

## System Reset IC

### ■FEATURES

- Full compatible with NJM2103
- Detection Voltage  $V_{SA}=4.2V\pm1.0\%$
- Adjustable Detection Voltage  $V_{SB}=1.22V\pm1.0\%$
- $V_{SA}$  and  $V_{SB}$  have Hysteresis Characteristics at Reset release
- Possible to detect Over Voltage  $V_{SC}=1.235V\pm1.0\%$
- Operating Temperature  $T_a=-40$  to  $125^\circ\text{C}$
- Low Quiescent Current  $280\mu\text{A typ.}$
- Reference Voltage can be taken out  $0.8V \text{ typ.}$
- Low reset operation voltage DMP8, DIP8, TVSP8

### ■GENERAL DESCRIPTION

The NJU2103A is a power supply voltage monitoring IC that instantaneously detects abnormality such as power supply voltage cutoff or drop and generates reset signal.

It can monitor 2 systems of 5 V power supply and arbitrarily set voltage.

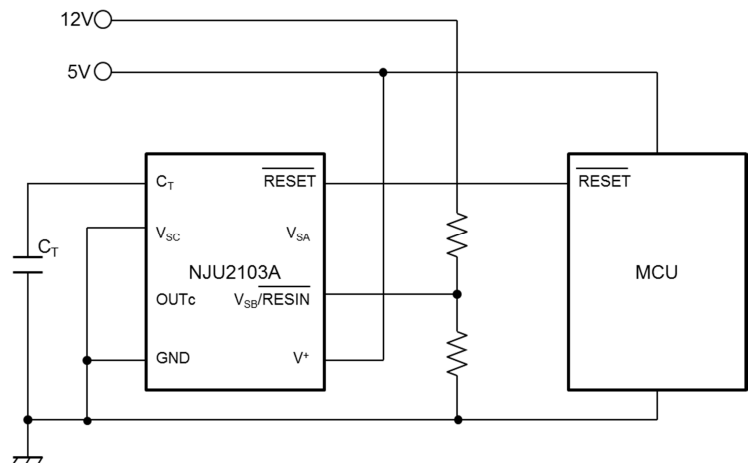
It is possible to direct replacement from NJM2103 and the small TVSP8 package is added.

Furthermore, it improves usability by extending operating temperature, standardizing AC characteristics, and making each parameter highly accurate.

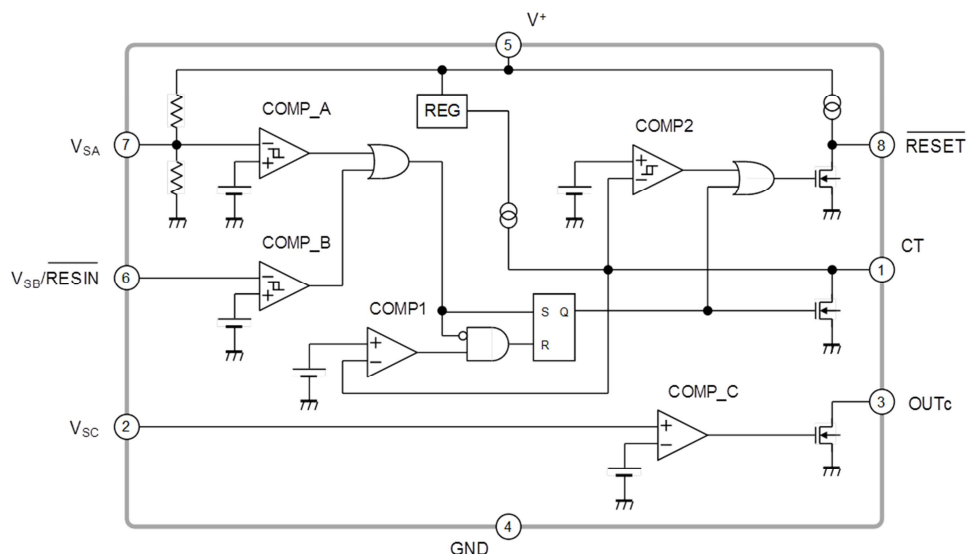
### ■APPLICATION

- Industrial equipment
- Housing and facility equipment
- OA equipment
- Amusement equipment

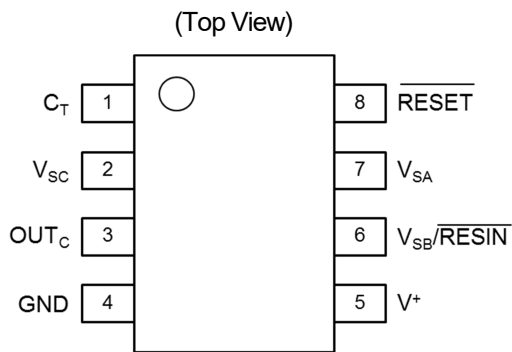
### ■TYPICAL APPLICATION



### ■BLOCK DIAGRAM



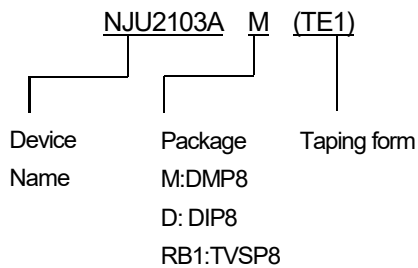
## ■PIN CONFIGURATION



DMP8/DIP8/TVSP8

| PIN No. | PIN NAME           | FUNCTION                                                    |
|---------|--------------------|-------------------------------------------------------------|
| 1       | $C_T$              | Connects Capacitor pin for setting RESET Output Pulse Width |
| 2       | $V_{SC}$           | Comparator C input pin                                      |
| 3       | $OUT_C$            | Comparator C output pin                                     |
| 4       | GND                | GND pin                                                     |
| 5       | $V^+$              | Power Supply pin                                            |
| 6       | $V_{SB}/RESIN$     | Comparator B input pin                                      |
| 7       | $V_{SA}$           | Comparator A input pin                                      |
| 8       | $\overline{RESET}$ | RESET output pin (Active Low)                               |

## ■PRODUCT NAME INFORMATION



## ■ORDERING INFORMATION

| PRODUCT NAME     | PACKAGE OUTLINE | RoHS | Halogen-Free | TERMINAL FINISH | MARKING | WEIGHT (mg) | MOQ (pcs) |
|------------------|-----------------|------|--------------|-----------------|---------|-------------|-----------|
| NJU2103AM(TE1)   | DMP8            | ○    | ○            | Sn-2Bi          | 2103A   | 95          | 2000      |
| NJU2103AD        | DIP8            | ○    | -            | Sn-2Bi          | 2103AD  | 510         | 2000      |
| NJU2103ARB1(TE1) | TVSP8           | ○    | ○            | Sn-2Bi          | 2103A   | 18          | 2000      |

## ■ABSOLUTE MAXIMUM RATINGS

| PARAMETER                       | SYMBOL      | RATINGS                                  | UNIT |
|---------------------------------|-------------|------------------------------------------|------|
| Supply Voltage                  | $V^+$       | -0.3 to 20                               | V    |
| Input Voltage                   | $V_{SA}$    | -0.3 to $V^+ + 0.3$ (<20)                | V    |
|                                 | $V_{SB}$    | -0.3 to 20                               | V    |
|                                 | $V_{SC}$    | -0.3 to 20                               | V    |
| $C_T$ Pin Voltage               | $V_{CT}$    | -0.3 to $V^+ + 0.3$ (<20)                | V    |
| RESET Output Voltage            | $V_{RESET}$ | -0.3 to $V^+ + 0.3$ (<20)                | V    |
| OUT <sub>C</sub> Output Voltage | $V_{OUTC}$  | -0.3 to 20                               | V    |
| Power Dissipation(Ta=25°C)      | $P_D$       | (2-layer / 4-layer)                      | mW   |
| DMP8                            |             | 470 <sup>(1)</sup> / 600 <sup>(2)</sup>  |      |
| DIP8                            |             | 910 <sup>(1)</sup> / 1300 <sup>(2)</sup> |      |
| TVSP8                           |             | 510 <sup>(1)</sup> / 680 <sup>(2)</sup>  |      |
| Junction Temperature            | $T_J$       | -40 to +150                              | °C   |
| Operating Temperature           | $T_{opr}$   | -40 to +125                              | °C   |
| Storage Temperature             | $T_{stg}$   | -50 to +150                              | °C   |

(1): Mounted on glass epoxy board.(76.2 x 114.3 x 1.6 :based on EIA/JEDEC standard, 2 Layers)

(2): Mounted on glass epoxy board.(76.2 x 114.3 x 1.6 :based on EIA/JEDEC standard, 4 Layers) internal Cu area: 74.2 x 74.2mm

## ■RECOMMENDED OPERATING CONDITIONS

| PARAMETER                | SYMBOL      | RATINGS      | UNIT |
|--------------------------|-------------|--------------|------|
| Supply Voltage           | $V^+$       | 2.5 to 18    | V    |
| Input Voltage            | $V_{SA}$    | 0 to $V^+$   | V    |
|                          | $V_{SB}$    | 0 to 18      | V    |
|                          | $V_{SC}$    | 0 to 18      | V    |
| Output Current           | $I_{RESET}$ | 0 to 20      | mA   |
|                          | $I_{OUTC}$  | 0 to 6       | mA   |
| RESET Output Pulse Width | $t_{PO}$    | 0.15 to 1500 | ms   |
| $C_T$ Capacitor          | $C_T$       | 0.001 to 10  | μF   |

## ■ELECTRICAL CHARACTERISTICS

### (DC Characteristics)

Unless other noted,  $V^+=5V$ ,  $V_{SB}=0V$ ,  $V_{SC}=0V$ ,  $C_T=0.01\mu F$ ,  $T_a=25^\circ C$

| PARAMETER                                     | SYMBOL                 | TEST CONDITION                                                      | MIN.  | TYP.  | MAX.  | UNIT    |
|-----------------------------------------------|------------------------|---------------------------------------------------------------------|-------|-------|-------|---------|
| Operating Current 1                           | $I_{CC1}$              | $V_{SB}=5V$                                                         | -     | 280   | 390   | $\mu A$ |
| Operating Current 2                           | $I_{CC2}$              |                                                                     | -     | 300   | 410   | $\mu A$ |
| $V_{SA}$ Detecting Voltage 1                  | $V_{SAL}$              | $V^+$ sweep down, $V_{SB}=V^+$                                      | 4.158 | 4.200 | 4.242 | V       |
|                                               |                        | $V^+$ sweep down, $V_{SB}=V^+$ , $T_a=-40^\circ C$ to $125^\circ C$ | 4.050 | -     | 4.350 |         |
| $V_{SA}$ Detecting Voltage 2                  | $V_{SAH}$              | $V^+$ sweep up, $V_{SB}=V^+$                                        | 4.210 | 4.300 | 4.390 | V       |
|                                               |                        | $V^+$ sweep up, $V_{SB}=V^+$ , $T_a=-40^\circ C$ to $125^\circ C$   | 4.150 | -     | 4.450 |         |
| $V_{SA}$ Hysteresis Width                     | $V_{HRSA}$             |                                                                     | 50    | 100   | 150   | mV      |
| $V_{SB}$ Detecting Voltage                    | $V_{SBL}$              | $V_{SB}$ sweep down                                                 | 1.208 | 1.220 | 1.232 | V       |
|                                               |                        | $V_{SB}$ sweep down, $T_a=-40^\circ C$ to $125^\circ C$             | 1.190 | -     | 1.250 |         |
| $V_{SB}$ Detecting Supply Voltage Fluctuation | $\Delta V_{SBL}$       | $V^+=2.5$ to $18V$                                                  | -     | 3     | 10    | mV      |
| $V_{SB}$ Hysteresis Width                     | $V_{HRSB}$             |                                                                     | 14    | 28    | 42    | mV      |
| $V_{SB}$ Input Current 1                      | $I_{IHB}$              | $V_{SB}=5V$                                                         | -     | 0     | 250   | nA      |
| $V_{SB}$ Input Current 2                      | $I_{ILB}$              |                                                                     | -     | 0     | 250   | nA      |
| High Level RESET Output Voltage               | $V_{OHR}$              | $\overline{I_{RESET}}=-5\mu A$ , $V_{SB}=5V$                        | 4.5   | 4.9   | -     | V       |
| RESET Output Saturation Voltage 1             | $V_{OLR1}$             | $\overline{I_{RESET}}=3mA$                                          | -     | 0.05  | 0.40  | V       |
| RESET Output Saturation Voltage 2             | $V_{OLR2}$             | $\overline{I_{RESET}}=10mA$                                         | -     | 0.15  | 0.50  | V       |
| RESET Output Sink Current                     | $\overline{I_{RESET}}$ | $V_{OLR}=1V$                                                        | 20    | 60    | -     | mA      |
| $C_T$ Charge Current                          | $I_{CT}$               | $V_{SB}=5V$ , $V_{CT}=0.5V$                                         | 6     | 9     | 13    | $\mu A$ |
| $V_{SC}$ Input Current1                       | $I_{IHC}$              | $V_{SC}=5V$                                                         | -     | 0     | 500   | nA      |
| $V_{SC}$ Input Current 2                      | $I_{ILC}$              |                                                                     | -     | 0     | 500   | nA      |
| $V_{SC}$ Detecting Voltage                    | $V_{SC}$               |                                                                     | 1.223 | 1.235 | 1.247 | V       |
|                                               |                        | $T_a=-40^\circ C$ to $125^\circ C$                                  | 1.195 | -     | 1.275 |         |
| $V_{SC}$ Detecting Supply Voltage Fluctuation | $\Delta V_{SC}$        | $V^+=2.5$ to $18V$                                                  | -     | 3     | 10    | mV      |
| $OUT_C$ Output Leak Current                   | $I_{OHC}$              | $V_{OHC}=18V$                                                       | -     | 0     | 1     | $\mu A$ |
| $OUT_C$ Output Saturation Voltage             | $V_{OLC}$              | $I_{OUTC}=4mA$ , $V_{SC}=5V$                                        | -     | 0.15  | 0.40  | V       |
| $OUT_C$ Output Sink Current                   | $I_{OUTC}$             | $V_{OLC}=1V$ , $V_{SC}=5V$                                          | 6     | 20    | -     | mA      |
| RESET Minimum Operating Voltage               | $V_L^+$                | $V_{OLR}=0.4V$ , $\overline{I_{RESET}}=200\mu A$                    | -     | 0.8   | 1.2   | V       |

## ■ELECTRICAL CHARACTERISTICS

### (AC Characteristics)

Unless other noted,  $V^+=5V$ ,  $V_{SB}=5V$ ,  $V_{SC}=0V$ ,  $C_T=0.01\mu F$ ,  $T_a=25^\circ C$

| PARAMETER                  | SYMBOL    | TEST CONDITION                                                    | MIN. | TYP. | MAX. | UNIT    |
|----------------------------|-----------|-------------------------------------------------------------------|------|------|------|---------|
| $V_{SA}$ Input Pulse Width | $t_{PIA}$ |                                                                   | 5    | -    | -    | $\mu s$ |
| $V_{SB}$ Input Pulse Width | $t_{PIB}$ |                                                                   | 5    | -    | -    | $\mu s$ |
| RESET Output Pulse Width   | $t_{PO}$  | $V_{SB}=V^+$                                                      | 1.0  | 1.5  | 2.0  | ms      |
| RESET Rise Time            | $t_r$     | $V_{SB}=V^+$ , $R_L=2.2k\Omega$ , $C_L=100pF$<br>RESET=10% to 90% | -    | 1.0  | 1.5  | $\mu s$ |
| RESET Fall Time            | $t_f$     | $V_{SB}=V^+$ , $R_L=2.2k\Omega$ , $C_L=100pF$<br>RESET=90% to 10% | -    | 0.1  | 0.5  | $\mu s$ |
| Output Delay Time          | $t_{PD}$  | $V_{SB}$ sweep down                                               | -    | 2    | 10   | $\mu s$ |
|                            | $t_{PHL}$ | $V_{SC}$ sweep up, $R_L=2.2k\Omega$ , $C_L=100pF$                 | -    | 0.5  | -    | $\mu s$ |
|                            | $t_{PLH}$ | $V_{SC}$ sweep down, $R_L=2.2k\Omega$ , $C_L=100pF$               | -    | 1.0  | -    | $\mu s$ |

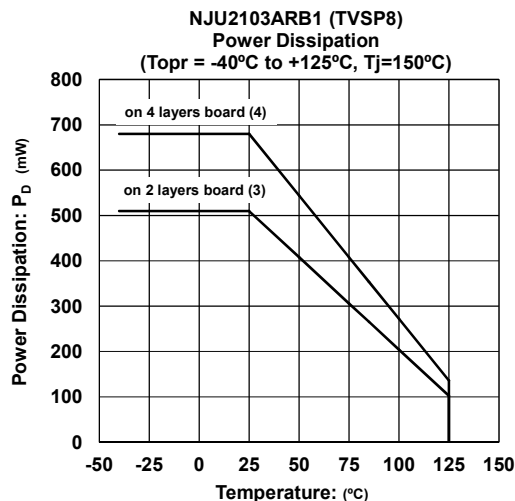
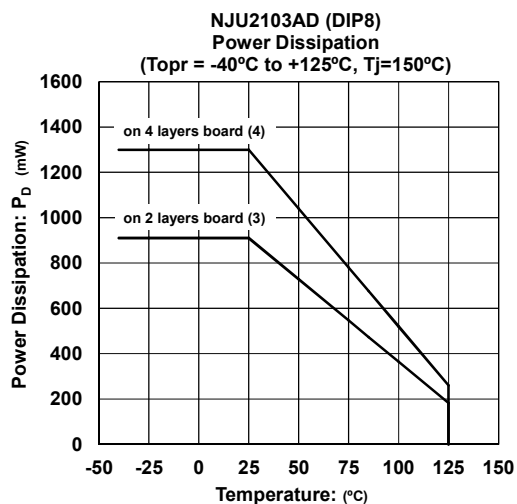
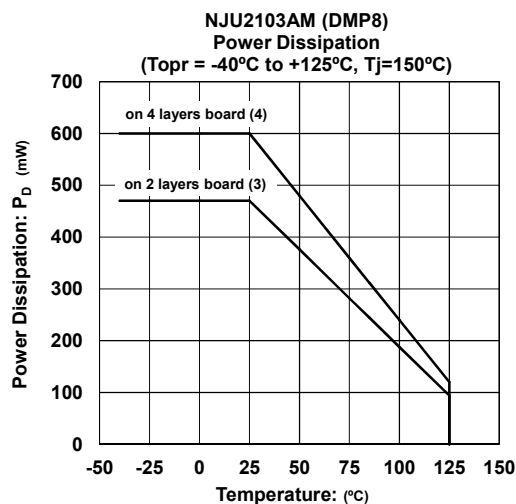
## ■ THERMAL CHARACTERISTICS

| PARAMETER                                             | SYMBOL        | VALUE |                                          | UNIT |
|-------------------------------------------------------|---------------|-------|------------------------------------------|------|
| Junction-to-ambient thermal resistance                | $\theta_{ja}$ | DMP8  | 262 <sup>(3)</sup><br>206 <sup>(4)</sup> | °C/W |
|                                                       |               | DIP8  | 138 <sup>(3)</sup><br>98 <sup>(4)</sup>  |      |
|                                                       |               | TVSP8 | 244 <sup>(3)</sup><br>185 <sup>(4)</sup> |      |
| Junction-to-Top of package characterization parameter | $\psi_{jt}$   | DMP8  | 72 <sup>(3)</sup><br>65 <sup>(4)</sup>   | °C/W |
|                                                       |               | DIP8  | 57 <sup>(3)</sup><br>54 <sup>(4)</sup>   |      |
|                                                       |               | TVSP8 | 51 <sup>(3)</sup><br>45 <sup>(4)</sup>   |      |

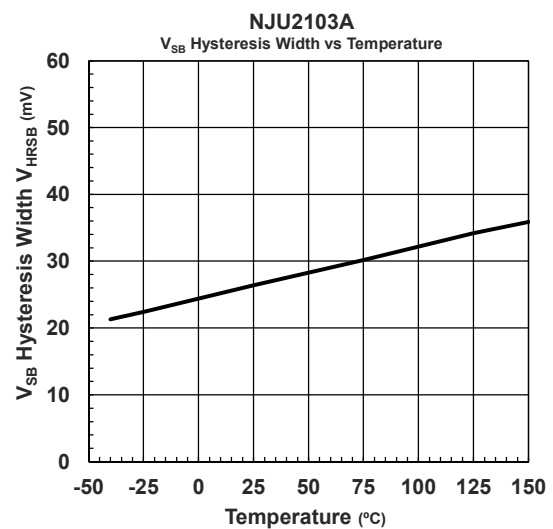
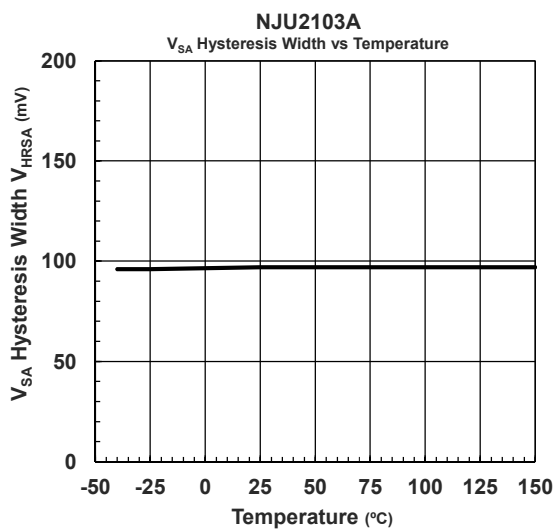
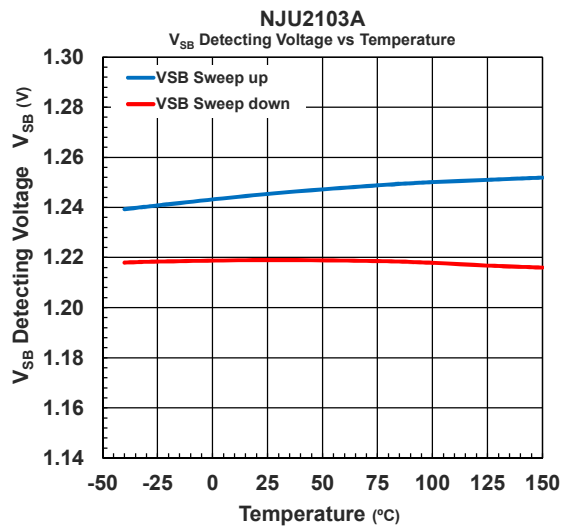
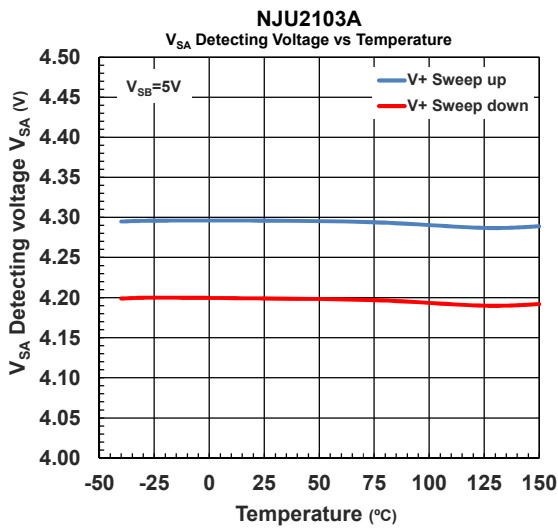
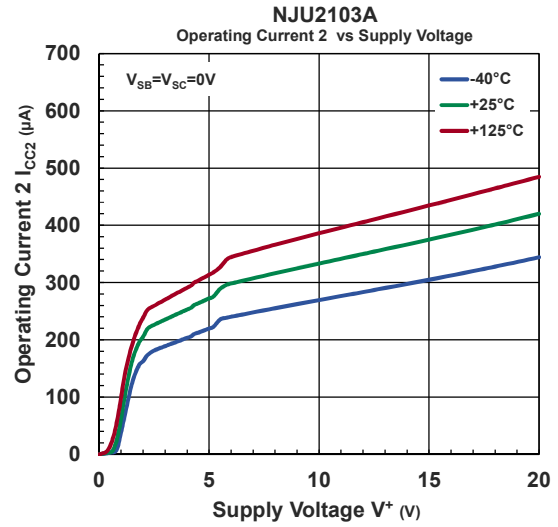
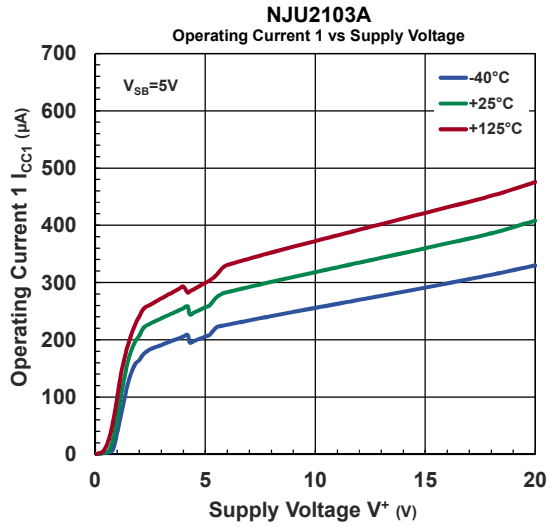
(3): Mounted on glass epoxy board.(76.2 x 114.3 x 1.6 :based on EIA/JEDEC standard, 2 Layers)

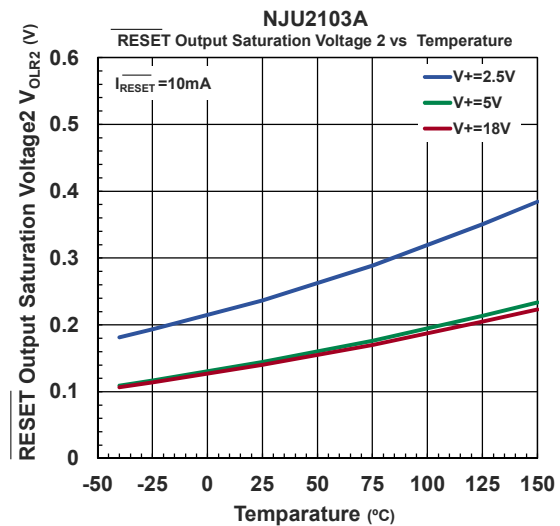
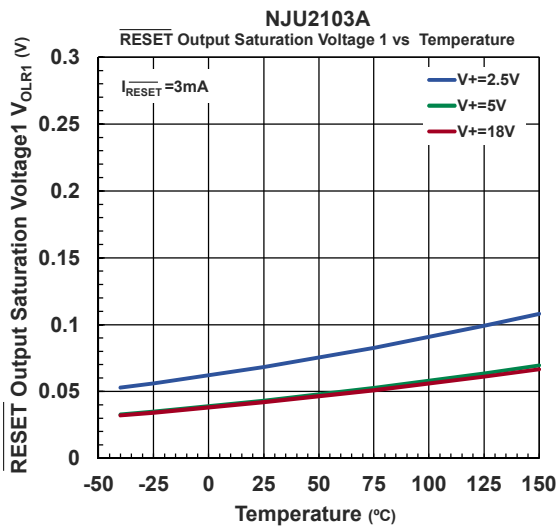
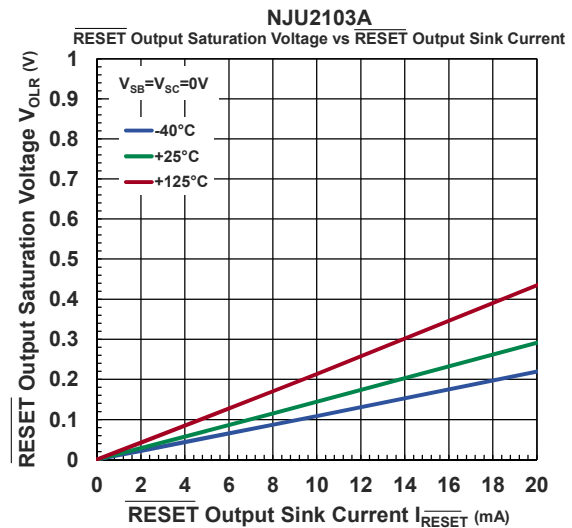
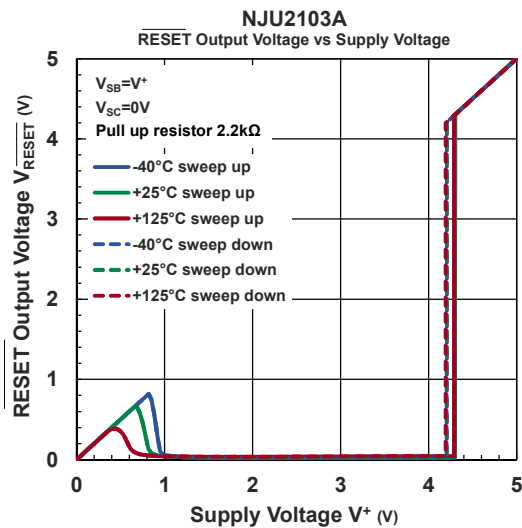
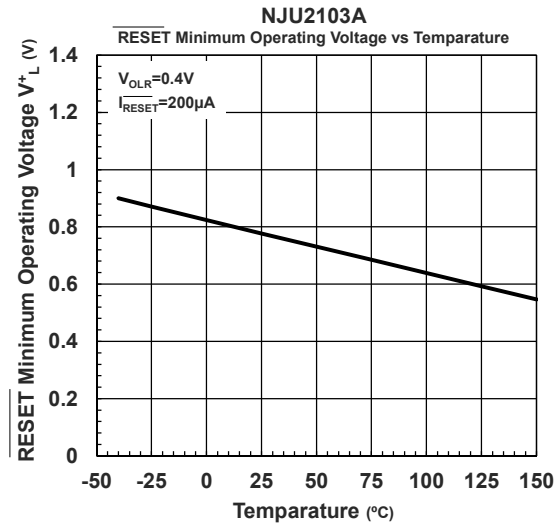
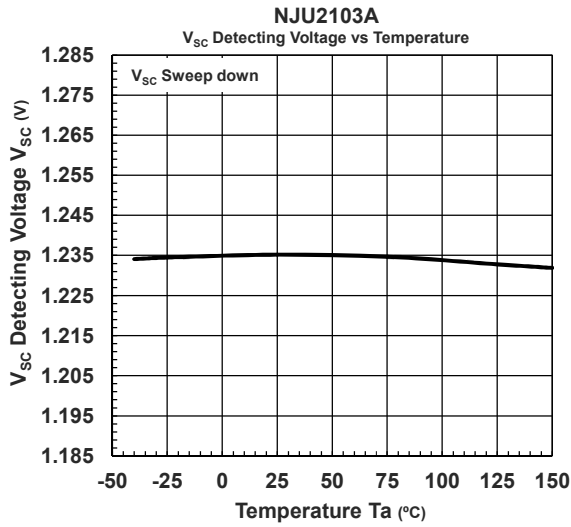
(4): Mounted on glass epoxy board.(76.2 x 114.3 x 1.6 :based on EIA/JEDEC standard, 4 Layers) internal Cu area: 74.2 x 74.2mm

## ■ POWER DISSIPATION vs. AMBIENT TEMPERATURE

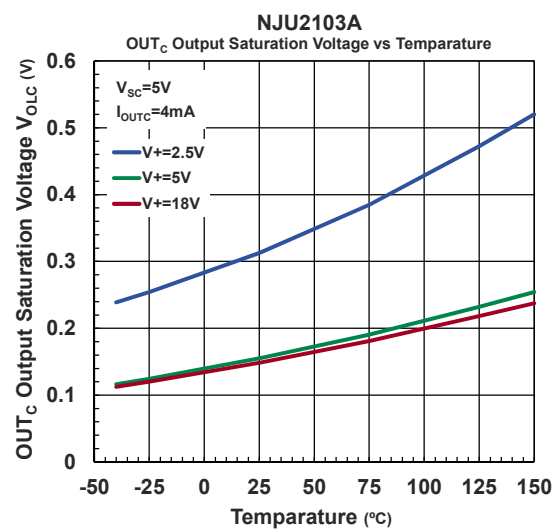
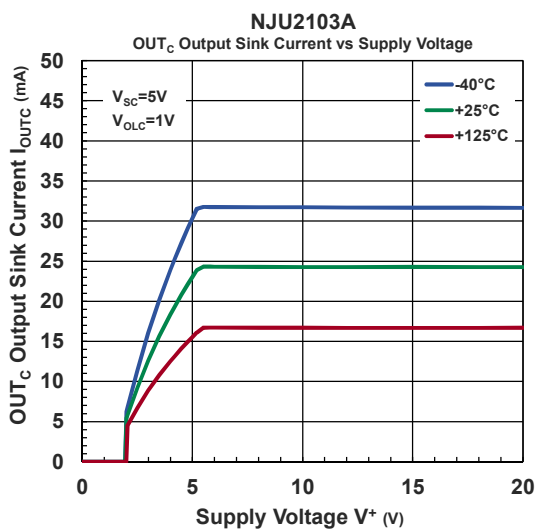
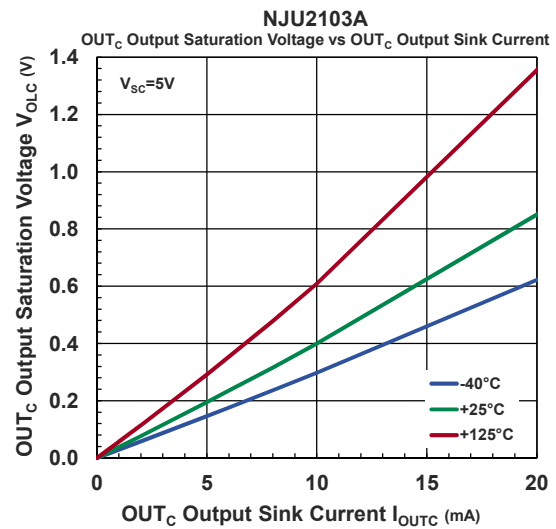
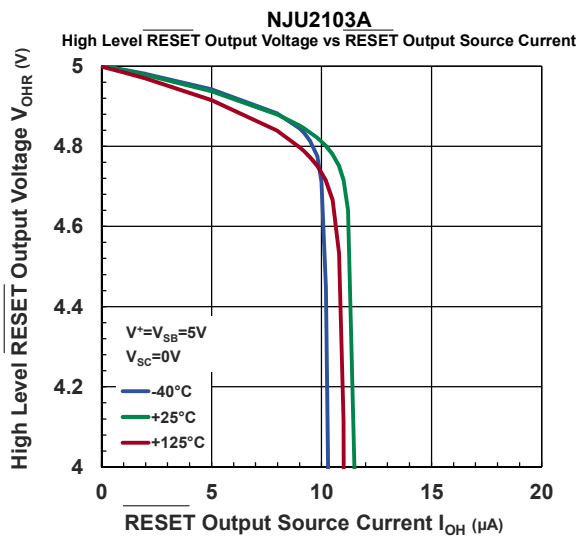
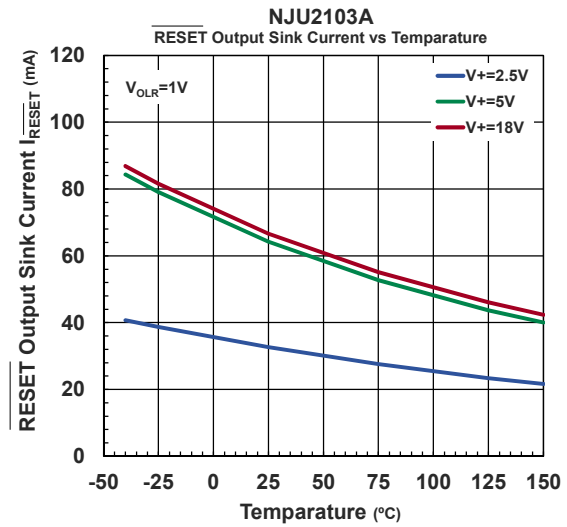
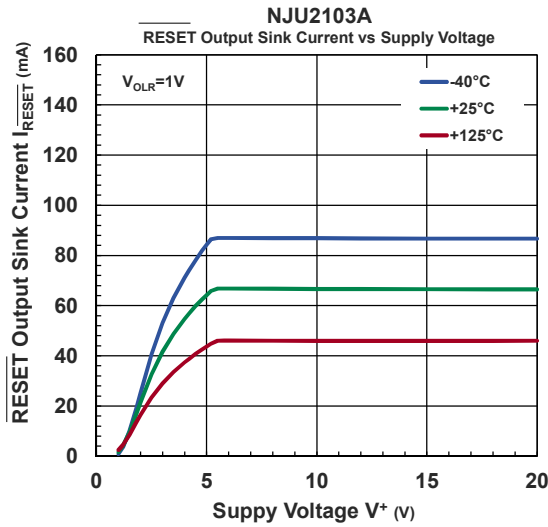


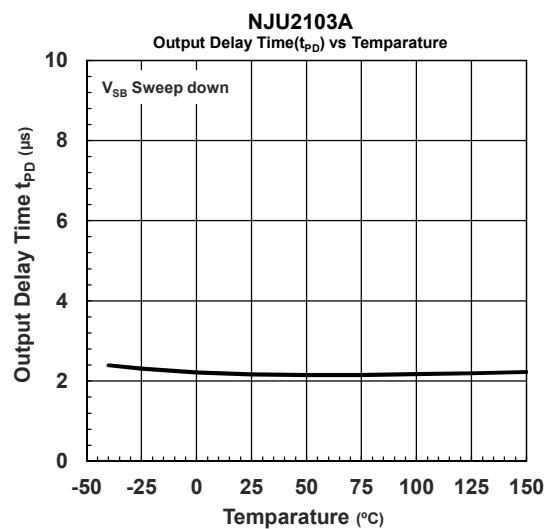
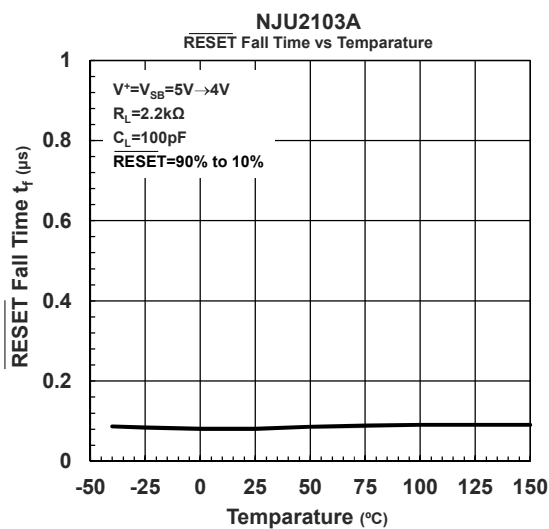
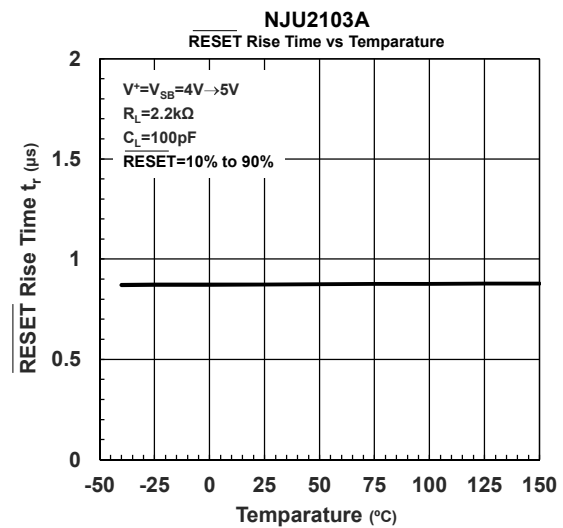
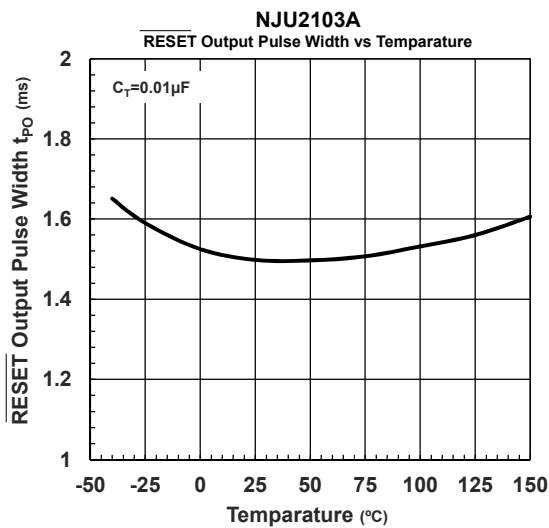
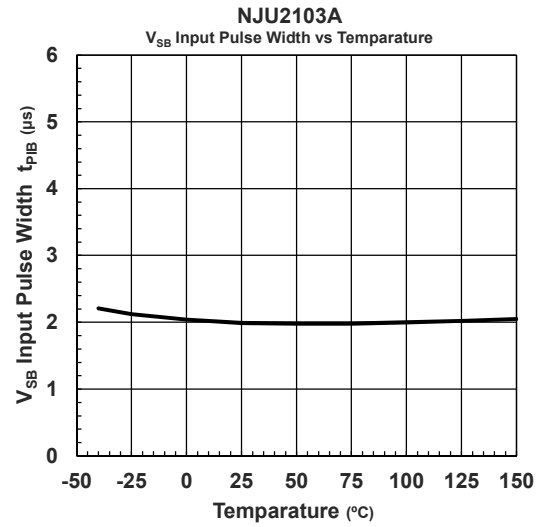
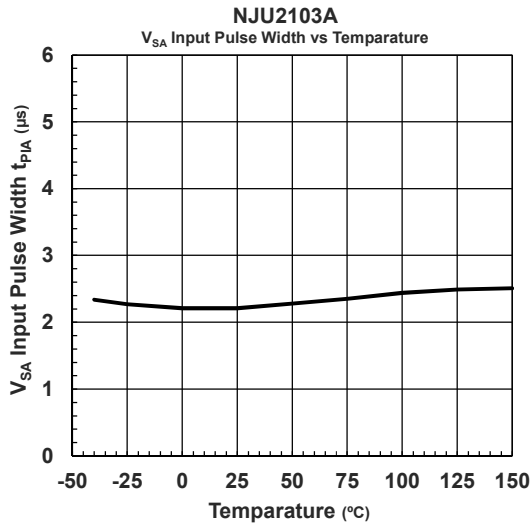
## ■ TYPICAL CHARACTERISTICS

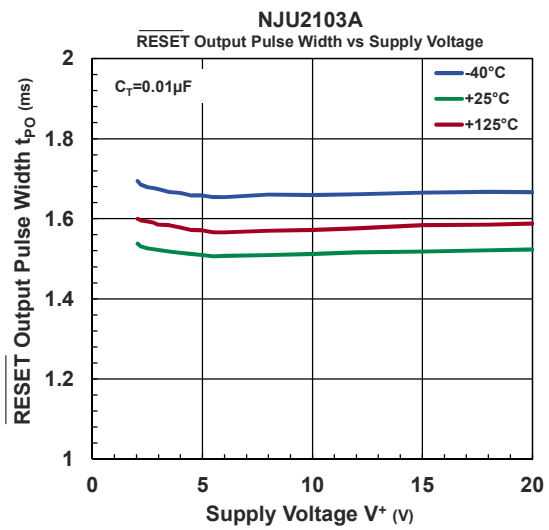
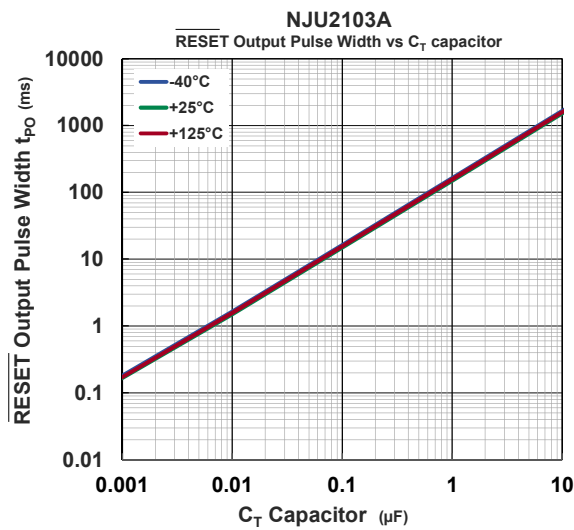
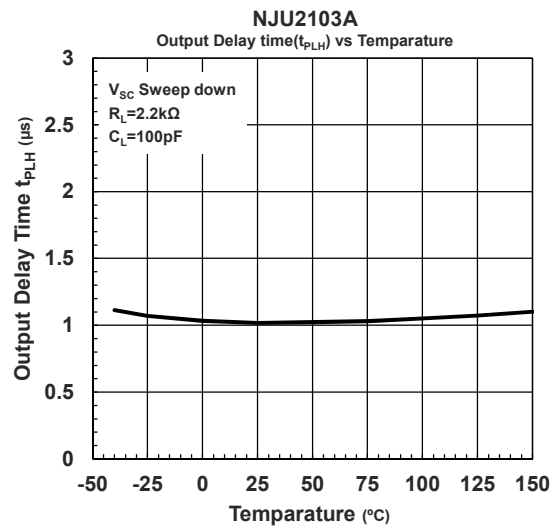
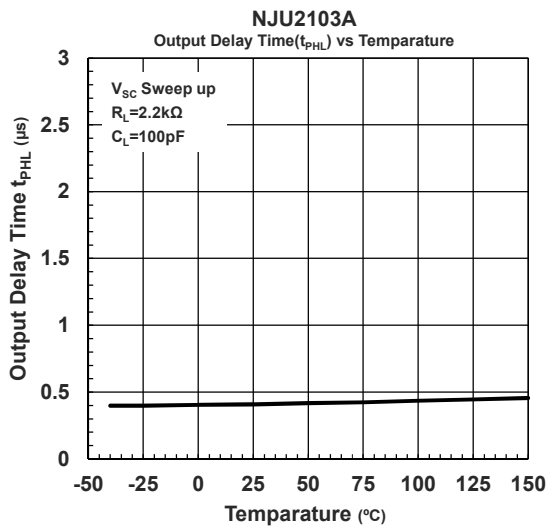










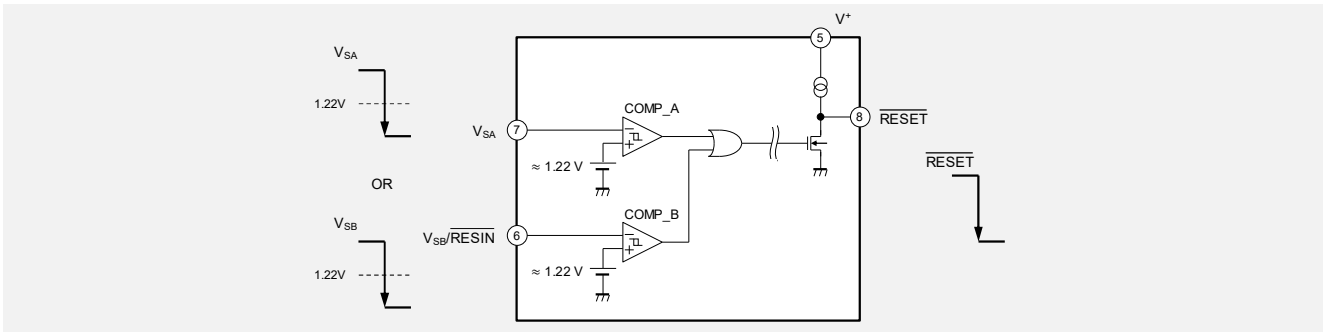


## Technical Information

### FUNCTION EXPLANATION

COMP\_A and COMP\_B are comparator with hysteresis in detection voltage.

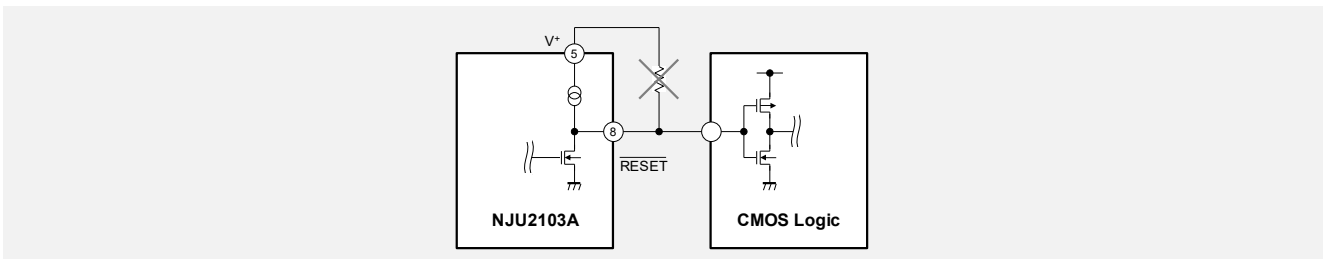
When either  $V_{SA}$  or  $V_{SB}$  pin voltage becomes about 1.22 V or less, the  $\overline{RESET}$  output becomes " Low ".



COMP\_B can be used for arbitrary voltage detection (refer to Fig.3 or Fig.4) and also can be used as a manual reset function with reset hold time by TTL signal input. (refer to Fig.7)

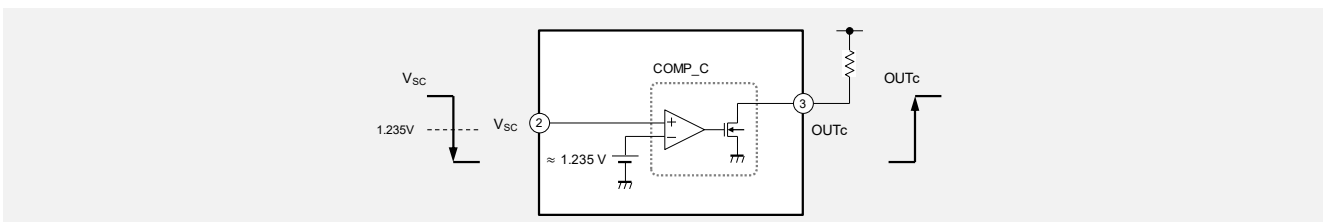
The NJU2103A can detect the instantaneous interruption and the instantaneous drop of the power line with a time of about 2  $\mu$ s width. If this level of instantaneous interruption or drop is not a problem, it can have a delayed trigger function by connecting capacitor to the  $V_{SA}$  and  $V_{SB}$  pins (refer to Fig.9).

Since the  $\overline{RESET}$  pin is internally pulled up to  $V^+$ , an external pull-up resistor isn't required in case of high impedance load like a CMOS logic IC.



COMP\_C is an open-drain output comparator without hysteresis which has anti-polarity input and output.

Therefore, it can be used for overvoltage detection (refer to Fig. 14), positive logic reset output (refer to Fig.8) and generating a reference voltage source.(refer to Fig.11 to 13)

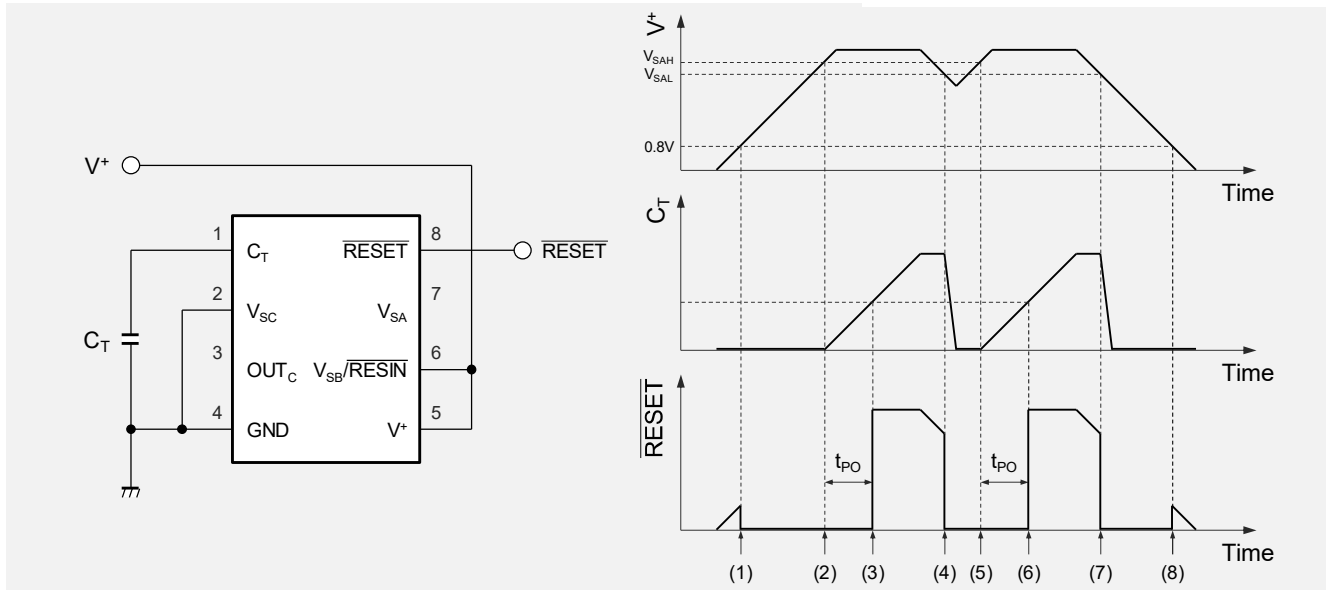


Unused Pin should be treated as shown in the table below.

| Pin. No. | Pin Name                  | Treatment method of unused Pin |
|----------|---------------------------|--------------------------------|
| 2        | $V_{SC}$                  | Connect to GND                 |
| 3        | $OUT_C$                   | OPEN                           |
| 6        | $V_{SB}/\overline{RESIN}$ | Connect to $V^+$               |
| 7        | $V_{SA}$                  | OPEN                           |
| 8        | $\overline{RESET}$        | OPEN                           |

## ■ OPERATION EXPLANATION

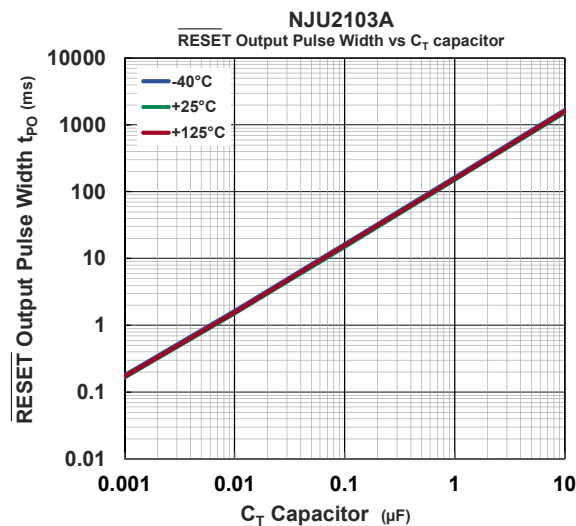
## Technical Information



- (1) When  $V^+$  increases to about 0.8V,  $\overline{\text{RESET}}$  becomes Low
- (2) When  $V^+$  increases to  $V_{SAH}$ , charging to capacitor  $C_T$  starts. At this time,  $\overline{\text{RESET}}$  holds Low.
- (3)  $\overline{\text{RESET}}$  switches from Low to High after the  $\overline{\text{RESET}}$  Output Pulse Width  $t_{PO}$ .

Refer to "Output Pulse Width vs  $C_T$  capacitor" in TYPICAL CHARACTERISTICS and  $t_{PO}$  can be calculated as following formula.

$$\overline{\text{RESET}} \text{ Output Pulse Width } t_{PO} [\text{ms}] \approx 150 \times C_T [\mu\text{F}]$$



- (4) After  $\overline{\text{RESET}}$  becomes High, When  $V^+$  decreases below  $V_{SAL}$ ,  $\overline{\text{RESET}}$  goes Low and discharges  $C_T$ .
- (5) After  $V^+$  decreases below  $V_{SAL}$ , it starts charging  $C_T$  when  $V^+$  increase to  $V_{SAH}$ .  
In case of instantaneous  $V^+$  drop, if the time from  $V^+$  decreases below  $V_{SAL}$  to increase to  $V_{SAH}$  is more than  $V_{SA}$  Input Pulse Width  $t_{PIA}$ , charging will start after discharging  $C_T$ .
- (6)  $V^+$  increase to  $V_{SAH}$  and  $\overline{\text{RESET}}$  switches from Low to High after  $\overline{\text{RESET}}$  Output Pulse Width  $t_{PO}$
- (7) When  $V^+$  becomes less than  $V_{SAL}$ , repeat steps (4) – (6).
- (8) When  $V^+$  decreases to 0 V,  $\overline{\text{RESET}}$  holds Low until  $V^+$  reaches about 0.8 V.

## Technical Information

### ■APPLICATION EXAMPLE

#### 1. 5V Power supply monitor

Monitor the 5V power supply with  $V_{SA}(\text{COMP\_A})$ . The detection voltage at falling is the  $V_{SA}$  detection voltage 1  $V_{SAL}$  (4.2 V typ.), and the detection voltage at rising is the  $V_{SA}$  detection voltage 2  $V_{SAH}$  (4.3V typ.).

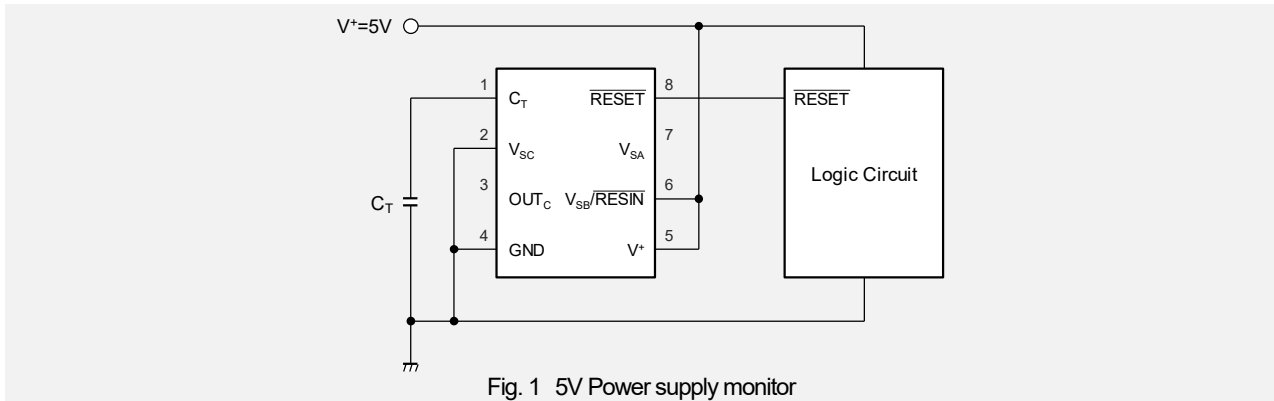


Fig. 1 5V Power supply monitor

#### 2. Power supply monitor (adjust detection voltage by external resistor)

$V_{SA}$  detection voltage1 can be adjusted with an external resistor.

- By selecting the external voltage-dividing resistors  $R_1$  and  $R_2$  to a sufficiently smaller value than internal voltage-dividing resistors  $R'$ ,  $R''$  (100 k $\Omega$ , 41 k $\Omega$ ), the detection voltage can be set by the resistance ratio of  $R_1$  and  $R_2$ .

The formula for calculating detection voltage is as follows and refer to Tab.1 for setting example.

##### Detection voltage calculate formula ( $R_1 \ll 100\text{k}\Omega$ , $R_2 \ll 41\text{k}\Omega$ )

$$\text{Detection Voltage(falling)} = \frac{(R_1 \parallel R') + (R_2 \parallel R'')}{R_2 \parallel R''} \times \frac{R''}{R' + R''} \times V_{SAL} \approx \frac{R_1 + R_2}{R_2} \times 1.2213 \text{ [V]}$$

$$\text{Detection Voltage(rising)} = \frac{(R_1 \parallel R') + (R_2 \parallel R'')}{R_2 \parallel R''} \times \frac{R''}{R' + R''} \times V_{SAH} \approx \frac{R_1 + R_2}{R_2} \times 1.2504 \text{ [V]}$$

Tab. 1 Setting example

| External resistor $R_1$<br>[k $\Omega$ ] | External resistor $R_2$<br>[k $\Omega$ ] | Detection Voltage(falling)<br>[V] | Detection Voltage(rising)<br>[V] |
|------------------------------------------|------------------------------------------|-----------------------------------|----------------------------------|
| 6.2                                      | 2.4                                      | 4.37                              | 4.47                             |
| 12                                       | 5.1                                      | 4.11                              | 4.20                             |

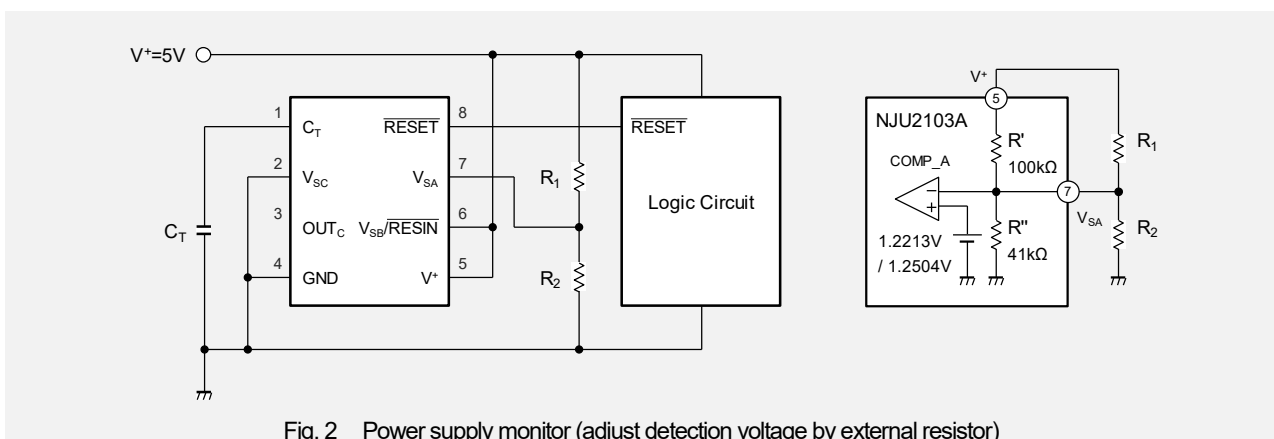


Fig. 2 Power supply monitor (adjust detection voltage by external resistor)

## Technical Information

### 3. Arbitrary power supply monitor (monitoring $V^+ \leq 18V$ )

Monitor the power supply of  $V^+ \leq 18V$  with  $V_{SB}$  (COMP\_B) and voltage-dividing resistors  $R_1$  and  $R_2$ .

- The detection voltage can be set by resistors  $R_1$  and  $R_2$ .

The formula for calculating  $R_1$  and  $R_2$  is as follows and refer to Tab.2.

#### Detection voltage calculate formula

$$\text{Detection Voltage(falling)} = \frac{R_1 + R_2}{R_2} \times V_{SBL} \approx \frac{R_1 + R_2}{R_2} \times 1.220 \text{ [V]}$$

$$\text{Detection Voltage(rising)} = \frac{R_1 + R_2}{R_2} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_1 + R_2}{R_2} \times 1.248 \text{ [V]}$$

- When  $V^+$  is 4.45V or less, connects  $V_{SA}$  (pin 7) to  $V^+$  to disable COMP\_A
- When  $V^+$  is greater than 4.45 V,  $V_{SA}$  (pin 7) should be opened. And in this case, current consumption decreases. (decrease value:  $17.3 \times V^+ \text{ [}\mu\text{A]}$ )

Tab. 2 setting example

| External resistor $R_1$<br>[kΩ] | External resistor $R_2$<br>[kΩ] | Detection Voltage(falling)<br>[V] | Detection Voltage(rising)<br>[V] |
|---------------------------------|---------------------------------|-----------------------------------|----------------------------------|
| 43                              | 16                              | 4.50                              | 4.60                             |
| 22                              | 15                              | 3.01                              | 3.08                             |

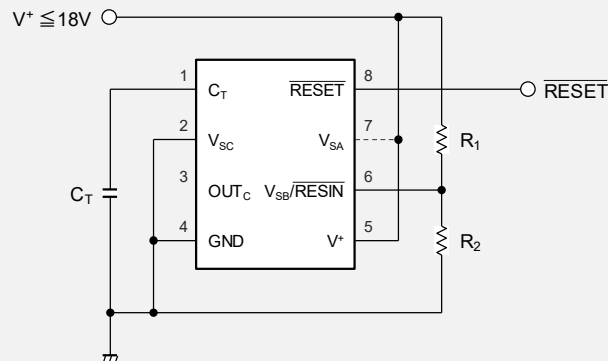


Fig. 3 Arbitrary power supply monitor (monitoring  $V^+ \leq 18V$ )

## Technical Information

### 4. Arbitrary power supply monitor (monitoring $V^+ > 18V$ )

Monitor the power supply of  $V^+ > 18V$  with  $V_{SB}$  (COMP\_B) and voltage-dividing resistors  $R_1$  and  $R_2$ .

The power supply of this IC (about 5V) is generated with  $V_{SC}$  (COMP\_C) and feedback resistors  $R_4$  and  $R_5$ .

- Set the detection voltage with resistors  $R_1$  and  $R_2$  according to the following formula.

#### Detection voltage calculate formula

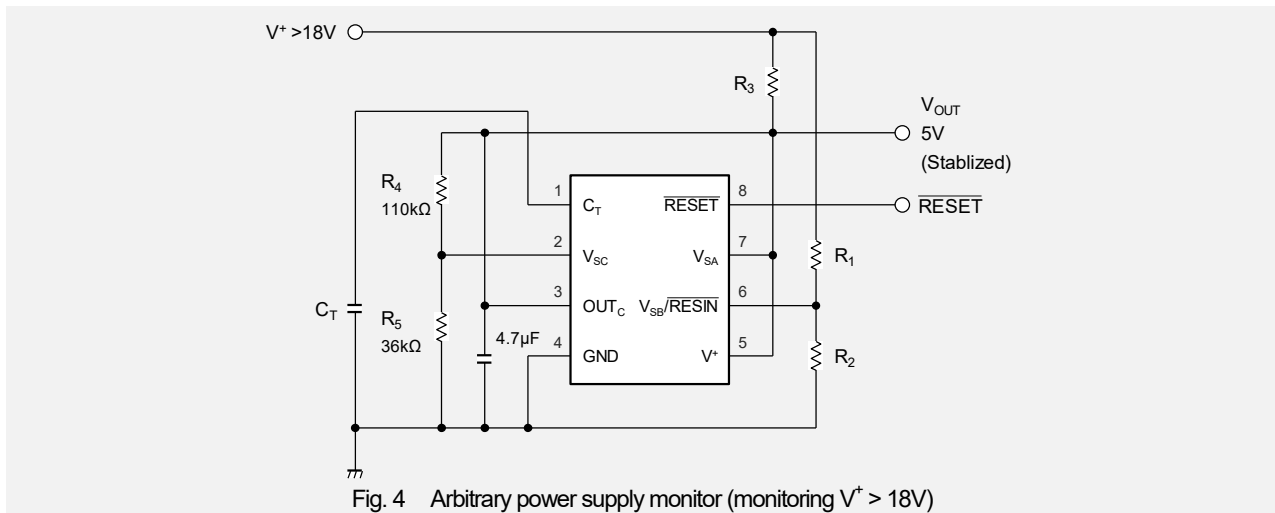
$$\text{Detection Voltage}(V^+ \text{ falling}) = \frac{R_1 + R_2}{R_2} \times V_{SBL} \approx \frac{R_1 + R_2}{R_1} \times 1.220 \text{ [V]}$$

$$\text{Detection Voltage}(V^+ \text{ rising}) = \frac{R_1 + R_2}{R_2} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_1 + R_2}{R_2} \times 1.248 \text{ [V]}$$

- The  $\overline{\text{RESET}}$  output is  $\approx 0V$  (low level) and  $\approx 5V$  (high level). Not outputs  $V^+$  voltage.  
 $\overline{\text{RESET}}$  should not be pulled up to  $V^+$
- If the resistor ratio of  $R_4$  and  $R_5$  is adjusted, high level  $\overline{\text{RESET}}$  voltage is changed according to constant voltage set by resistor ratio of  $R_4$  and  $R_5$ . Constant voltage  $V_{OUT}$  is calculated as the following formula. However, shouldn't be exceed 18V.

$$\text{Constant Voltage } V_{OUT} = \frac{R_4 + R_5}{R_5} \times V_{SC} \approx \frac{R_4 + R_5}{R_5} \times 1.235 \text{ [V]}$$

- The constant voltage (5V output) can be used as the power supply for the small current consumption circuit.
- When deciding the value of  $R_3$ , it is necessary to be careful about power consumption.





## Technical Information

5. 5V, 12V power supply monitor ( dual power supply monitor e.g.  $V_1^+ = 5V$ ,  $V_2^+ = 12V$ )

Monitor the  $V_1^+$ (5V) power supply with  $V_{SA}$ (COMP\_A) and monitor the  $V_2^+$ (12V) power supply with  $V_{SB}$ (COMP\_B) and voltage-dividing resistors  $R_1$  and  $R_2$ .

- $V_1^+$  detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.
- $V_2^+$  detection voltage (falling, rising) is set by  $R_1$  and  $R_2$  according to following formula. In case of resistor value in Fig.5, the detection voltage (falling) is about 9.0 V and the detection voltage (rising) is about 9.2 V.

 **$V_2^+$  detection voltage calculate formula**

$$\text{Detection Voltage}(V_2^+ \text{ falling}) = \frac{R_1 + R_2}{R_2} \times V_{SBL} \approx \frac{R_1 + R_2}{R_2} \times 1.220 \text{ [V]}$$

$$\text{Detection Voltage}(V_2^+ \text{ rising}) = \frac{R_1 + R_2}{R_2} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_1 + R_2}{R_2} \times 1.248 \text{ [V]}$$

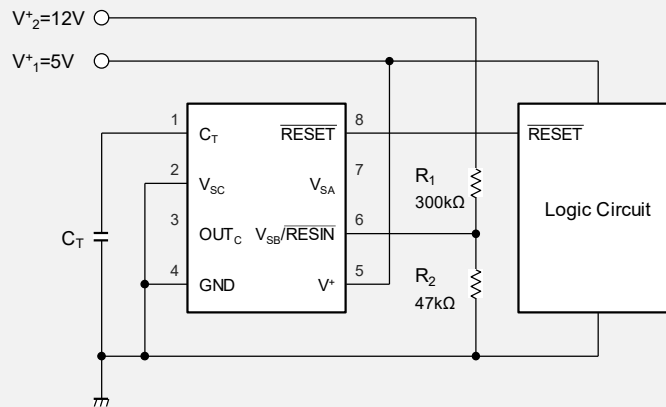


Fig. 5 5V, 12V power supply monitor ( dual power supply monitor :  $V_1^+ = 5V$ ,  $V_2^+ = 12V$ )

## Technical Information

**6. 5V, 12V power supply monitor (e.g.  $V_1^+ = 5V$ ,  $V_2^+ = 12V$ ,  $\overline{\text{RESET}}$  output is only  $V_1^+$  detection result)**

Monitor the  $V_1^+$ (5V) power supply with  $V_{SA}(\text{COMP\_A})$  and output signal from  $\overline{\text{RESET}}$ .

In addition, monitor the  $V_2^+$ (12V) power supply with  $V_{SC}(\text{COMP\_C})$  and voltage-dividing resistors  $R_1$ ,  $R_2$ ,  $R_3$ ,  $R_4$ , NPN transistor, base current limiting resistor  $R_5$ , and output signal from  $\text{OUT}_C$ .

- $V_1^+$  detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.
- $V_2^+$  detection voltage (falling) and hysteresis width at rising are calculated as following formula. In case of resistor value in Fig.6, the detection voltage (falling) is about 9.0 V and the hysteresis width at rising is about 0.2 V

 **$V_2^+$  detection voltage and hysteresis width at rising calculate formula**

$$\text{Detection Voltage}(V_2^+ \text{ falling}) = \frac{R_1 + R_2 + R_3}{R_2 + R_3} \times V_{SC} \approx \frac{R_1 + R_2 + R_3}{R_2 + R_3} \times 1.235 \text{ [V]}$$

$$\text{Hysteresis width at rising} = \frac{R_1(R_3 - R_3 \parallel R_4)}{(R_2 + R_3)(R_2 + R_3 \parallel R_4)} \times V_{SC} \approx \frac{R_1(R_3 - R_3 \parallel R_4)}{(R_2 + R_3)(R_2 + R_3 \parallel R_4)} \times 1.235 \text{ [V]}$$

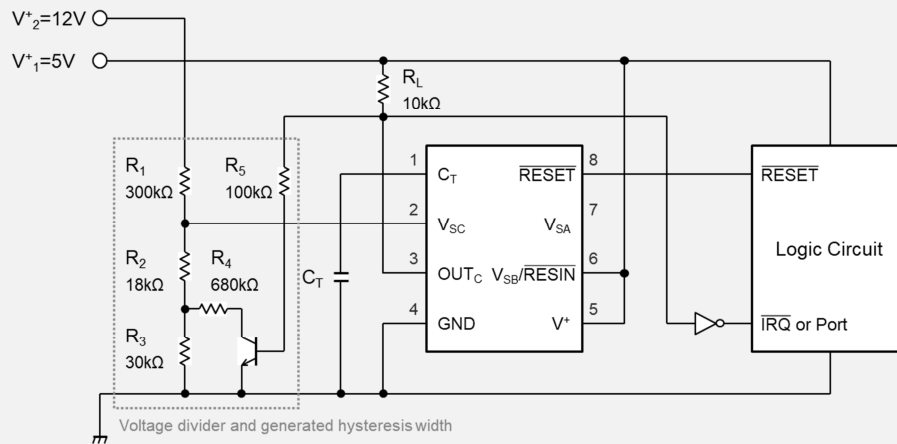


Fig. 6 5V, 12V power supply monitor ( $V_1^+ = 5V$ ,  $V_2^+ = 12V$ ,  $\overline{\text{RESET}}$  output is only  $V_1^+$  detection result)

**7. Manual Reset function ( $V^+ = 5V$ )**

By inputting the TTL signal to  $V_{SB}/\overline{\text{RESIN}}$ , it realizes manual reset output signal regardless of the state of  $V^+$ .

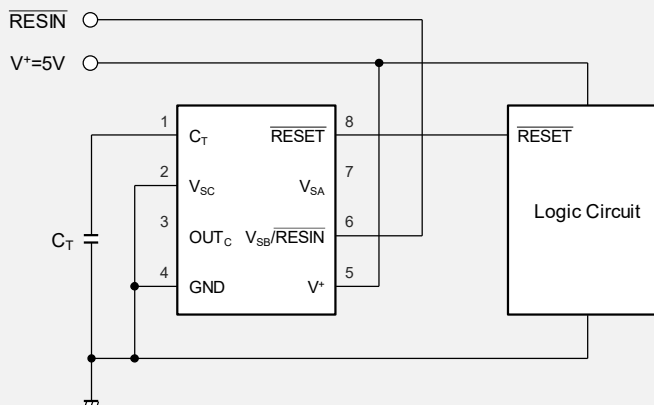


Fig. 7 Manual Reset function ( $V^+ = 5V$ )

## Technical Information

### 8. Non-inverting reset output

If a positive output is required for reset signal, invert the  $\overline{\text{RESET}}$  output with COMP\_C and output from OUT<sub>C</sub>. Since OUT<sub>C</sub> is an open drain output, It is required the pull-up resistor.(shown as R<sub>L</sub> in Fig.8)

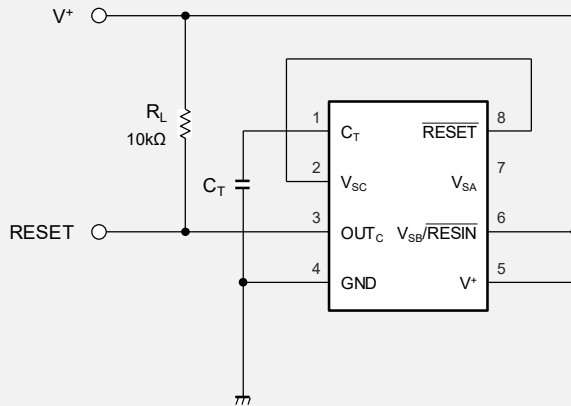


Fig. 8 Non-inverting reset output

### 9. Power supply voltage monitoring by delayed trigger

An arbitrary delay is added to the COMP\_A operation by connecting capacitor C<sub>1</sub> between V<sub>SA</sub> and GND.

When C<sub>1</sub> is connected minimum input pulse width becomes longer. E.g. t<sub>PI</sub> = 40μs (C<sub>1</sub>=1000pF)

Minimum input pulse width t<sub>PI</sub> is calculated as following formula.

#### Minimum input pulse width calculate formula

$$t_{PI} [\mu s] \approx (R' \parallel R'') \times \ln \left( \frac{5 - 4}{V_{SAL} - 4} \right) \times 10^{-6} \times C_1 [pF] \approx 4.7 \times 10^{-2} \times C_1 [pF]$$

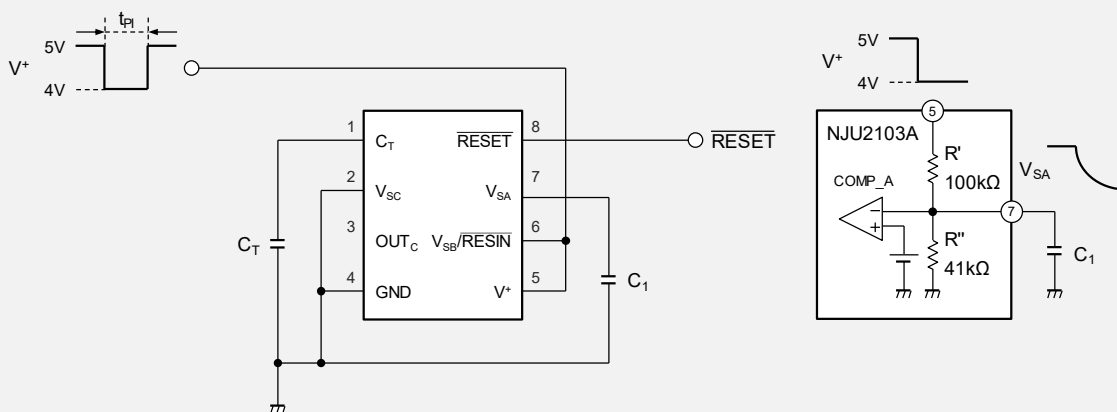


Fig. 9 Power supply voltage monitoring by delayed trigger

## Technical Information

10. Positive and negative dual power supply monitoring (e.g.  $V^+ = 5V$ ,  $V^-$ =negative voltage)

Monitor the positive power supply with  $V_{SA}$  (COMP\_A) and monitor the negative voltage with  $V_{SB}$  (COMP\_B)

$V_{SC}$  (COMP\_C) is used to shift negative voltage to positive voltage.

- $R_1$ ,  $R_2$ ,  $R_3$  should have the same resistance value
- $V^+$  detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.
- $V^-$  detection voltage is calculated as following formula. In case of resistor value in Fig.10, the detection voltage (falling) is about -4.3V and the detection voltage (rising) is about -4.4V.

 **$V^-$  detection voltage calculate formula**

$$\text{Detection voltage } (V^- \text{ falling}) = \frac{R_3 + R_4}{R_3} \times V_{SC} - \frac{2R_4}{R_3} \times V_{SBL} \approx \frac{R_3 - R_4}{R_3} \times 1.220 \text{ [V]}$$

$$\text{Detection voltage } (V^- \text{ rising}) = \frac{R_3 + R_4}{R_3} \times V_{SC} - \frac{2R_4}{R_3} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_3 - R_4}{R_3} \times 1.248 \text{ [V]}$$

- When using a power supply that outputs  $V^-$  without  $V^+$  output, it is necessary to connect a Schottky barrier diode (SBD) between the  $V_{SC}$  and GND to prevent being applied negative voltage to this IC.

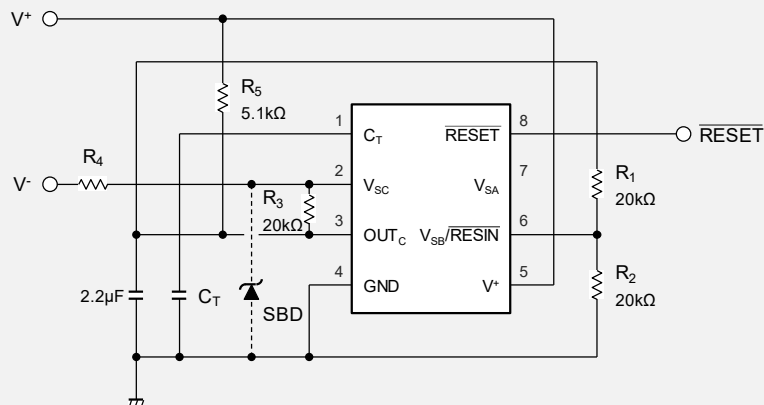


Fig. 10 Positive and negative dual power supply monitoring ( $V^+ = 5V$ ,  $V^-$ =negative voltage)

## Technical Information

### 11. Reference voltage output and voltage drop monitoring (e.g. 9V reference output, 5V, 9V monitoring)

Monitor the  $V^+$  (5V) power supply with  $V_{SA}(COMP\_A)$ .

9V Reference voltage  $V_{OUT}$  is generated by  $V_{SC}(COMP\_C)$ , feedback resistors  $R_3$ ,  $R_4$  and NPN transistor, and its 9V is monitored by  $V_{SB}(COMP\_B)$  and voltage-dividing resistors  $R_1$  and  $R_2$ .

- $V^+$  detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.
- The reference voltage  $V_{OUT}$  and its detection voltage are calculated as the following formula. In case of resistor value in Fig.11, the reference voltage  $V_{OUT}$  is about 9.0V, the detection voltage ( $V_{OUT}$  falling) is about 7.1V and the detection voltage ( $V_{OUT}$  rising) is about 7.3V

#### Reference Voltage and detection voltage calculate formula

$$\text{Reference voltage } V_{OUT} = \frac{R_3 + R_4}{R_4} \times V_{SC} \approx \frac{R_3 + R_4}{R_4} \times 1.235 \text{ [V]}$$

$$\text{Detection Voltage}(V_{OUT} \text{ falling}) = \frac{R_1 + R_2}{R_2} \times V_{SBL} \approx \frac{R_1 + R_2}{R_2} \times 1.220 \text{ [V]}$$

$$\text{Detection Voltage}(V_{OUT} \text{ rising}) = \frac{R_1 + R_2}{R_2} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_1 + R_2}{R_2} \times 1.248 \text{ [V]}$$

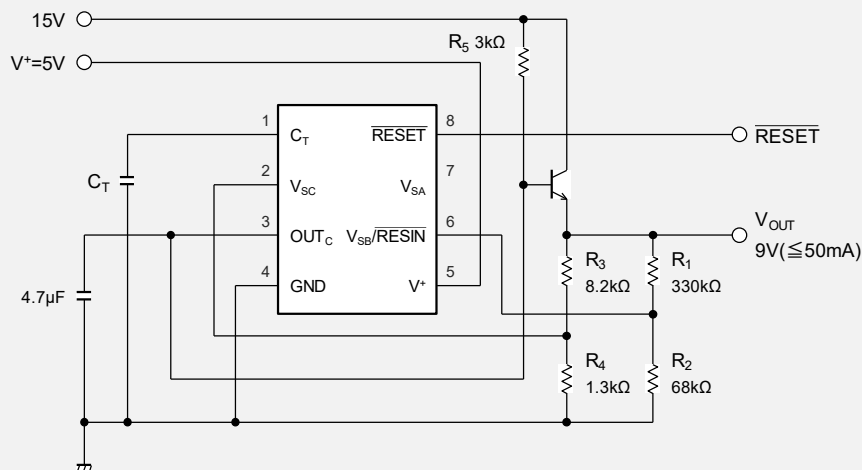


Fig. 11 Reference voltage output and voltage drop monitoring (e.g. 9V reference output, 5V, 9V monitoring)

## Technical Information

### 12-1. Reference Voltage output and Voltage drop monitoring (e.g. 5V output, 5V monitoring)

5V Reference voltage  $V_{OUT}$  is generated by  $V_{SC}(COMP\_C)$ , feedback resistors  $R_3$ ,  $R_4$ , NPN transistor and drive resistor  $R_5$ , and its 5V is monitored by  $V_{SA}(COMP\_A)$ .

- The reference voltage  $V_{OUT}$  is calculated as following formula. In case of resistor value In Fig.12-1, the reference voltage  $V_{OUT}$  is about 5.0V

### Reference Voltage calculate formula

$$\text{Reference voltage } V_{OUT} = \frac{R_3 + R_4}{R_3} \times V_{SC} \approx \frac{R_3 + R_4}{R_3} \times 1.235 \text{ [V]}$$

- The detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.

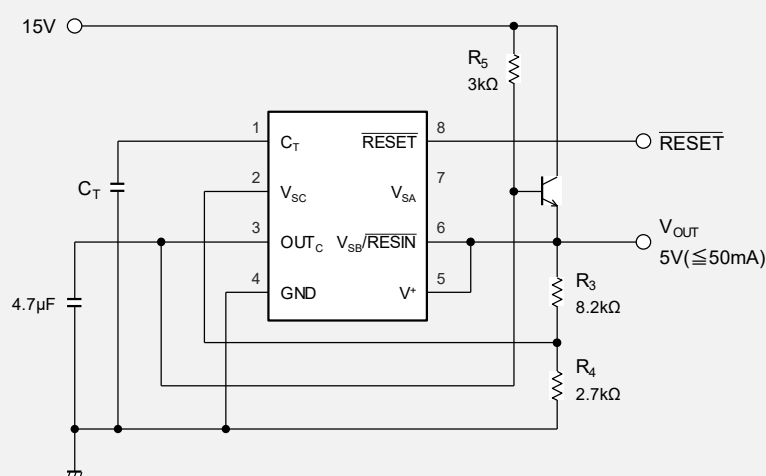


Fig. 12-1 Reference Voltage output and Voltage drop monitoring (e.g. 5V output, 5V monitoring)

## Technical Information

### 12-2. Reference Voltage output and Voltage drop monitoring (e.g. 5V output, 5V monitoring)

5V Reference voltage  $V_{OUT}$  is generated by  $V_{SC}(COMP\_C)$  and feedback resistors  $R_2$ ,  $R_3$ , and its 5V is monitored by  $V_{SA}(COMP\_A)$ . Unlike Fig.12-1, 5V output cannot supply large current.

- The reference voltage  $V_{OUT}$  is calculated as following formula.

In case of resistor value In Fig.12-2, the reference voltage  $V_{OUT}$  is about 5.0V

$$\text{Reference voltage } V_{OUT} = \frac{R_2 + R_3}{R_3} \times V_{SC} \approx \frac{R_2 + R_3}{R_3} \times 1.235 \text{ [V]}$$

- The detection voltage (falling) is 4.2V and detection voltage (rising) is 4.3V.
- $R_1$  value should be calculated from current consumption of NJU2103A, the current flowing through  $R_2$  and  $R_3$ , and 5V output current. .

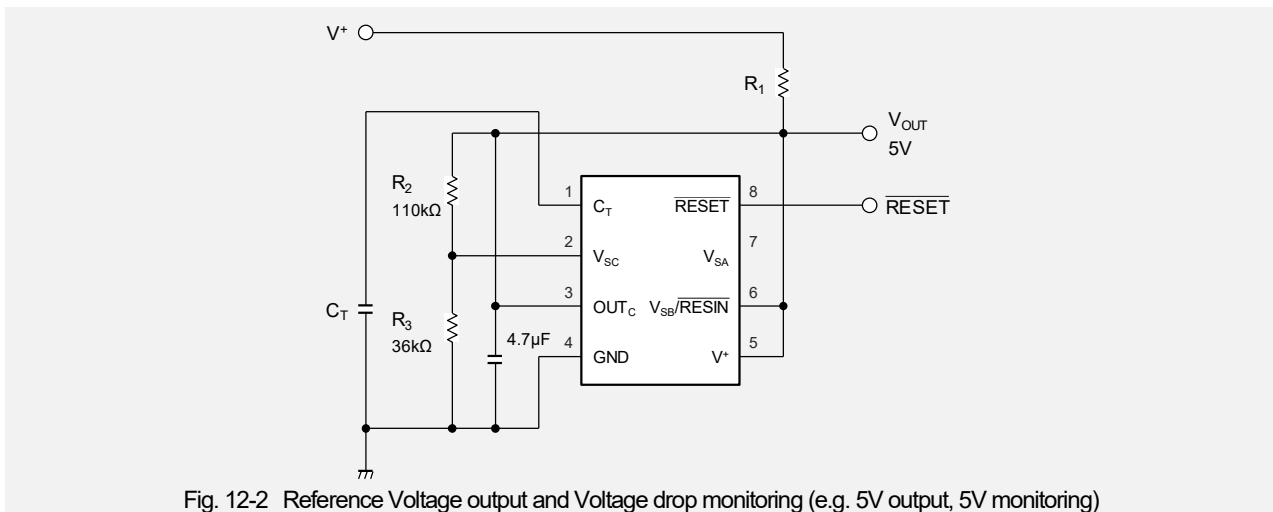


Fig. 12-2 Reference Voltage output and Voltage drop monitoring (e.g. 5V output, 5V monitoring)

### 13. Reference Voltage output and Voltage drop monitoring (e.g. 1.235V output, 5V monitoring)

Buffer-connect the  $V_{SC}(COMP\_C)$  and output the reference voltage of  $COMP\_C$ .

The output current of the reference voltage output is limited by  $R_1$ . If  $R_1$  is 1.2 kΩ, it can output about 2 mA

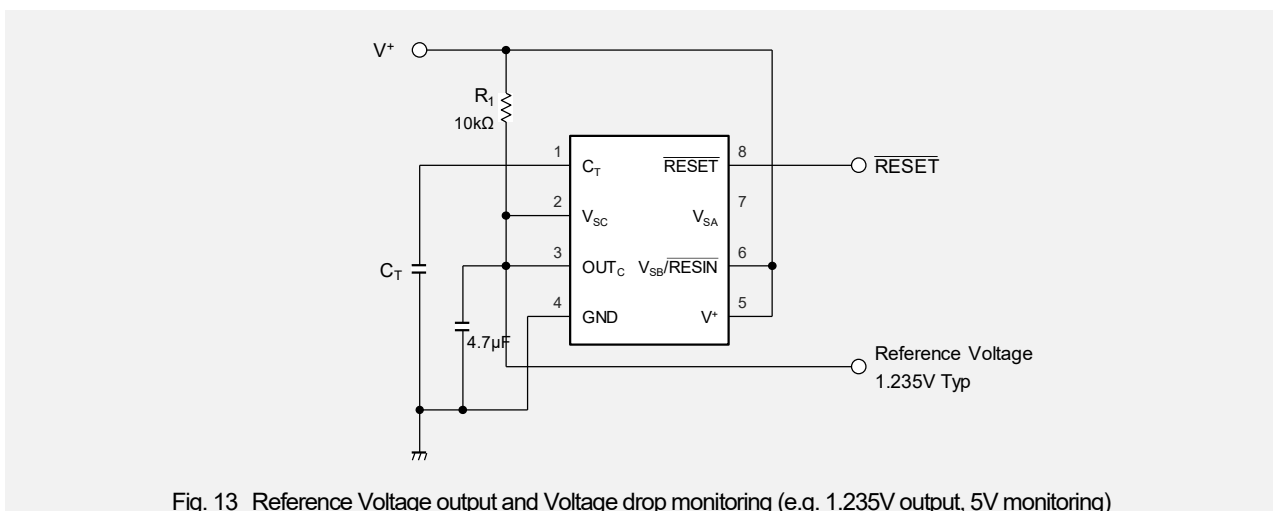


Fig. 13 Reference Voltage output and Voltage drop monitoring (e.g. 1.235V output, 5V monitoring)

## Technical Information

### 14. Low voltage and over voltage detection ( $V^+ = 5V$ )

$V_{SB}$  (COMP\_B) for low voltage detection and  $V_{SC}$  (COMP\_C) for overvoltage detection.

Logically synthesizes low voltage and over voltage detection by connecting  $OUT_C$  to  $V_{SA}$ , and output from  $\overline{RESET}$ .

- Low voltage detection and over voltage detection are calculated as following formula.

$$\text{Low voltage detection (falling)} V_{SL1} = \frac{R_1 + R_2}{R_2} \times V_{SBL} \approx \frac{R_1 + R_2}{R_2} \times 1.220 \text{ [V]}$$

$$\text{Low voltage detection (rising)} V_{SL2} = \frac{R_1 + R_2}{R_2} \times (V_{SBL} + V_{HRSB}) \approx \frac{R_1 + R_2}{R_2} \times 1.248 \text{ [V]}$$

$$\text{Over voltage detection } V_{SH} = \frac{R_3 + R_4}{R_4} \times V_{SC} \approx \frac{R_3 + R_4}{R_4} \times 1.235 \text{ [V]}$$

- There is no hysteresis characteristics for over voltage detection.
- " $\overline{RESET}$  Output Pulse Width  $t_{PO}$ " is valid even when overvoltage is detected

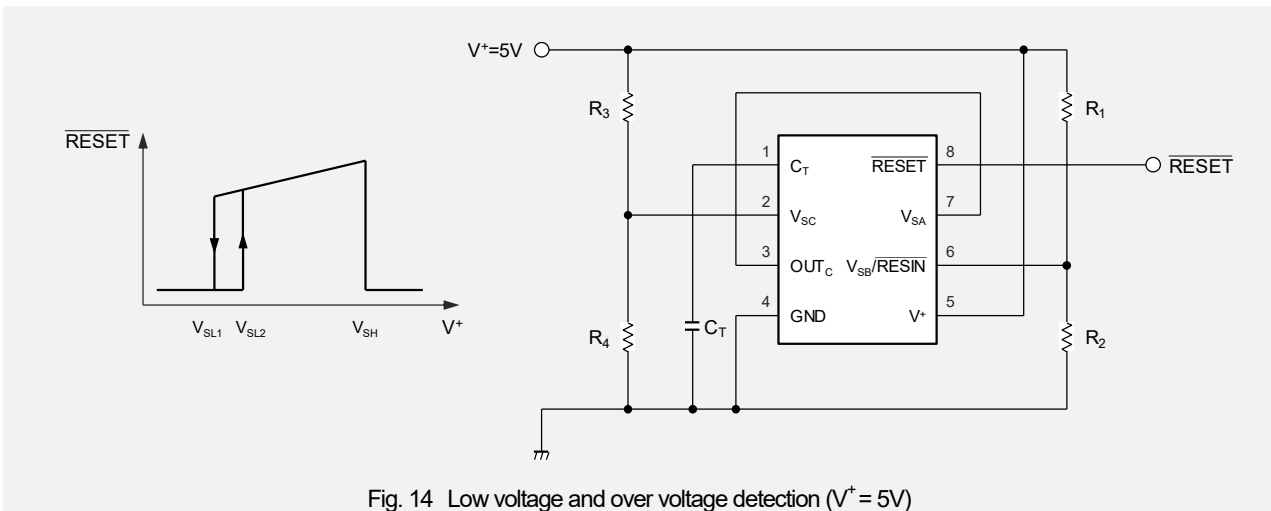
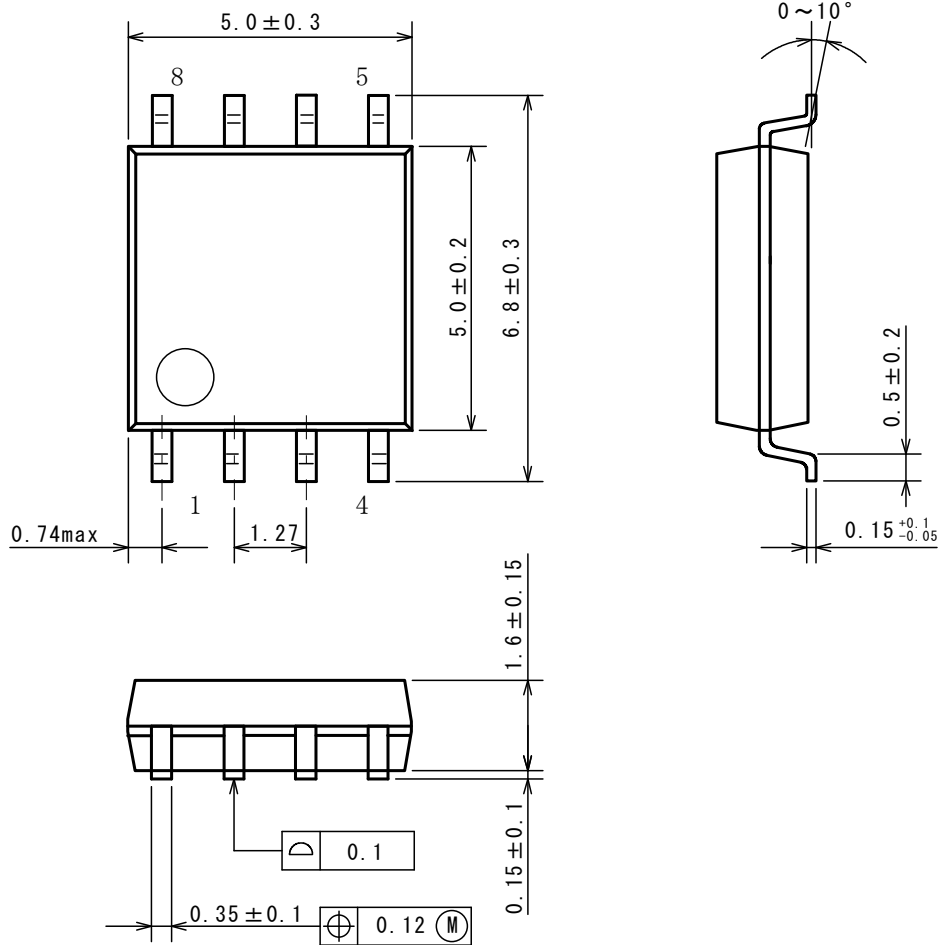


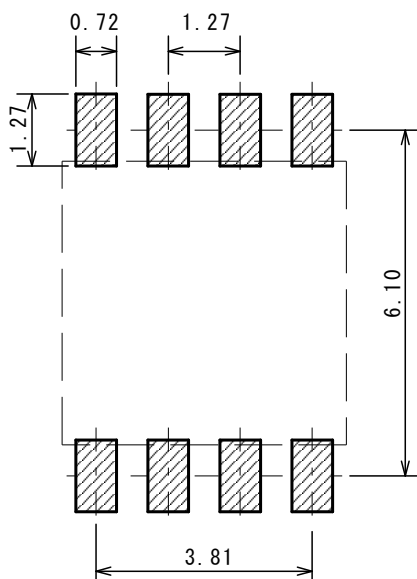
Fig. 14 Low voltage and over voltage detection ( $V^+ = 5V$ )



### ■PACKAGE DIMENSIONS



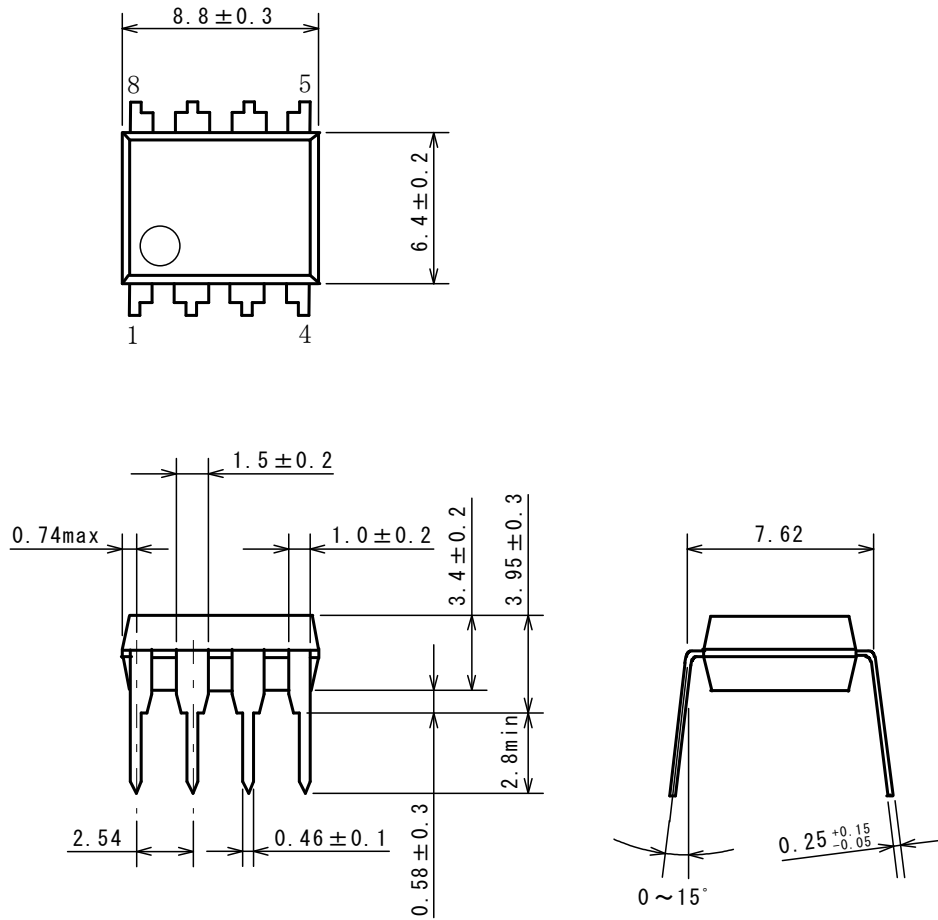
### ■EXAMPLE OF SOLDER PADS DIMENSIONS



## ■PACKAGE DIMENSIONS

**DIP8**

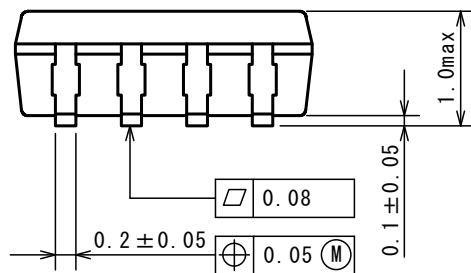
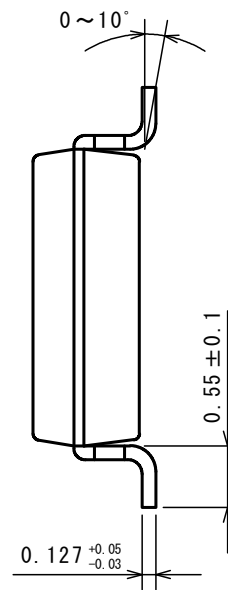
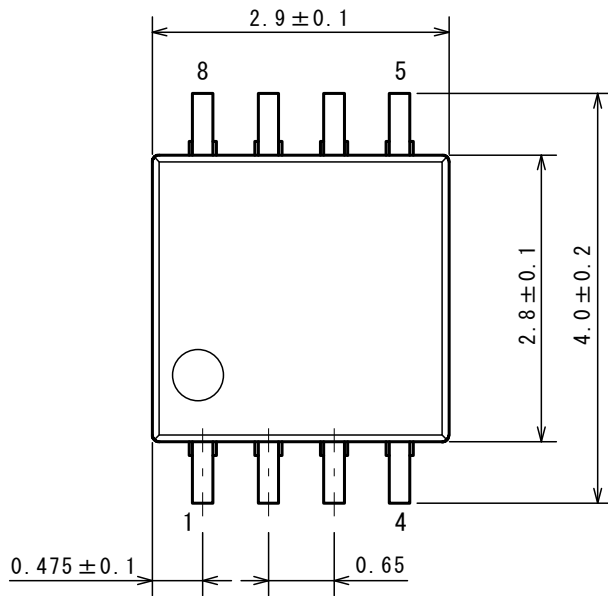
Unit: mm



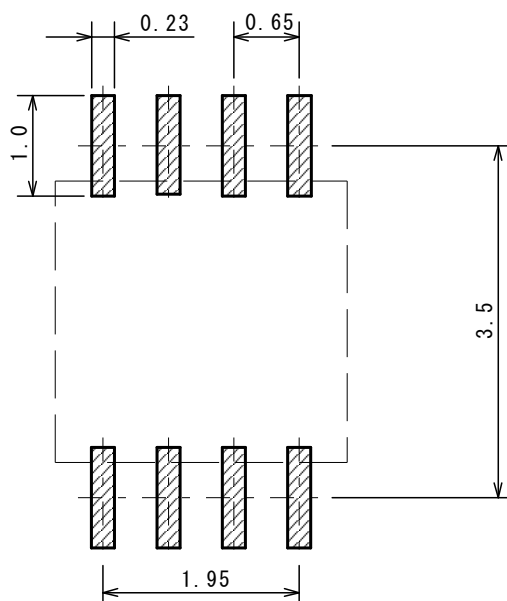
## TVSP8(MSOP8) MEET JEDEC MO-187-DA/THIN TYPE

### ■PACKAGE DIMENSIONS

Unit: mm

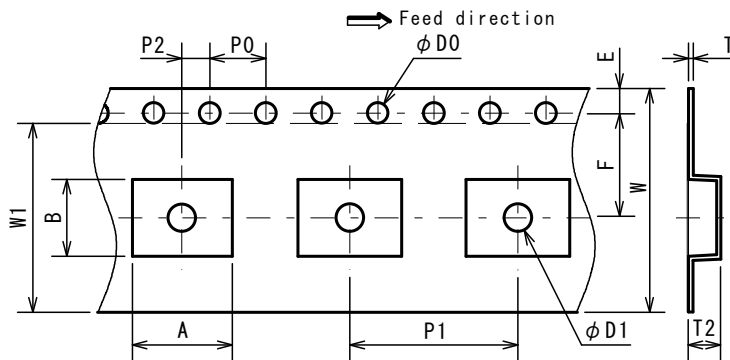


### ■EXAMPLE OF SOLDER PADS DIMENSIONS



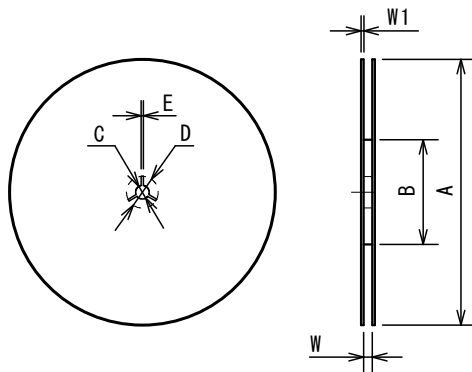
### PACKING SPEC

#### TAPING DIMENSIONS



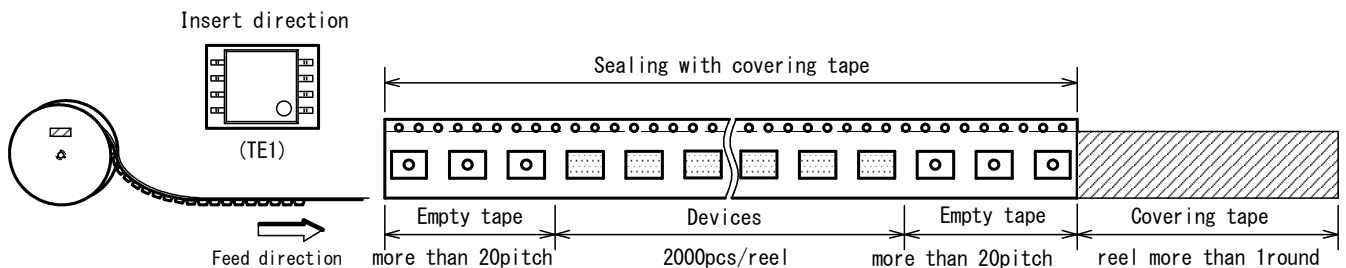
| SYMBOL | DIMENSION       | REMARKS          |
|--------|-----------------|------------------|
| A      | 7.1             | BOTTOM DIMENSION |
| B      | 5.4             | BOTTOM DIMENSION |
| D0     | $1.55 \pm 0.05$ |                  |
| D1     | $2.05 \pm 0.1$  |                  |
| E      | $1.75 \pm 0.1$  |                  |
| F      | $7.5 \pm 0.1$   |                  |
| P0     | $4.0 \pm 0.1$   |                  |
| P1     | $12.0 \pm 0.1$  |                  |
| P2     | $2.0 \pm 0.1$   |                  |
| T      | $0.3 \pm 0.05$  |                  |
| T2     | 2.3             |                  |
| W      | $16.0 \pm 0.3$  |                  |
| W1     | 13.5            | THICKNESS 0.1max |

#### REEL DIMENSIONS

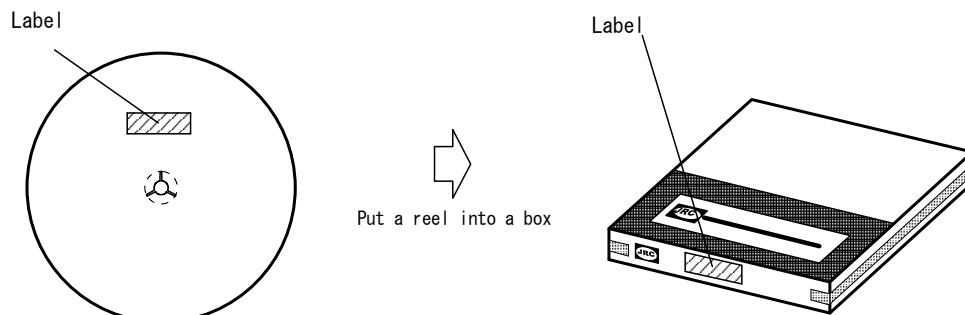


| SYMBOL | DIMENSION         |
|--------|-------------------|
| A      | $\phi 330 \pm 2$  |
| B      | $\phi 80 \pm 1$   |
| C      | $\phi 13 \pm 0.2$ |
| D      | $\phi 21 \pm 0.8$ |
| E      | $2 \pm 0.5$       |
| W      | $17.5 \pm 0.5$    |
| W1     | $2 \pm 0.2$       |

#### TAPING STATE



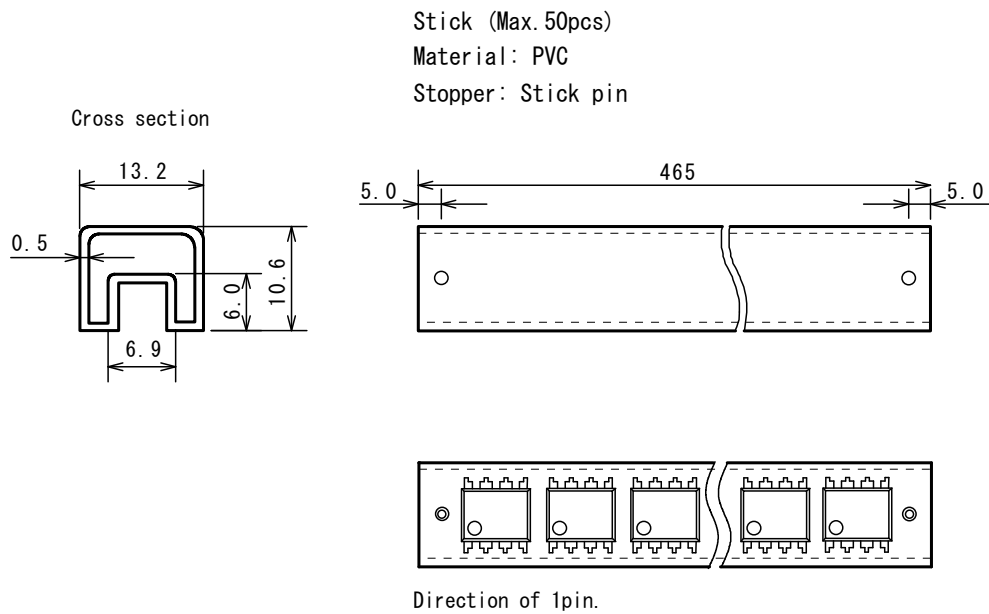
#### PACKING STATE



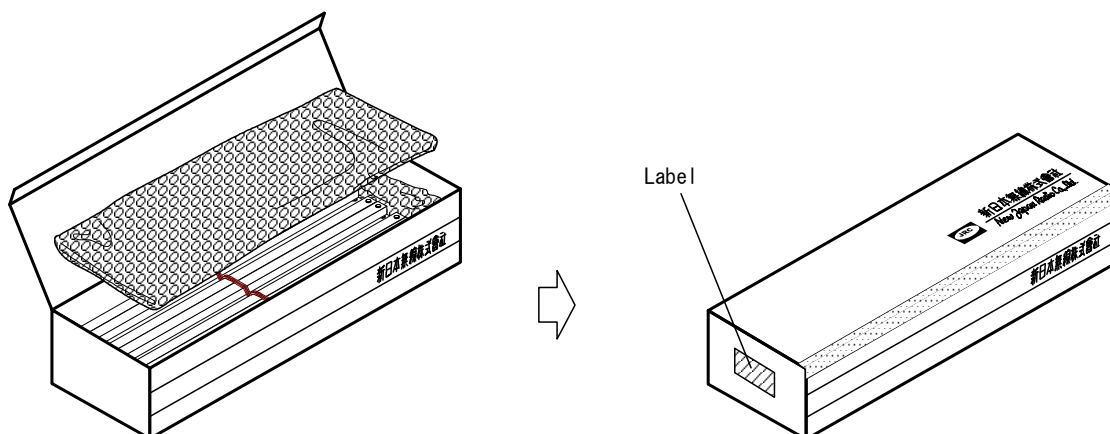
### ■PACKING SPEC

Unit: mm

#### STICK STATE



#### PACKING STATE

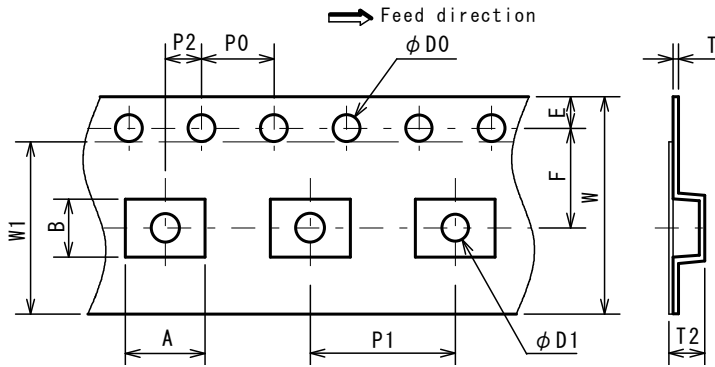


## TVSP8(MSOP8) MEET JEDEC MO-187-DA/THIN TYPE

### PACKING SPEC

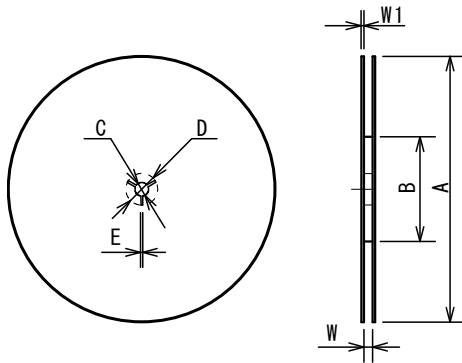
Unit: mm

#### TAPING DIMENSIONS



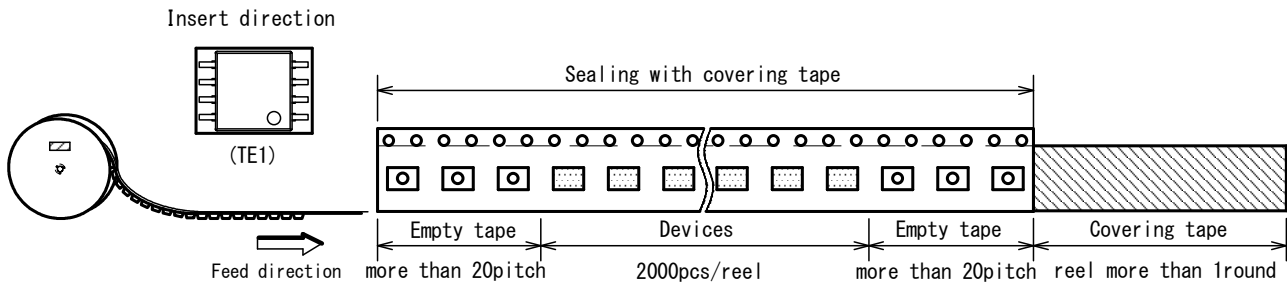
| SYMBOL | DIMENSION       | REMARKS          |
|--------|-----------------|------------------|
| A      | 4.4             | BOTTOM DIMENSION |
| B      | 3.2             | BOTTOM DIMENSION |
| D0     | $1.5^{+0.1}_0$  |                  |
| D1     | $1.5^{+0.1}_0$  |                  |
| E      | $1.75 \pm 0.1$  |                  |
| F      | $5.5 \pm 0.05$  |                  |
| P0     | $4.0 \pm 0.1$   |                  |
| P1     | $8.0 \pm 0.1$   |                  |
| P2     | $2.0 \pm 0.05$  |                  |
| T      | $0.30 \pm 0.05$ |                  |
| T2     | 1.75 (MAX.)     |                  |
| W      | $12.0 \pm 0.3$  |                  |
| W1     | 9.5             | THICKNESS 0.1max |

#### REEL DIMENSIONS

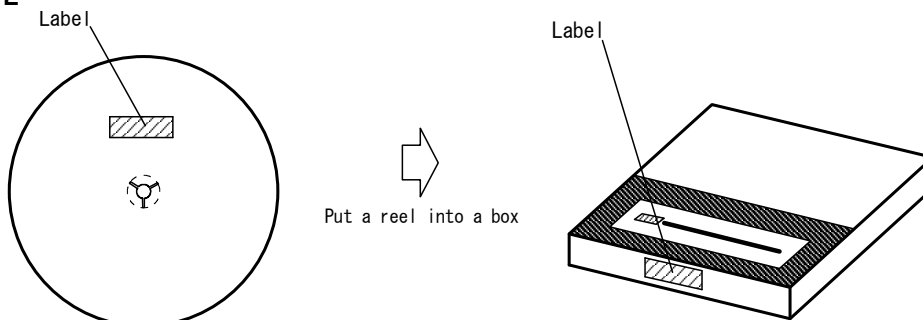


| SYMBOL | DIMENSION         |
|--------|-------------------|
| A      | $\phi 254 \pm 2$  |
| B      | $\phi 100 \pm 1$  |
| C      | $\phi 13 \pm 0.2$ |
| D      | $\phi 21 \pm 0.8$ |
| E      | $2 \pm 0.5$       |
| W      | $13.5 \pm 0.5$    |
| W1     | $2.0 \pm 0.2$     |

#### TAPING STATE

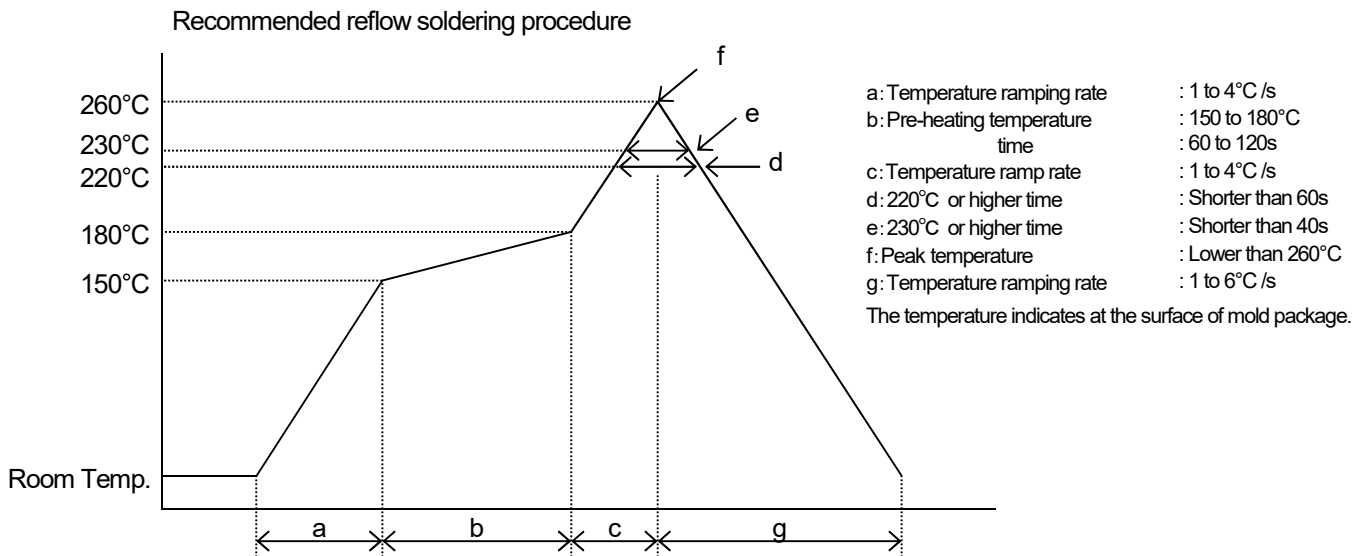


#### PACKING STATE

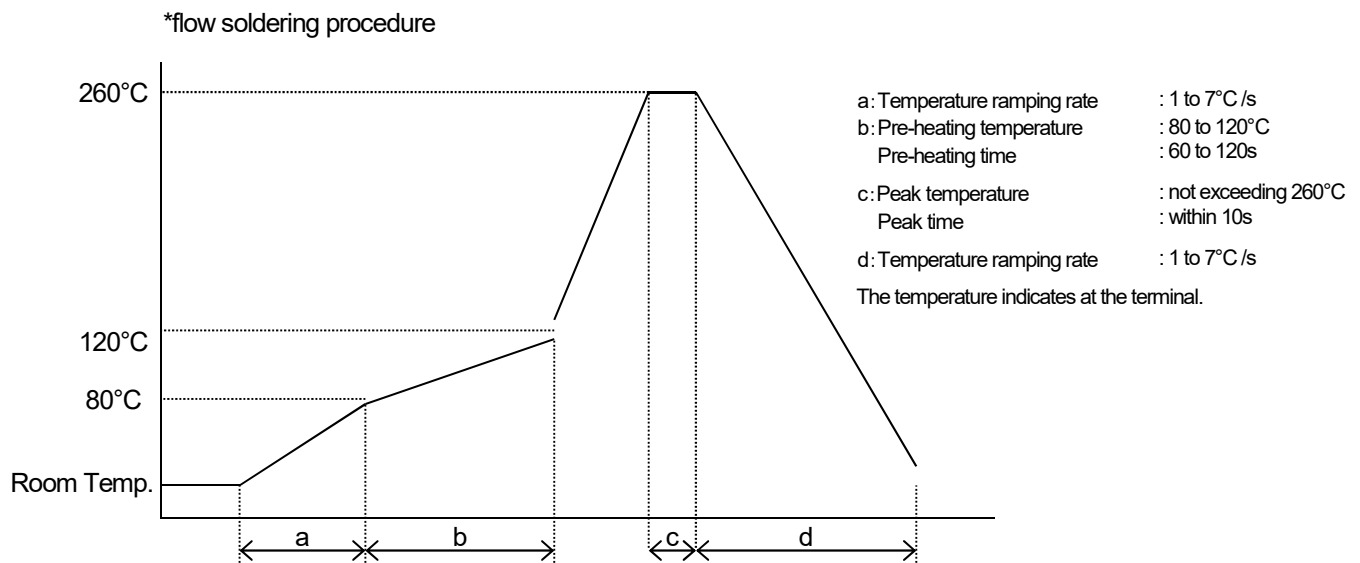


## ■RECOMMENDED MOUNTING METHOD

### INFRARED REFLOW SOLDERING METHOD



### INFRARED FLOW SOLDERING METHOD



**■REVISION HISTORY**

| Date         | Revision | Changes                                                                                              |
|--------------|----------|------------------------------------------------------------------------------------------------------|
| 11.Jun.2018. | 1.0      | New Release                                                                                          |
| 22.Aug.2018  | 1.1      | Add the Technical Information.<br>(FUNCTION EXPLAMATION, OPERATION EXPLAMATION, APPLICATION EXAMPLE) |
| 07.Feb.2019  | 1.2      | Correction of error                                                                                  |



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