by Analog Devices for maximum working voltage. In the case of unipolar ac or dc voltage, the stress on the insulation is significantly lower. This allows operation at higher working voltages while still achieving a 50-year service life. Treat any cross-insulation voltage waveform that does not conform to Figure 30 or Figure 31 as a bipolar ac waveform, and limit its peak voltage to the 50-year lifetime voltage value listed in Table 8.

Note that the voltage presented in Figure 30 is shown as sinusoidal for illustration purposes only. It is meant to represent any voltage waveform varying between 0 V and some limiting value. The limiting value can be positive or negative, but the voltage cannot cross 0 V.



PACKAGING AND ORDERING INFORMATION

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-137-AB
CONTROLLING DIMENSIONS ARE IN INCHES; MILLIMETER DIMENSIONS
(IN PARENTHESES) ARE ROUNDED-OFF INCH EQUIVALENTS FOR
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

Figure 32. 16-Lead Shrink Small Outline Package [QSOP]
(RQ-16)

Dimensions shown in inches and (millimeters)

01-28-2008-A

ORDERING GUIDE

Model ¹	Temperature Range	Bandwidth (Typical)	Package Description	Package Option
ADuM3190ARQZ	-40°C to +85°C	200 kHz	16-Lead QSOP	RQ-16
ADuM3190ARQZ-RL7	-40°C to +85°C	200 kHz	16-Lead QSOP	RQ-16
ADuM3190BRQZ	-40°C to +85°C	400 kHz	16-Lead QSOP	RQ-16
ADuM3190BRQZ-RL7	-40°C to +85°C	400 kHz	16-Lead QSOP	RQ-16
ADuM3190SRQZ	-40°C to +125°C	200 kHz	16-Lead QSOP	RQ-16
ADuM3190SRQZ-RL7	-40°C to +125°C	200 kHz	16-Lead QSOP	RQ-16
ADuM3190TRQZ	-40°C to +125°C	400 kHz	16-Lead QSOP	RQ-16
ADuM3190TRQZ-RL7	-40°C to +125°C	400 kHz	16-Lead QSOP	RQ-16
EVAL-ADuM3190EBZ			Evaluation Board	

¹ Z = RoHS Compliant Part.