TEXAS INSTRUMENTS

CC2541-Q1
SWRS128 – JUNE 2014

CC2541-Q1 SimpleLink™ *Bluetooth*® Low Energy Wireless MCU for Automotive

1 Device Overview

1.1 Features

- RF
  - 2.4-GHz *Bluetooth* Low Energy Compliant and Proprietary RF Wireless MCU
  - Supports Data Rates of 250 kbps, 500 kbps, 1 Mbps, and 2 Mbps
  - Excellent Link Budget, Enabling Long-Range Applications Without External Front End
  - Programmable Output Power up to 0 dBm
  - Excellent Receiver Sensitivity (–94 dBm at 1 Mbps), Selectivity, and Blocking Performance
  - Suitable for Systems Targeting Compliance With Worldwide Radio Frequency Regulations: ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan)
- Layout
  - Few External Components
  - 6 mm × 6 mm QFN-40 Package
- Low Power
  - Active-Mode RX Down to: 18.3 mA
  - Active-Mode TX (0 dBm): 18.6 mA
  - Power Mode 1 (4-µs Wake-Up): 270 µA
  - Power Mode 2 (Sleep Timer On): 1 µA
  - Power Mode 3 (External Interrupts): 0.5 µA
  - Wide Supply-Voltage Range (2 V to 3.6 V)
- Microcontroller
  - High-Performance and Low-Power 8051 Microcontroller Core With Code Prefetch
  - 256KB In-System Programmable Flash
  - 8KB of RAM With Retention in All Power Modes
  - Hardware Debug Support
  - Extensive Baseband Automation, Including Auto-Acknowledgment and Address Decoding
  - Retention of All Relevant Registers in All Power Modes
- Peripherals
  - Powerful Five-Channel DMA
  - IR Generation Circuitry
  - General-Purpose Timers (One 16-Bit, Two 8-Bit)
  - 32-kHz Sleep Timer With Capture

- Accurate Digital RSSI Support
- Battery Monitor and Temperature Sensor
- 12-Bit ADC With Eight Channels and Configurable Resolution
- AES Security Coprocessor
- Two Powerful USARTs With Support for Several Serial Protocols
- 23 General-Purpose I/O Pins (21 × 4 mA, 2 × 20 mA)
- I^{2}C interface
- 2 I/O Pins Have LED Driving Capabilities
- Watchdog Timer
- Integrated High-Performance Comparator
- Development Tools
  - CC2541 Evaluation Module
  - SmartRF™ Software
  - IAR Embedded Workbench™ Available
- *Bluetooth* v4.0 Compliant Protocol Stack for Single-Mode BLE Solution
  - Complete Power-Optimized Stack, Including Controller and Host
    - GAP – Central, Peripheral, Observer, or Broadcaster (Including Combination Roles)
    - ATT / GATT – Client and Server
    - SMP – AES-128 Encryption and Decryption
    - L2CAP
  - Sample Applications and Profiles
    - Generic Applications for GAP Central and Peripheral Roles
    - Proximity, Accelerometer, Simple Keys, and Battery GATT Services
    - More Applications Supported in BLE Software Stack
  - Multiple Configuration Options
    - Single-Chip Configuration, Allowing Applications to Run on CC2541-Q1
    - Network Processor Interface for Applications Running on an External Microcontroller
  - BTool–Windows PC Application for Evaluation, Development, and Test
  - Over the Air Update Capable

1.2 Applications

- 2.4-GHz *Bluetooth* Low-Energy Systems
- Proprietary 2.4-GHz Systems
- Keyless Entry (Passive and Remote)
- Tire Pressure Monitoring
- Proximity Sensing
- Interface and Control

- Diagnostics and Maintenance
- Cable Replacement
- Sensor Nodes
- Infotainment and Media
- Smart Phone Connectivity
- Beacons

1.3 Description

The CC2541-Q1 is a power-optimized true Wireless MCU solution for both *Bluetooth* low energy and proprietary 2.4-GHz applications. This device enables the building of robust nework nodes with low total bill-of-material costs. The CC2541-Q1 combines the excellent performance of a leading RF transceiver with an industry-standard enhanced 8051 MCU, in-system programmable flash memory, 8KB of RAM, and many other powerful supporting features and peripherals. The CC2541-Q1 is highly suited for systems in which ultralow power consumption is required, which is specified by various operating modes. Short transition times between operating modes further enable low power consumption.

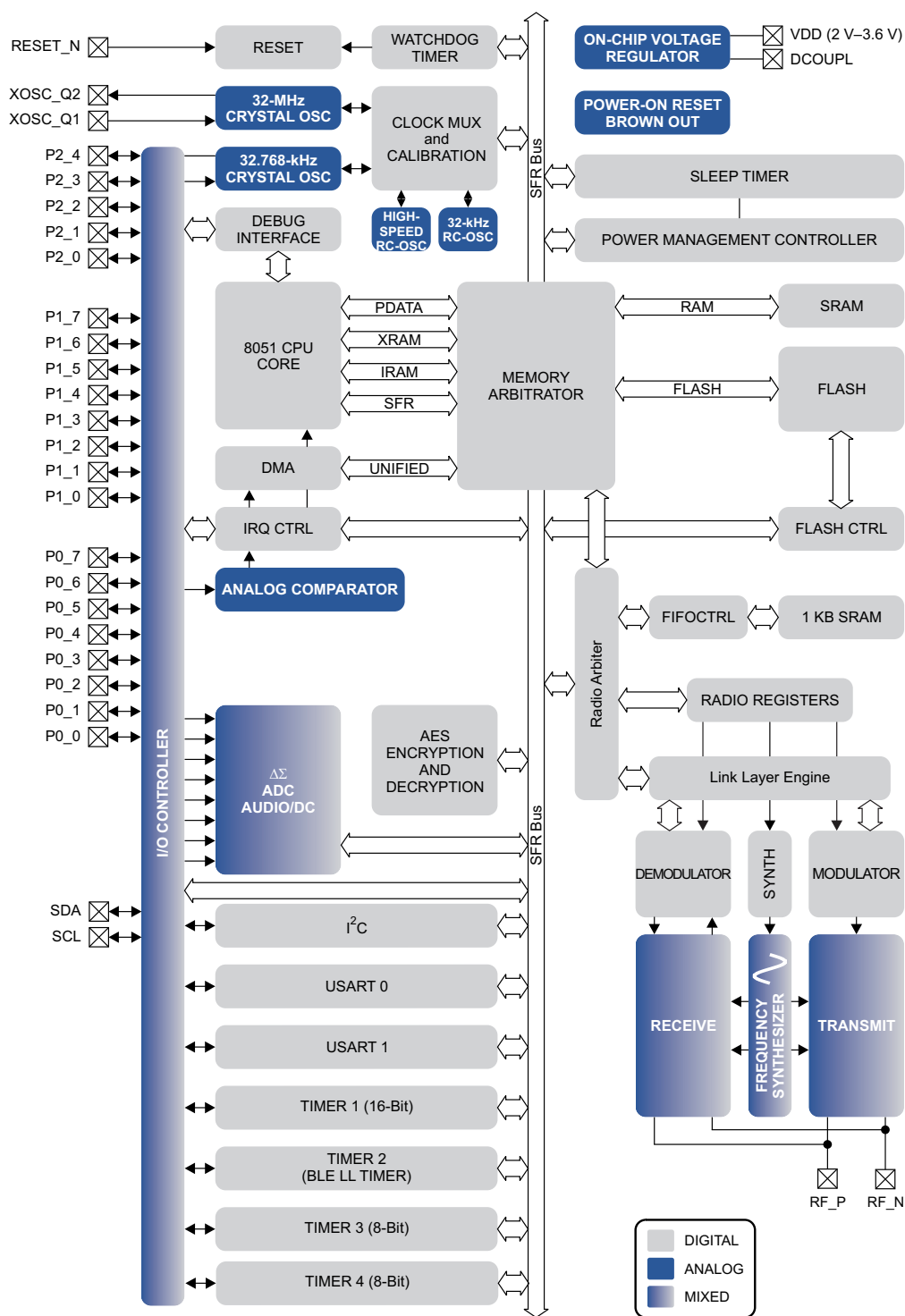The CC2541-Q1 comes in a 6 mm x 6 mm QFN40 package.

Device Information[(1)]

PART NUMBER	PACKAGE	BODY SIZE
CC2541F256TRHARQ1	RHA (40)	6.00 mm × 6.00 mm
CC2541F256TRHATQ1	RHA (40)	6.00 mm × 6.00 mm

(1) For more information, see Section 8, Mechanical Packaging and Orderable Information.

1.4 Functional Block Diagram

Figure 1-1 shows the CC2541-Q1 block diagram.

Figure 1-1. Block Diagram

Table of Contents

Submit Documentation Feedback

Product Folder Links: *CC2541-Q1*

2 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
June 2014	*	Initial release.

3 Terminal Configuration and Functions

The CC2541-Q1 pinout is shown in Figure 3-1 and a short description of the pins follows.

3.1 Pin Diagram
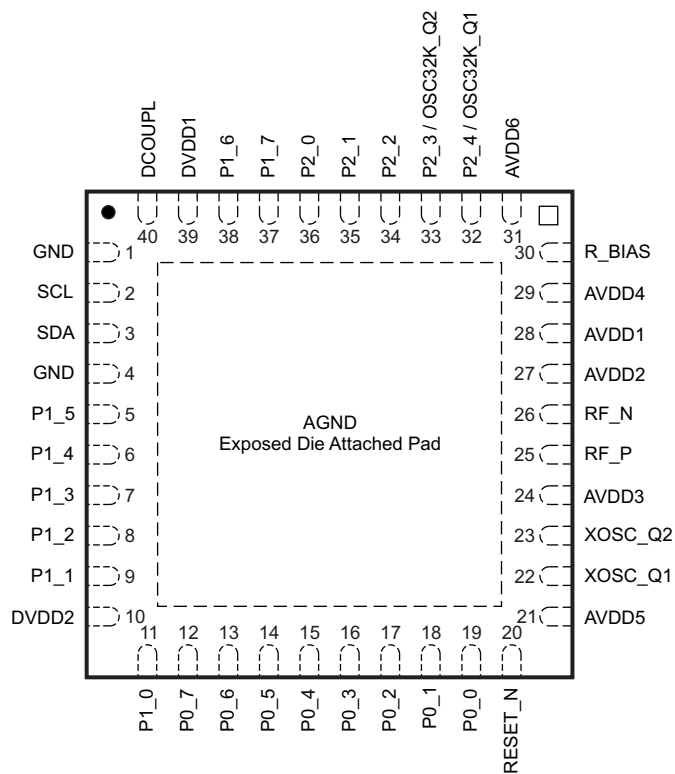
NOTE: The exposed ground pad must be connected to a solid ground plane, as this is the ground connection for the chip.

Figure 3-1. RHA PACKAGE (TOP VIEW)

Submit Documentation Feedback

Product Folder Links: *CC2541-Q1*

3.2 Pin Descriptions

Table 3-1. Pin Descriptions

PINS		TYPE	DESCRIPTION
NAME	NO.		
AVDD1	28	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD2	27	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD3	24	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD4	29	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD5	21	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD6	31	Power (analog)	2-V–3.6-V analog power-supply connection
DCOUPL	40	Power (digital)	1.8-V digital power-supply decoupling. Do not use for supplying external circuits.
DVDD1	39	Power (digital)	2-V–3.6-V digital power-supply connection
DVDD2	10	Power (digital)	2-V–3.6-V digital power-supply connection
GND	1	Ground pin	Connect to GND
GND	—	Ground	The ground pad must be connected to a solid ground plane.
GND	4	Ground pin	Connect to GND
P0_0	19	Digital I/O	Port 0.0
P0_1	18	Digital I/O	Port 0.1
P0_2	17	Digital I/O	Port 0.2
P0_3	16	Digital I/O	Port 0.3
P0_4	15	Digital I/O	Port 0.4
P0_5	14	Digital I/O	Port 0.5
P0_6	13	Digital I/O	Port 0.6
P0_7	12	Digital I/O	Port 0.7
P1_0	11	Digital I/O	Port 1.0 – 20-mA drive capability
P1_1	9	Digital I/O	Port 1.1 – 20-mA drive capability
P1_2	8	Digital I/O	Port 1.2
P1_3	7	Digital I/O	Port 1.3
P1_4	6	Digital I/O	Port 1.4
P1_5	5	Digital I/O	Port 1.5
P1_6	38	Digital I/O	Port 1.6
P1_7	37	Digital I/O	Port 1.7
P2_0	36	Digital I/O	Port 2.0
P2_1/DD	35	Digital I/O	Port 2.1 / debug data
P2_2/DC	34	Digital I/O	Port 2.2 / debug clock
P2_3/ OSC32K_Q2	33	Digital I/O, Analog I/O	Port 2.3/32.768 kHz XOSC
P2_4/ OSC32K_Q1	32	Digital I/O, Analog I/O	Port 2.4/32.768 kHz XOSC
RBIAS	30	Analog I/O	External precision bias resistor for reference current
RESET_N	20	Digital input	Reset, active-low
RF_N	26	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RF_P	25	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
SCL	2	I^{2}C clock or digital I/O	Can be used as I^{2}C clock pin or digital I/O. Leave floating if not used. If grounded disable pull up
SDA	3	I^{2}C clock or digital I/O	Can be used as I^{2}C data pin or digital I/O. Leave floating if not used. If grounded disable pull up
XOSC_Q1	22	Analog I/O	32-MHz crystal oscillator pin 1 or external clock input
XOSC_Q2	23	Analog I/O	32-MHz crystal oscillator pin 2

4 Specifications

4.1 Absolute Maximum Ratings[1]

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Supply voltage	All supply pins must have the same voltage	−0.3	3.9	V
Voltage on any digital pin		−0.3	VDD + 0.3 ≤ 3.9	V
Input RF level			10	dBm

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

4.2 Handling Ratings

				MIN	MAX	UNIT
T_{stg}	Storage temperature range			−40	125	°C
V_{ESD}	Electrostatic discharge (ESD) performance:	Human Body Model (HBM), per AEC Q100-002[1]	All pins	−1	1	kV
			All pins (Excluding pins 25 and 26)	−2	2	
		Charged Device Model (CDM), per AEC Q100-011		−500	500	V

(1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

4.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Operating ambient temperature range, T_A	−40		105	°C
Operating supply voltage	2		3.6	V

4.4 Thermal Characteristics for RHA Package

NAME	DESCRIPTION	°C/W	AIR FLOW (m/s)[1]
$R\Theta_{JC}$	Junction-to-case (top)	16.1	0.00
$R\Theta_{JB}$	Junction-to-board	5.5	0.00
$R\Theta_{JA}$	Junction-to-free air	30.6	0.00
Psi_{JT}	Junction-to-package top	0.2	0.00
Psi_{JB}	Junction-to-board	5.4	0.00
$R\theta_{JC}$	Junction-to-case (bottom)	1.0	0.00

(1) m/s = meters per second

4.5 Electrical Characteristics

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V,
1 Mbps, GFSK, 250-kHz deviation, **Bluetooth** low energy mode, and 0.1% BER

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{core}	Core current consumption	RX mode, standard mode, no peripherals active, low MCU activity		18.3		mA
		RX mode, high-gain mode, no peripherals active, low MCU activity		20.8		
		TX mode, −20 dBm output power, no peripherals active, low MCU activity		17.2		
		TX mode, 0 dBm output power, no peripherals active, low MCU activity		18.6		
		Power mode 1. Digital regulator on; 16-MHz RCOSC and 32-MHz crystal oscillator off; 32.768-kHz XOSC, POR, BOD and sleep timer active; RAM and register retention		270		µA
		Power mode 2. Digital regulator off; 16-MHz RCOSC and 32-MHz crystal oscillator off; 32.768-kHz XOSC, POR, and sleep timer active; RAM and register retention		1		
		Power mode 3. Digital regulator off; no clocks; POR active; RAM and register retention		0.5		
		Low MCU activity: 32-MHz XOSC running. No radio or peripherals. Limited flash access, no RAM access.		6.7		mA
I_{peri}	Peripheral current consumption (Adds to core current I_{core} for each peripheral unit activated)	Timer 1. Timer running, 32-MHz XOSC used		90		µA
		Timer 2. Timer running, 32-MHz XOSC used		90		
		Timer 3. Timer running, 32-MHz XOSC used		60		
		Timer 4. Timer running, 32-MHz XOSC used		70		
		Sleep timer, including 32.753-kHz RCOSC		0.6		
		ADC, when converting		1.2		mA

4.6 General Characteristics

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
WAKE-UP AND TIMING					
Power mode 1 → Active	Digital regulator on, 16-MHz RCOSC and 32-MHz crystal oscillator off. Start-up of 16-MHz RCOSC		4		µs
Power mode 2 or 3 → Active	Digital regulator off, 16-MHz RCOSC and 32-MHz crystal oscillator off. Start-up of regulator and 16-MHz RCOSC		120		µs
Active → TX or RX	Crystal ESR = 16 Ω. Initially running on 16-MHz RCOSC, with 32-MHz XOSC OFF		500		µs
	With 32-MHz XOSC initially on		180		µs
RX/TX turnaround	Proprietary auto mode		130		µs
	BLE mode		150		
RADIO PART					
RF frequency range	Programmable in 1-MHz steps	2379		2496	MHz
Data rate and modulation format	2 Mbps, GFSK, 500-kHz deviation 2 Mbps, GFSK, 320-kHz deviation 1 Mbps, GFSK, 250-kHz deviation 1 Mbps, GFSK, 160-kHz deviation 500 kbps, MSK 250 kbps, GFSK, 160-kHz deviation 250 kbps, MSK				

4.7 RF Receive Section

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C, VDD = 3 V, f_c = 2440 MHz

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
1 Mbps, GFSK, 250-kHz Deviation, *Bluetooth* low energy Mode, 0.1% BER					
Receiver sensitivity[1][2]	High-gain mode		–94		dBm
	Standard mode		–88		
Saturation[2]	BER < 0.1%		5		dBm
Co-channel rejection[2]	Wanted signal –67 dBm		–6		dB
In-band blocking rejection[2]	±1 MHz offset, 0.1% BER, wanted signal –67 dBm		–2		dB
	±2 MHz offset, 0.1% BER, wanted signal –67 dBm		26		
	±3 MHz offset, 0.1% BER, wanted signal –67 dBm		34		
	>6 MHz offset, 0.1% BER, wanted signal –67 dBm		33		
Out-of-band blocking rejection[2]	Minimum interferer level < 2 GHz (Wanted signal –67 dBm)		–21		dBm
	Minimum interferer level [2 GHz, 3 GHz] (Wanted signal –67 dBm)		–27		
	Minimum interferer level > 3 GHz (Wanted signal –67 dBm)		–8		
Intermodulation[2]	Minimum interferer level		–36		dBm
Frequency error tolerance[3]	Including both initial tolerance and drift. Sensitivity better than -67dBm, 250 byte payload. BER 0.1%	–250		250	kHz
Symbol rate error tolerance[4]	Maximum packet length. Sensitivity better than –67 dBm, 250 byte payload. BER 0.1%	–80		80	ppm
ALL RATES/FORMATS					
Spurious emission in RX. Conducted measurement	f < 1 GHz		–67		dBm
Spurious emission in RX. Conducted measurement	f > 1 GHz		–57		dBm

(1) The receiver sensitivity setting is programmable using a TI BLE stack vendor-specific API command. The default value is standard mode.
(2) Results based on standard-gain mode.
(3) Difference between center frequency of the received RF signal and local oscillator frequency
(4) Difference between incoming symbol rate and the internally generated symbol rate

4.8 RF Transmit Section

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C, VDD = 3 V and f_c = 2440 MHz

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output power	Delivered to a single-ended 50-Ω load through a balun using maximum recommended output power setting		0		dBm
	Delivered to a single-ended 50-Ω load through a balun using minimum recommended output power setting		–20		
Programmable output power range	Delivered to a single-ended 50-Ω load through a balun using minimum recommended output power setting		20		dB
Spurious emission conducted measurement	f < 1 GHz			–52	dBm
	f > 1 GHz			–48	dBm
	Suitable for systems targeting compliance with worldwide radio-frequency regulations ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan)				
Optimum load impedance	Differential impedance as seen from the RF port (RF_P and RF_N) toward the antenna		70 +j30		Ω

Designs with antenna connectors that require conducted ETSI compliance at 64 MHz should insert an LC resonator in front of the antenna connector. Use a 1.6-nH inductor in parallel with a 1.8-pF capacitor. Connect both from the signal trace to a good RF ground.

4.9 32-MHz Crystal Oscillator

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Crystal frequency			32		MHz
	Crystal frequency accuracy requirement[1]		–40		40	ppm
ESR	Equivalent series resistance		6		60	Ω
C_0	Crystal shunt capacitance		1		7	pF
C_L	Crystal load capacitance		10		16	pF
	Start-up time			0.25		ms
	Power-down guard time	The crystal oscillator must be in power down for a guard time before it is used again. This requirement is valid for all modes of operation. The need for power-down guard time can vary with crystal type and load.		3		ms

(1) Including aging and temperature dependency, as specified by [1]

4.10   32.768-kHz Crystal Oscillator

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Crystal frequency			32.768		kHz
	Crystal frequency accuracy requirement[1]		–40		40	ppm
ESR	Equivalent series resistance			40	130	kΩ
C_0	Crystal shunt capacitance			0.9	2	pF
C_L	Crystal load capacitance			12	16	pF
	Start-up time			0.4		s

(1)   Including aging and temperature dependency, as specified by [1]

4.11   32-kHz RC Oscillator

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Calibrated frequency[1]			32.753		kHz
Frequency accuracy after calibration			±0.2%		
Temperature coefficient[2]			0.4		%/°C
Supply-voltage coefficient[3]			3		%/V
Calibration time[4]			2		ms

(1)   The calibrated 32-kHz RC oscillator frequency is the 32-MHz XTAL frequency divided by 977.
(2)   Frequency drift when temperature changes after calibration
(3)   Frequency drift when supply voltage changes after calibration
(4)   When the 32-kHz RC oscillator is enabled, it is calibrated when a switch from the 16-MHz RC oscillator to the 32-MHz crystal oscillator
        is performed while SLEEPCMD.OSC32K_CALDIS is set to 0.

4.12   16-MHz RC Oscillator

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency[1]			16		MHz
Uncalibrated frequency accuracy			±18%		
Calibrated frequency accuracy			±0.6%		
Start-up time			10		µs
Initial calibration time[2]			50		µs

(1)   The calibrated 16-MHz RC oscillator frequency is the 32-MHz XTAL frequency divided by 2.
(2)   When the 16-MHz RC oscillator is enabled, it is calibrated when a switch from the 16-MHz RC oscillator to the 32-MHz crystal oscillator
        is performed while SLEEPCMD.OSC_PD is set to 0.

4.13 RSSI Characteristics

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
2 Mbps, GFSK, 320-kHz Deviation, 0.1% BER and 2 Mbps, GFSK, 500-kHz Deviation, 0.1% BER					
Useful RSSI range[1]	Reduced gain by AGC algorithm		64		dB
	High gain by AGC algorithm		64		
RSSI offset[1]	Reduced gain by AGC algorithm		79		dBm
	High gain by AGC algorithm		99		
Absolute uncalibrated accuracy[1]			±6		dB
Step size (LSB value)			1		dB
All Other Rates/Formats					
Useful RSSI range[1]	Standard mode		64		dB
	High-gain mode		64		
RSSI offset[1]	Standard mode		98		dBm
	High-gain mode		107		
Absolute uncalibrated accuracy[1]			±3		dB
Step size (LSB value)			1		dB

(1) Assuming CC2541-Q1 EM reference design. Other RF designs give an offset from the reported value.

4.14 Frequency Synthesizer Characteristics

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C, VDD = 3 V and f_c = 2440 MHz

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise, unmodulated carrier	At ±1-MHz offset from carrier		−109		dBc/Hz
	At ±3-MHz offset from carrier		−112		
	At ±5-MHz offset from carrier		−119		

4.15 Analog Temperature Sensor

Measured on Texas Instruments CC2541-Q1 EM reference design with T_A = 25°C and VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output			1480		12-bit
Temperature coefficient			4.5		/ 1°C
Voltage coefficient	Measured using integrated ADC, internal band-gap voltage reference, and maximum resolution		1		0.1 V
Initial accuracy without calibration			±10		°C
Accuracy using 1-point calibration			±5		°C
Current consumption when enabled			0.5		mA

4.16 Comparator Characteristics

T_A = 25°C, VDD = 3 V. All measurement results are obtained using the CC2541-Q1 reference designs, post-calibration.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Common-mode maximum voltage			VDD		V
Common-mode minimum voltage			−0.3		
Input offset voltage			1		mV
Offset vs temperature			16		µV/°C
Offset vs operating voltage			4		mV/V
Supply current			230		nA
Hysteresis			0.15		mV

4.17 ADC Characteristics

$T_A = 25°C$ and VDD = 3 V

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage	VDD is voltage on AVDD5 pin	0		VDD	V
	External reference voltage	VDD is voltage on AVDD5 pin	0		VDD	V
	External reference voltage differential	VDD is voltage on AVDD5 pin	0		VDD	V
	Input resistance, signal	Simulated using 4-MHz clock speed		197		kΩ
	Full-scale signal[1]	Peak-to-peak, defines 0 dBFS		2.97		V
ENOB[1]	Effective number of bits	Single-ended input, 7-bit setting		5.7		bits
		Single-ended input, 9-bit setting		7.5		
		Single-ended input, 10-bit setting		9.3		
		Single-ended input, 12-bit setting		10.3		
		Differential input, 7-bit setting		6.5		
		Differential input, 9-bit setting		8.3		
		Differential input, 10-bit setting		10		
		Differential input, 12-bit setting		11.5		
		10-bit setting, clocked by RCOSC		9.7		
		12-bit setting, clocked by RCOSC		10.9		
	Useful power bandwidth	7-bit setting, both single and differential		0–20		kHz
THD	Total harmonic distortion	Single ended input, 12-bit setting, –6 dBFS[1]		–75.2		dB
		Differential input, 12-bit setting, –6 dBFS[1]		–86.6		
	Signal to nonharmonic ratio	Single-ended input, 12-bit setting[1]		70.2		dB
		Differential input, 12-bit setting[1]		79.3		
		Single-ended input, 12-bit setting, –6 dBFS[1]		78.8		
		Differential input, 12-bit setting, –6 dBFS[1]		88.9		
CMRR	Common-mode rejection ratio	Differential input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
	Crosstalk	Single ended input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
	Offset	Midscale		–3		mV
	Gain error			0.68%		
DNL	Differential nonlinearity	12-bit setting, mean[1]		0.05		LSB
		12-bit setting, maximum[1]		0.9		
INL	Integral nonlinearity	12-bit setting, mean[1]		4.6		LSB
		12-bit setting, maximum[1]		13.3		
		12-bit setting, mean, clocked by RCOSC		10		
		12-bit setting, max, clocked by RCOSC		29		
SINAD (–THD+N)	Signal-to-noise-and-distortion	Single ended input, 7-bit setting[1]		35.4		dB
		Single ended input, 9-bit setting[1]		46.8		
		Single ended input, 10-bit setting[1]		57.5		
		Single ended input, 12-bit setting[1]		66.6		
		Differential input, 7-bit setting[1]		40.7		
		Differential input, 9-bit setting[1]		51.6		
		Differential input, 10-bit setting[1]		61.8		
		Differential input, 12-bit setting[1]		70.8		
	Conversion time	7-bit setting		20		µs
		9-bit setting		36		
		10-bit setting		68		
		12-bit setting		132		

(1) Measured with 300-Hz sine-wave input and VDD as reference.

ADC Characteristics *(continued)*

T_A = 25°C and VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Power consumption			1.2		mA
Internal reference VDD coefficient			4		mV/V
Internal reference temperature coefficient			0.4		mV/10°C
Internal reference voltage			1.24		V

4.18 DC Characteristics

T_A = 25°C, VDD = 3 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Logic-0 input voltage				0.5	V
Logic-1 input voltage		2.4			V
Logic-0 input current	Input equals 0 V	−50		50	nA
Logic-1 input current	Input equals VDD	−50		50	nA
I/O-pin pullup and pulldown resistors			20		kΩ
Logic-0 output voltage, 4- mA pins	Output load 4 mA			0.5	V
Logic-1 output voltage, 4-mA pins	Output load 4 mA	2.5			V
Logic-0 output voltage, 20- mA pins	Output load 20 mA			0.5	V
Logic-1 output voltage, 20-mA pins	Output load 20 mA	2.5			V

4.19 Control Input AC Characteristics

T_A = −40°C to 105°C, VDD = 2 V to 3.6 V

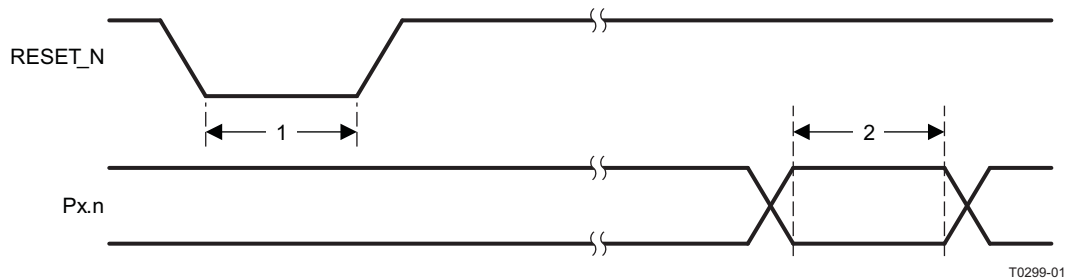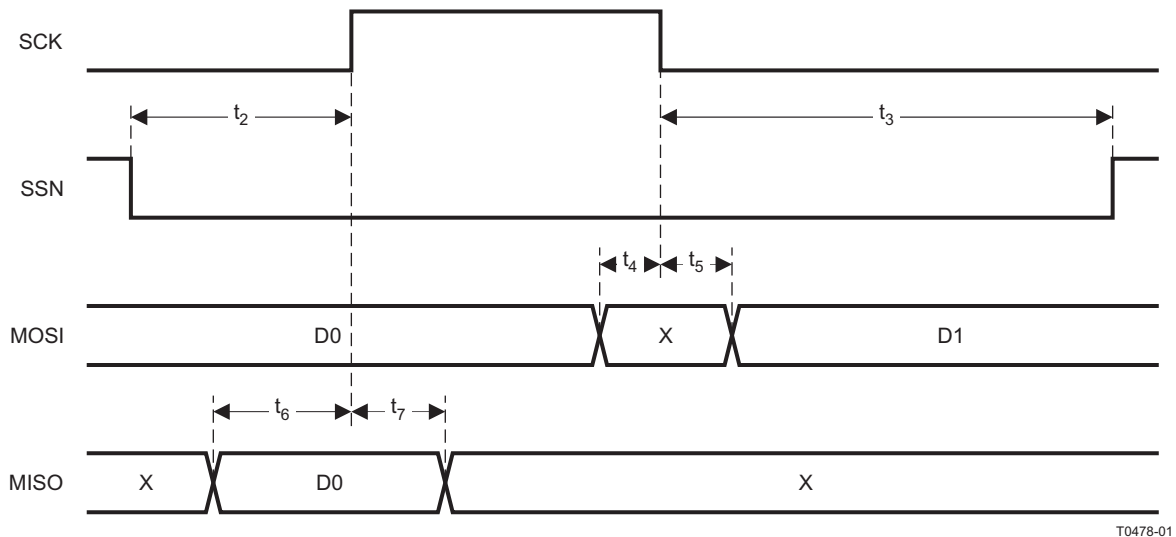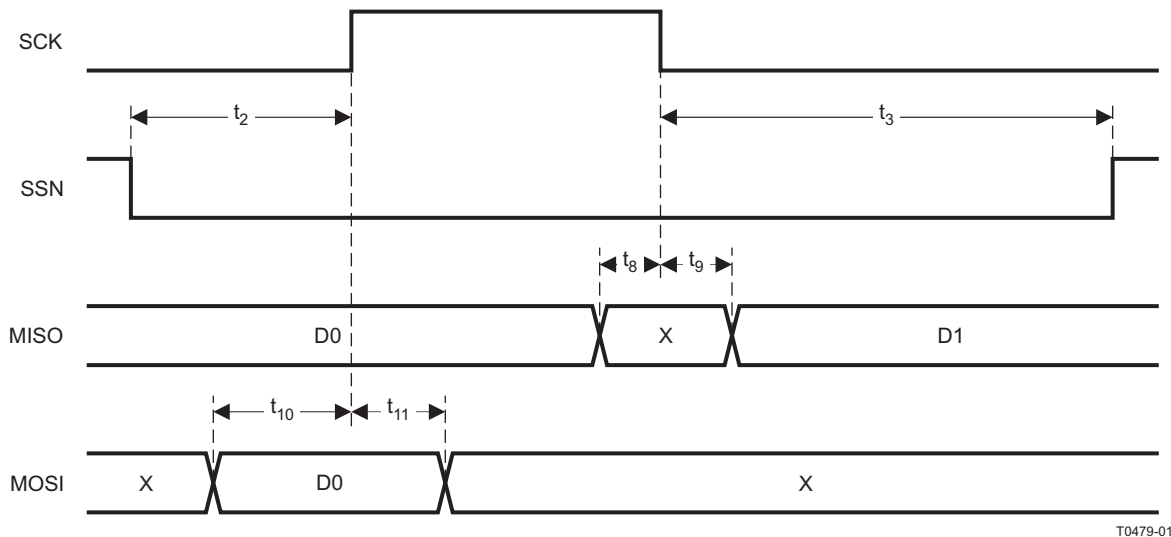| PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
System clock, f_{SYSCLK} t_{SYSCLK} = 1/ f_{SYSCLK}	The undivided system clock is 32 MHz when crystal oscillator is used. The undivided system clock is 16 MHz when calibrated 16-MHz RC oscillator is used.	16		32	MHz
RESET_N low duration	See item 1, Figure 4-1. This is the shortest pulse that is recognized as a complete reset pin request. Note that shorter pulses may be recognized but do not lead to complete reset of all modules within the chip.	1			µs
Interrupt pulse duration	See item 2, Figure 4-1.This is the shortest pulse that is recognized as an interrupt request.	20			ns

Figure 4-1. Control Input AC Characteristics

Submit Documentation Feedback
Product Folder Links: *CC2541-Q1*

4.20 SPI AC Characteristics

T_A = −40°C to 105°C, VDD = 2 V to 3.6 V

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_1	SCK period	Master, RX and TX	250			ns
		Slave, RX and TX	250			
	SCK duty cycle	Master		50%		
t_2	SSN low to SCK	Master	63			ns
		Slave	63			
t_3	SCK to SSN high	Master	63			ns
		Slave	63			
t_4	MOSI early out	Master, load = 10 pF			7	ns
t_5	MOSI late out	Master, load = 10 pF			10	ns
t_6	MISO setup	Master	90			ns
t_7	MISO hold	Master	10			ns
	SCK duty cycle	Slave		50%		ns
t_{10}	MOSI setup	Slave	35			ns
t_{11}	MOSI hold	Slave	10			ns
t_9	MISO late out	Slave, load = 10 pF			95	ns
	Operating frequency	Master, TX only			8	MHz
		Master, RX and TX			4	
		Slave, RX only			8	
		Slave, RX and TX			4	

Figure 4-2. SPI Master AC Characteristics

Submit Documentation Feedback

Product Folder Links: CC2541-Q1

Figure 4-3. SPI Slave AC Characteristics

4.21 Debug Interface AC Characteristics
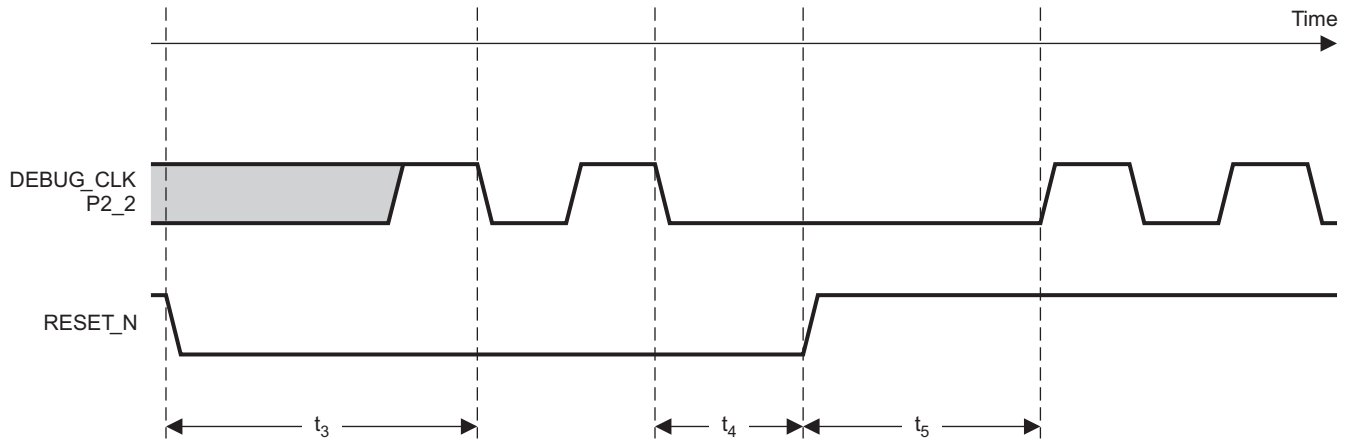
T_A = –40°C to 105°C, VDD = 2 V to 3.6 V

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f_{clk_dbg}	Debug clock frequency (see Figure 4-4)				12	MHz
t_1	Allowed high pulse on clock (see Figure 4-4)		35			ns
t_2	Allowed low pulse on clock (see Figure 4-4)		35			ns
t_3	EXT_RESET_N low to first falling edge on debug clock (see Figure 4-6)		167			ns
t_4	Falling edge on clock to EXT_RESET_N high (see Figure 4-6)		83			ns
t_5	EXT_RESET_N high to first debug command (see Figure 4-6)		83			ns
t_6	Debug data setup (see Figure 4-5)		2			ns
t_7	Debug data hold (see Figure 4-5)		4			ns
t_8	Clock-to-data delay (see Figure 4-5)	Load = 10 pF			30	ns
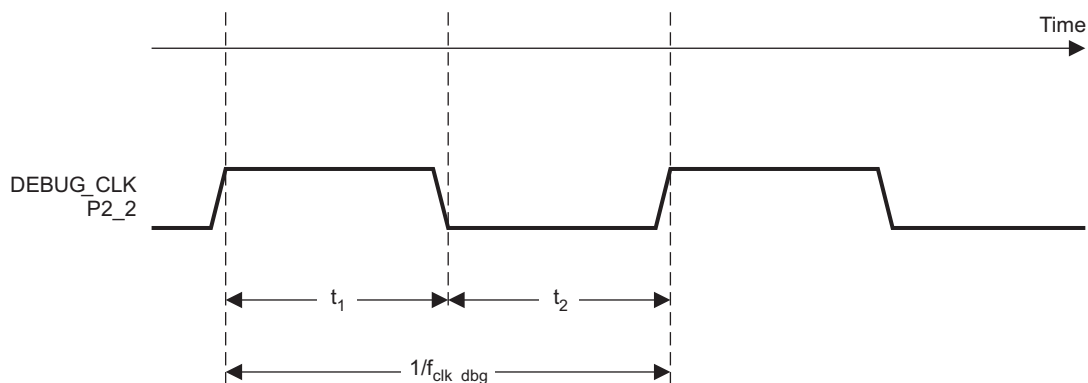
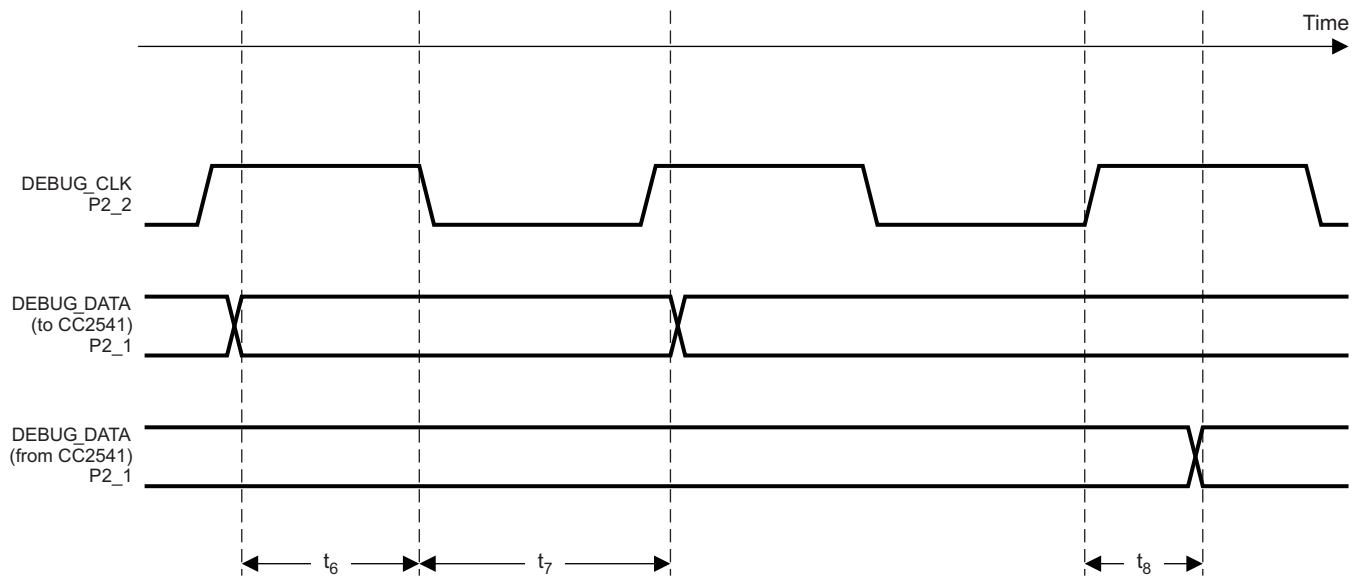**Figure 4-4. Debug Clock – Basic Timing**

Figure 4-5. Debug Enable Timing

Figure 4-6. Data Setup and Hold Timing

4.22 Timer Inputs AC Characteristics

T_A = –40°C to 105°C, VDD = 2 V to 3.6 V

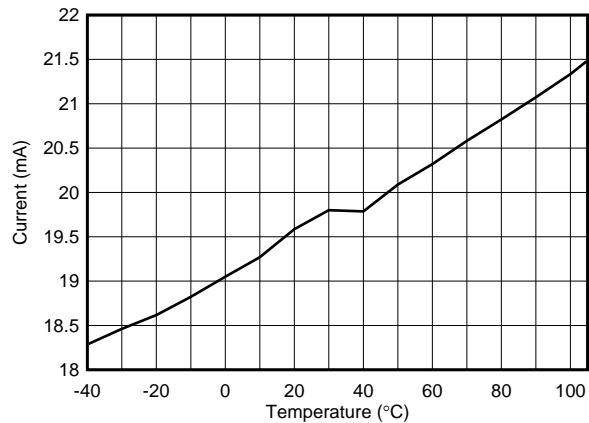| PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input capture pulse duration	Synchronizers determine the shortest input pulse that can be recognized. The synchronizers operate at the current system clock rate (16 MHz or 32 MHz).	1.5			t_{SYSCLK}

4.23 Typical Characteristics

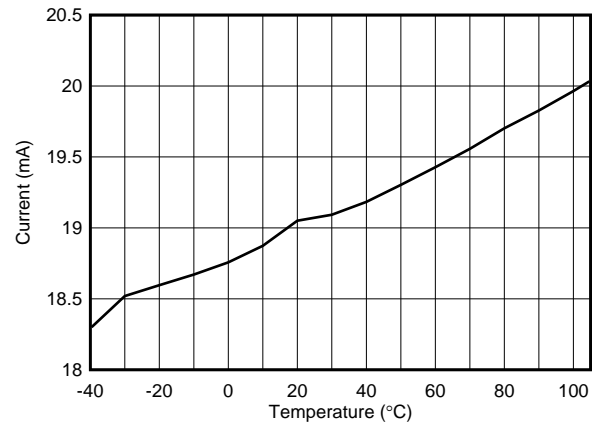**Figure 4-7. RX Current vs Temperature**

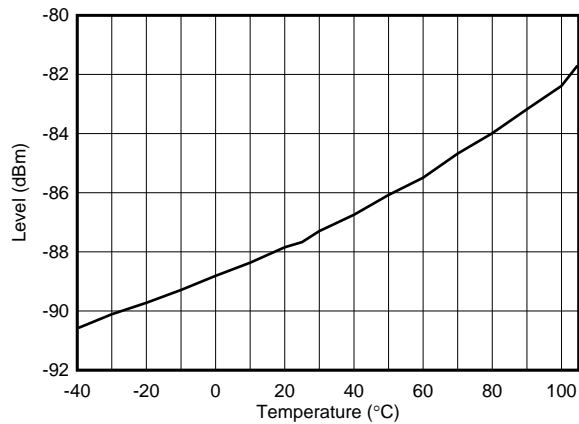**Figure 4-8. TX Current vs Temperature**

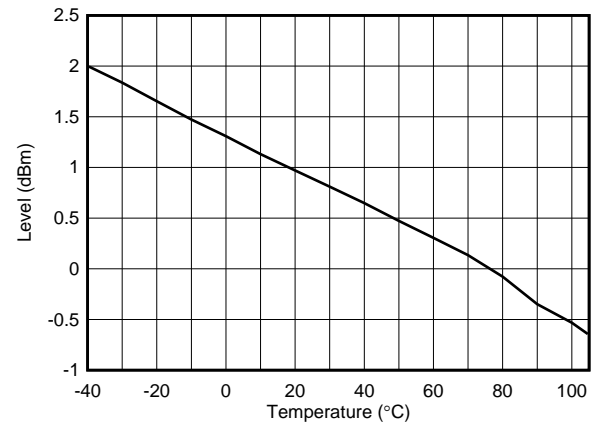**Figure 4-9. RX Sensitivity vs Temperature**

Figure 4-10. TX Power vs Temperature

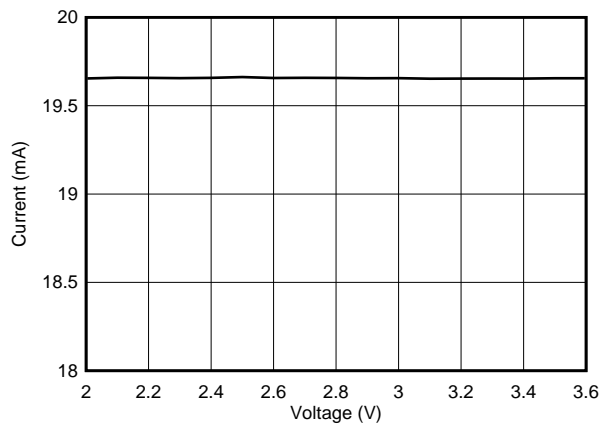**Figure 4-11. RX Current vs Supply Voltage**

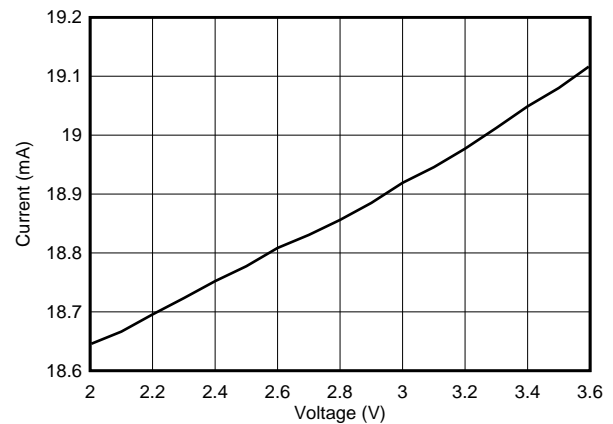**Figure 4-12. TX Current vs Supply Voltage**

Typical Characteristics *(continued)*

1-Mbps GFSK 250-kHz Standard Gain Setting

$T_A = 25°C$
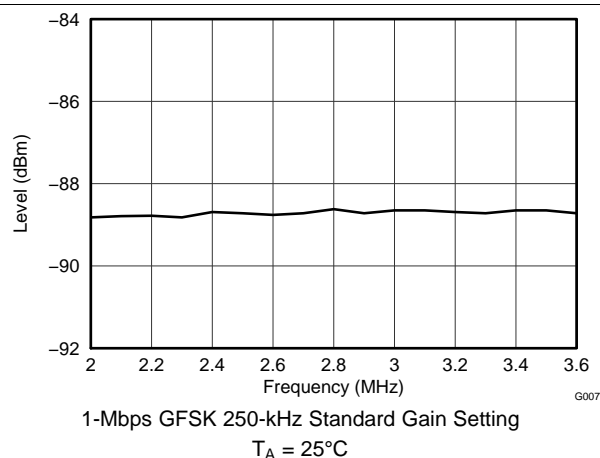
Figure 4-13. RX Sensitivity vs Supply Voltage

TX Power Setting = 0 dBm

$T_A = 25°C$
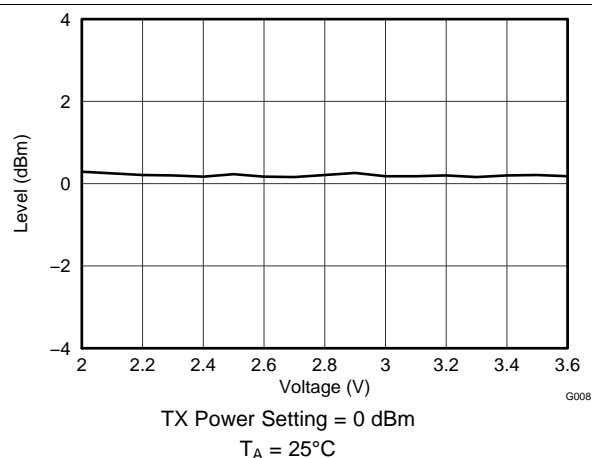
Figure 4-14. TX Power vs Supply Voltage

1-Mbps GFSK 250-kHz Standard Gain Setting

$T_A = 25°C$

$V_{cc} = 3$ V
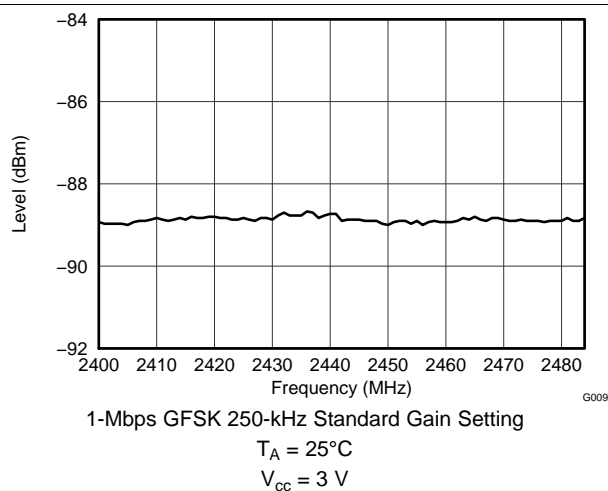
Figure 4-15. RX Sensitivity vs Frequency

TX Power Setting = 0 dBm

$T_A = 25°C$

$V_{cc} = 3$ V
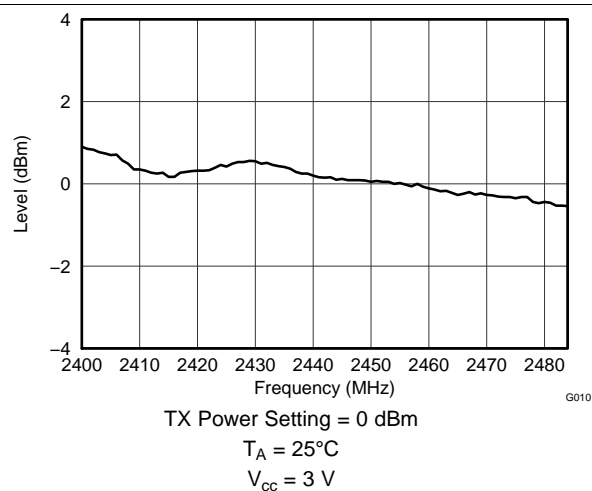
Figure 4-16. TX Power vs Frequency

Table 4-1. Output Power[1][2]

TX POWER Setting	Typical Output Power (dBm)
0xE1	0
0xD1	−2
0xC1	−4
0xB1	−6
0xA1	−8
0x91	−10
0x81	−12
0x71	−14
0x61	−16
0x51	−18
0x41	−20

(1) Measured on Texas Instruments CC2541-Q1 EM reference design with $T_A = 25°C$, VDD = 3 V and f_c = 2440 MHz. See SWRU191 for recommended register settings.
(2) 1 Mbsp, GFSK, 250-kHz deviation, *Bluetooth* low energy mode, 1% BER

5 Detailed Description

5.1 Functional Block Diagram

A block diagram of the CC2541-Q1 is shown in Figure 5-1. The modules can be roughly divided into one of three categories: CPU-related modules; modules related to power, test, and clock distribution; and radio-related modules. In the following subsections, a short description of each module is given.

Figure 5-1. CC2541-Q1 Block Diagram

5.2 Block Descriptions

A block diagram of the CC2541-Q1 is shown in Figure 5-1. The modules can be roughly divided into one of three categories: CPU-related modules; modules related to power, test, and clock distribution; and radio-related modules. In the following subsections, a short description of each module is given.

5.2.1 CPU and Memory

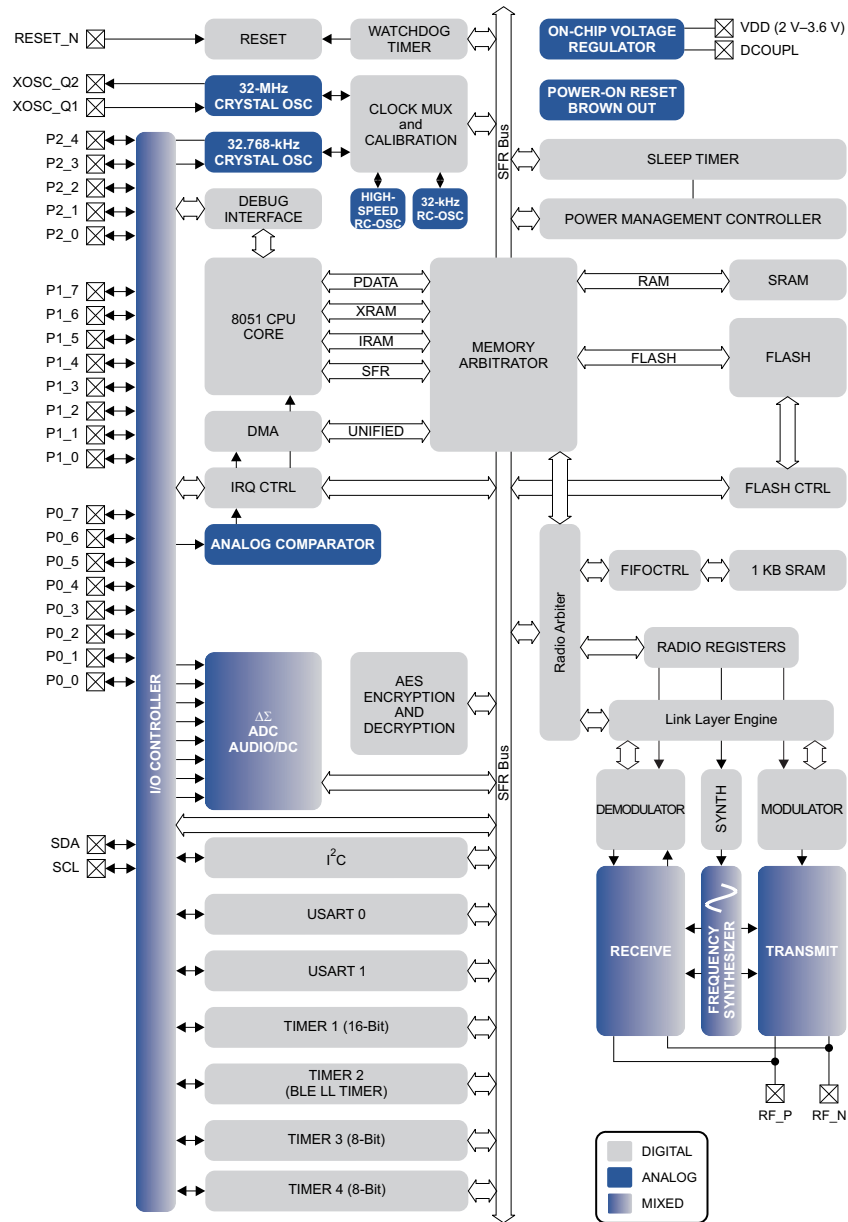The **8051 CPU core** is a single-cycle 8051-compatible core. It has three different memory access busses (SFR, DATA, and CODE/XDATA), a debug interface, and an 18-input extended interrupt unit.

The **memory arbiter** is at the heart of the system, as it connects the CPU and DMA controller with the physical memories and all peripherals through the SFR bus. The memory arbiter has four memory-access points, access of which can map to one of three physical memories: an SRAM, flash memory, and XREG/SFR registers. It is responsible for performing arbitration and sequencing between simultaneous memory accesses to the same physical memory.

The **SFR bus** is drawn conceptually in Figure 5-1 as a common bus that connects all hardware peripherals to the memory arbiter. The SFR bus in the block diagram also provides access to the radio registers in the radio register bank, even though these are indeed mapped into XDATA memory space.

The **8-KB SRAM** maps to the DATA memory space and to parts of the XDATA memory spaces. The SRAM is an ultralow-power SRAM that retains its contents even when the digital part is powered off (power mode 2 and mode 3).

The **256 KB flash block** provides in-circuit programmable non-volatile program memory for the device, and maps into the CODE and XDATA memory spaces.

5.2.2 Peripherals

Writing to the flash block is performed through a **flash controller** that allows page-wise erasure and 4-bytewise programming. See User Guide for details on the flash controller.

A versatile five-channel **DMA controller** is available in the system, accesses memory using the XDATA memory space, and thus has access to all physical memories. Each channel (trigger, priority, transfer mode, addressing mode, source and destination pointers, and transfer count) is configured with DMA descriptors that can be located anywhere in memory. Many of the hardware peripherals (AES core, flash controller, USARTs, timers, ADC interface, etc.) can be used with the DMA controller for efficient operation by performing data transfers between a single SFR or XREG address and flash/SRAM.

Each CC2541-Q1 contains a unique 48-bit IEEE address that can be used as the public device address for a *Bluetooth* device. Designers are free to use this address, or provide their own, as described in the *Bluetooth* specfication.

The **interrupt controller** services a total of 18 interrupt sources, divided into six interrupt groups, each of which is associated with one of four interrupt priorities. I/O and sleep timer interrupt requests are serviced even if the device is in a sleep mode (power modes 1 and 2) by bringing the CC2541-Q1 back to the active mode.

The **debug interface** implements a proprietary two-wire serial interface that is used for in-circuit debugging. Through this debug interface, it is possible to erase or program the entire flash memory, control which oscillators are enabled, stop and start execution of the user program, execute instructions on the 8051 core, set code breakpoints, and single-step through instructions in the code. Using these techniques, it is possible to perform in-circuit debugging and external flash programming elegantly.

The **I/O controller** is responsible for all general-purpose I/O pins. The CPU can configure whether peripheral modules control certain pins or whether they are under software control, and if so, whether each pin is configured as an input or output and if a pullup or pulldown resistor in the pad is connected. Each peripheral that connects to the I/O pins can choose between two different I/O pin locations to ensure flexibility in various applications.

The **sleep timer** is an ultralow-power timer that can either use an external 32.768-kHz crystal oscillator or an internal 32.753-kHz RC oscillator. The sleep timer runs continuously in all operating modes except power mode 3. Typical applications of this timer are as a real-time counter or as a wake-up timer to get out of power mode 1 or mode 2.

A built-in **watchdog timer** allows the CC2541-Q1 to reset itself if the firmware hangs. When enabled by software, the watchdog timer must be cleared periodically; otherwise, it resets the device when it times out.

Timer 1 is a 16-bit timer with timer/counter/PWM functionality. It has a programmable prescaler, a 16-bit period value, and five individually programmable counter/capture channels, each with a 16-bit compare value. Each of the counter/capture channels can be used as a PWM output or to capture the timing of edges on input signals. It can also be configured in IR generation mode, where it counts timer 3 periods and the output is ANDed with the output of timer 3 to generate modulated consumer IR signals with minimal CPU interaction.

Timer 2 is a 40-bit timer. It has a 16-bit counter with a configurable timer period and a 24-bit overflow counter that can be used to keep track of the number of periods that have transpired. A 40-bit capture register is also used to record the exact time at which a start-of-frame delimiter is received/transmitted or the exact time at which transmission ends. There are two 16-bit output compare registers and two 24-bit overflow compare registers that can be used to give exact timing for start of RX or TX to the radio or general interrupts.

Timer 3 and timer 4 are 8-bit timers with timer/counter/PWM functionality. They have a programmable prescaler, an 8-bit period value, and one programmable counter channel with an 8-bit compare value. Each of the counter channels can be used as PWM output.

USART 0 and USART 1 are each configurable as either an SPI master/slave or a UART. They provide double buffering on both RX and TX and hardware flow control and are thus well suited to high-throughput full-duplex applications. Each USART has its own high-precision baud-rate generator, thus leaving the ordinary timers free for other uses. When configured as SPI slaves, the USARTs sample the input signal using SCK directly instead of using some oversampling scheme, and are thus well-suited for high data rates.

The **AES encryption/decryption core** allows the user to encrypt and decrypt data using the AES algorithm with 128-bit keys. The AES core also supports ECB, CBC, CFB, OFB, CTR, and CBC-MAC, as well as hardware support for CCM.

The **ADC** supports 7 to 12 bits of resolution with a corresponding range of bandwidths from 30-kHz to 4-kHz, respectively. DC and audio conversions with up to eight input channels (I/O controller pins) are possible. The inputs can be selected as single-ended or differential. The reference voltage can be internal, AVDD, or a single-ended or differential external signal. The ADC also has a temperature-sensor input channel. The ADC can automate the process of periodic sampling or conversion over a sequence of channels.

The **I²C** module provides a digital peripheral connection with two pins and supports both master and slave operation. I²C support is compliant with the NXP I²C specification version 2.1 and supports standard mode (up to 100 kbps) and fast mode (up to 400 kbps). In addition, 7-bit device addressing modes are supported, as well as master and slave modes.

The ultralow-power **analog comparator** enables applications to wake up from PM2 or PM3 based on an analog signal. Both inputs are brought out to pins; the reference voltage must be provided externally. The comparator output is connected to the I/O controller interrupt detector and can be treated by the MCU as a regular I/O pin interrupt.

6 Application Information

Few external components are required for the operation of the CC2541-Q1. A typical application circuit is shown in Figure 6-1.
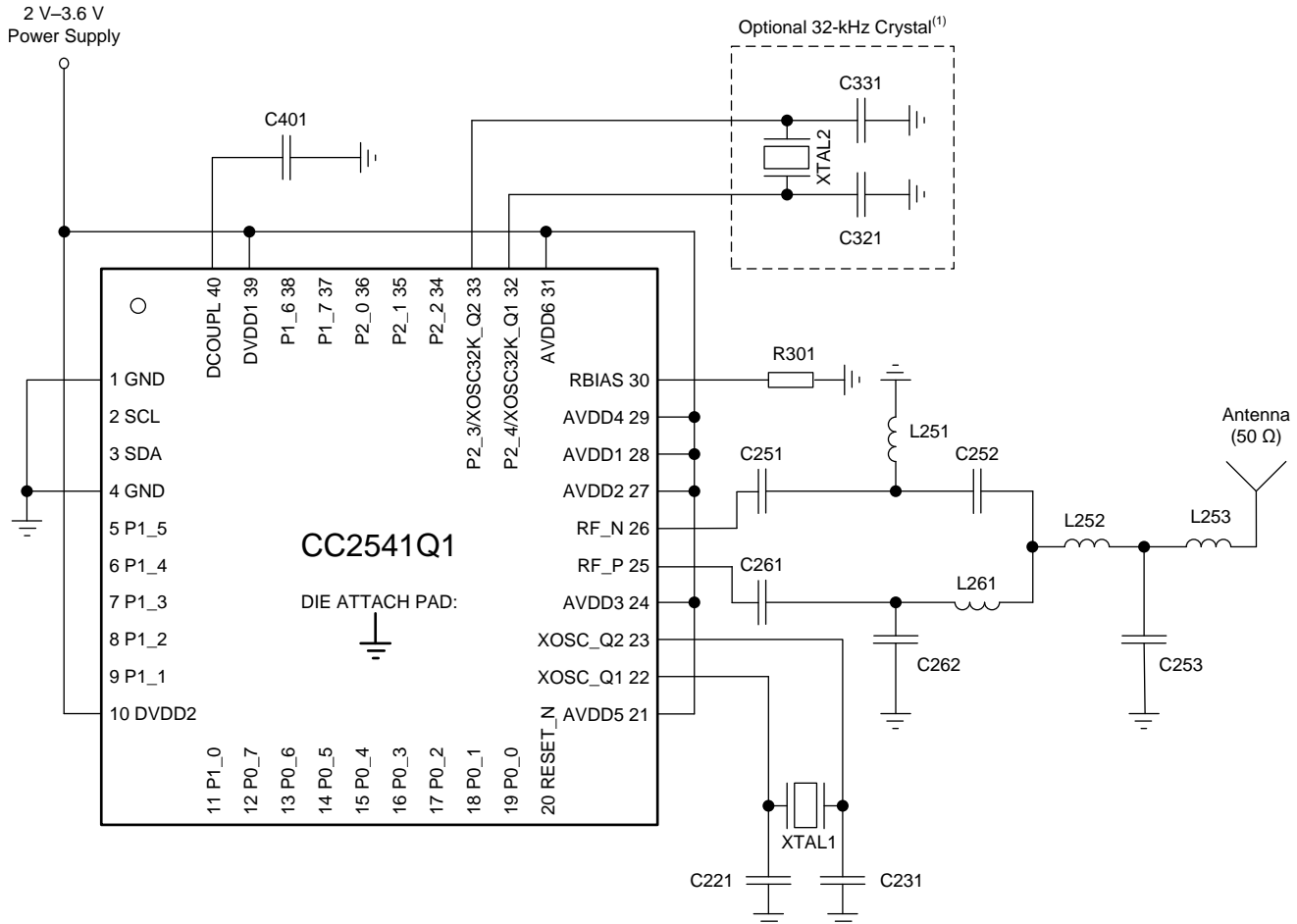
(1) 32-kHz crystal is mandatory when running the BLE protocol stack in low-power modes, except if the link layer is in the standby state (Vol. 6 Part B Section 1.1 in [1]).
NOTE: Different antenna alternatives will be provided as reference designs.
Power supply decoupling capacitors are not shown. Digital I/O not connected

Figure 6-1. CC2541-Q1 Application Circuit

Table 6-1. Overview of External Components (Excluding Supply Decoupling Capacitors)

Component	Description	Value
C401	Decoupling capacitor for the internal 1.8-V digital voltage regulator	1 µF
R301	Precision resistor ±1%, used for internal biasing	56 kΩ

6.1 Input/Output Matching

When using an unbalanced antenna such as a monopole, a balun should be used to optimize performance. The balun can be implemented using low-cost discrete inductors and capacitors. See reference design, CC2541-Q1EM, for recommended balun.

6.2 Crystal

An external 32-MHz crystal, XTAL1, with two loading capacitors (C221 and C231) is used for the 32-MHz crystal oscillator. See Section 4.9 for details. The load capacitance seen by the 32-MHz crystal is given by:

$$C_L = \frac{1}{\dfrac{1}{C_{221}} + \dfrac{1}{C_{231}}} + C_{parasitic}$$

(1)

XTAL2 is an optional 32.768-kHz crystal, with two loading capacitors (C321 and C331) used for the 32.768-kHz crystal oscillator. The 32.768-kHz crystal oscillator is used in applications where both very low sleep-current consumption and accurate wake-up times are needed. The load capacitance seen by the 32.768-kHz crystal is given by:

$$C_L = \frac{1}{\dfrac{1}{C_{321}} + \dfrac{1}{C_{331}}} + C_{parasitic}$$

(2)

A series resistor may be used to comply with the ESR requirement.

6.3 On-Chip 1.8-V Voltage Regulator Decoupling

The 1.8-V on-chip voltage regulator supplies the 1.8-V digital logic. This regulator requires a decoupling capacitor (C471) for stable operation.

6.4 Power-Supply Decoupling and Filtering

Proper power-supply decoupling must be used for optimum performance. The placement and size of the decoupling capacitors and the power supply filtering are very important to achieve the best performance in an application. TI provides a compact reference design that should be followed very closely.

7 Device and Documentation Support

7.1 Documentation Support

7.1.1 Related Documentation

1. *Bluetooth*® Core Technical Specification document, version 4.0
   http://www.bluetooth.com/SiteCollectionDocuments/Core_V40.zip
2. CC253x System-on-Chip Solution for 2.4-GHz IEEE 802.15.4 and ZigBee® Applications/CC2541-Q1 System-on-Chip Solution for 2.4-GHz *Bluetooth* low energy Applications (SWRU191)
3. Current Savings in CC254x Using the TPS62730 (SWRA365).

7.1.1.1 Additional Information

Texas Instruments offers a wide selection of cost-effective, low-power RF solutions for proprietary and standard-based wireless applications for use in industrial and consumer applications. Our selection includes RF transceivers, RF transmitters, RF front ends, and System-on-Chips as well as various software solutions for the sub-1- and 2.4-GHz frequency bands.

In addition, Texas Instruments provides a large selection of support collateral such as development tools, technical documentation, reference designs, application expertise, customer support, third-party and university programs.

The Low-Power RF E2E Online Community provides technical support forums, videos and blogs, and the chance to interact with fellow engineers from all over the world.

With a broad selection of product solutions, end application possibilities, and a range of technical support, Texas Instruments offers the broadest low-power RF portfolio. We make RF easy!

For more information on low-power RF, see Section 7.1.1.2, Section 7.1.1.3, and Section 7.1.1.4.

7.1.1.2 Texas Instruments Low-Power RF Website

- Forums, videos, and blogs
- RF design help
- E2E interaction

Join us today at www.ti.com/lprf-forum.

7.1.1.3 Texas Instruments Low-Power RF Developer Network

Texas Instruments has launched an extensive network of low-power RF development partners to help customers speed up their application development. The network consists of recommended companies, RF consultants, and independent design houses that provide a series of hardware module products and design services, including:

- RF circuit, low-power RF, and ZigBee® design services
- Low-power RF and ZigBee module solutions and development tools
- RF certification services and RF circuit manufacturing

Need help with modules, engineering services or development tools?

Search the Low-Power RF Developer Network tool to find a suitable partner.
www.ti.com/lprfnetwork

7.1.1.4 Low-Power RF eNewsletter

The Low-Power RF eNewsletter keeps you up-to-date on new products, news releases, developers' news, and other news and events associated with low-power RF products from TI. The Low-Power RF eNewsletter articles include links to get more online information.

Sign up today on
www.ti.com/lprfnewsletter

7.2 Trademarks

SimpleLink is a trademark of Texas Instruments.
Bluetooth is a registered trademark of Bluetooth SIG, Inc..
ZigBee is a registered trademark of ZigBee Alliance.
All other trademarks are the property of their respective owners.

7.3 Electrostatic Discharge Caution

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

7.4 Glossary

SLYZ022 — *TI Glossary*.
This glossary lists and explains terms, acronyms, and definitions.

8 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
CC2541F256TRHARQ1	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 105	CC2541Q1 F256	Samples
CC2541F256TRHATQ1	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 105	CC2541Q1 F256	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

![Texas Instruments logo]

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF CC2541-Q1 :

● Catalog: CC2541

NOTE: Qualified Version Definitions:

     ● Catalog - TI's standard catalog product

PACKAGE MATERIALS INFORMATION

TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width	
B0	Dimension designed to accommodate the component length	
K0	Dimension designed to accommodate the component thickness	
W	Overall width of the carrier tape	
P1	Pitch between successive cavity centers	

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC2541F256TRHARQ1	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2541F256TRHATQ1	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC2541F256TRHARQ1	VQFN	RHA	40	2500	336.6	336.6	28.6
CC2541F256TRHATQ1	VQFN	RHA	40	250	213.0	191.0	55.0

RHA (S−PVQFN−N40)   PLASTIC QUAD FLATPACK NO−LEAD

6,15
5,85   [A]

[B]

6,15
5,85

Pin 1 Index Area
Top and Bottom

40

1

1,00
0,80

0,20 Nominal
Lead Frame

◁ 0,08 C

Seating Plane

0,05
0,00

[C]

40X 0,50
     0,30

1   10

40

THERMAL PAD

11

SIZE AND SHAPE
SHOWN ON SEPARATE SHEET

31

20

30   21

0,50

40X 0,23 +0,07
        −0,05

⊕ 0,10 Ⓜ C A B
   0,05 Ⓜ C

Bottom View
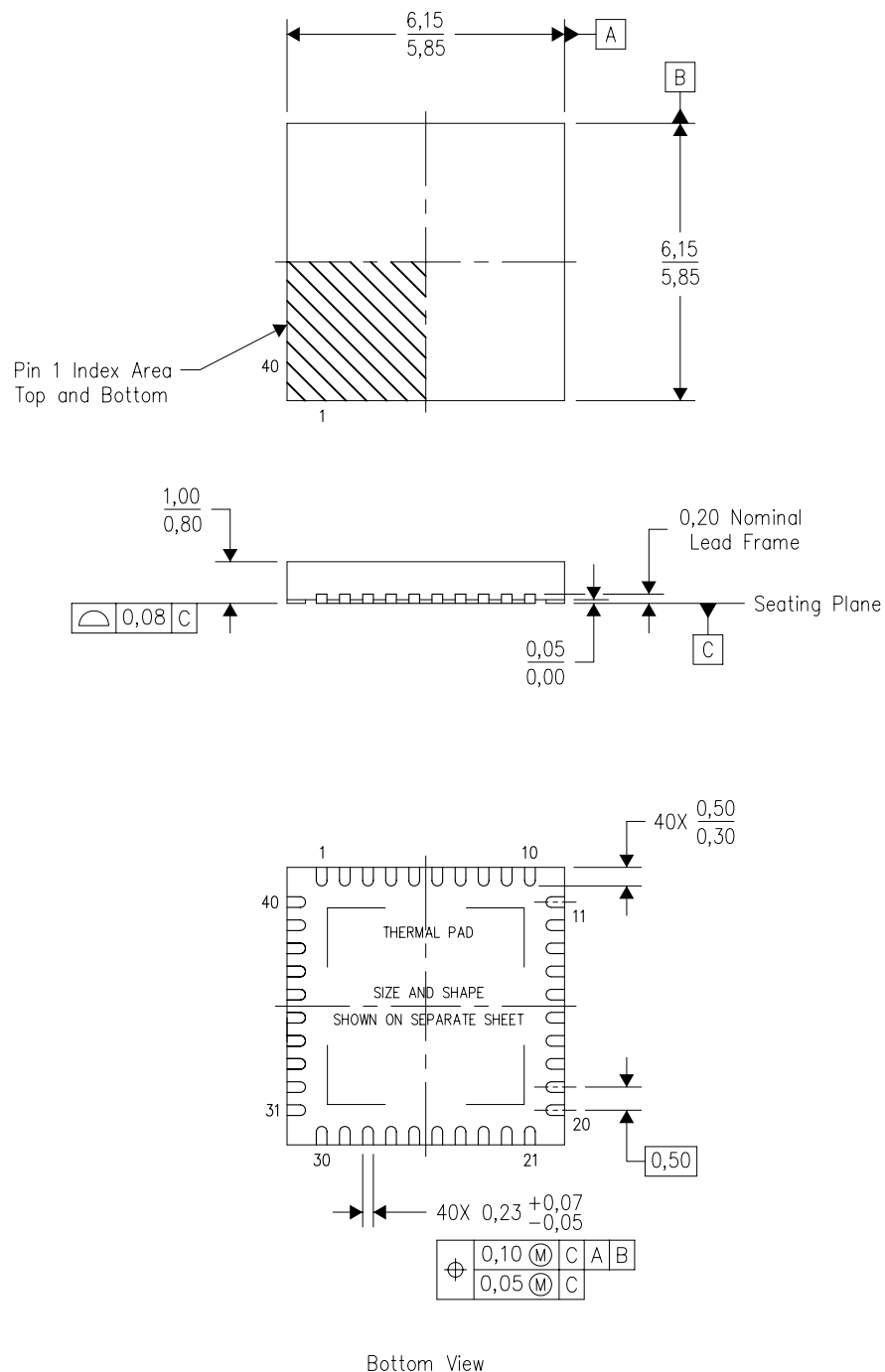
4204276/E 06/11

NOTES: A.   All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M−1994.
         B.   This drawing is subject to change without notice.
         C.   QFN (Quad Flatpack No−Lead) Package configuration.
         D.   The package thermal pad must be soldered to the board for thermal and mechanical performance.
         E.   See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
         F.   Package complies to JEDEC MO−220 variation VJJD−2.

**TEXAS
INSTRUMENTS**
www.ti.com

RHA (S−PVQFN−N40)   PLASTIC QUAD FLATPACK NO−LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No−Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

PIN 1 INDICATOR
C0,30

Exposed Thermal Pad

4,50±0,10

4,50±0,10
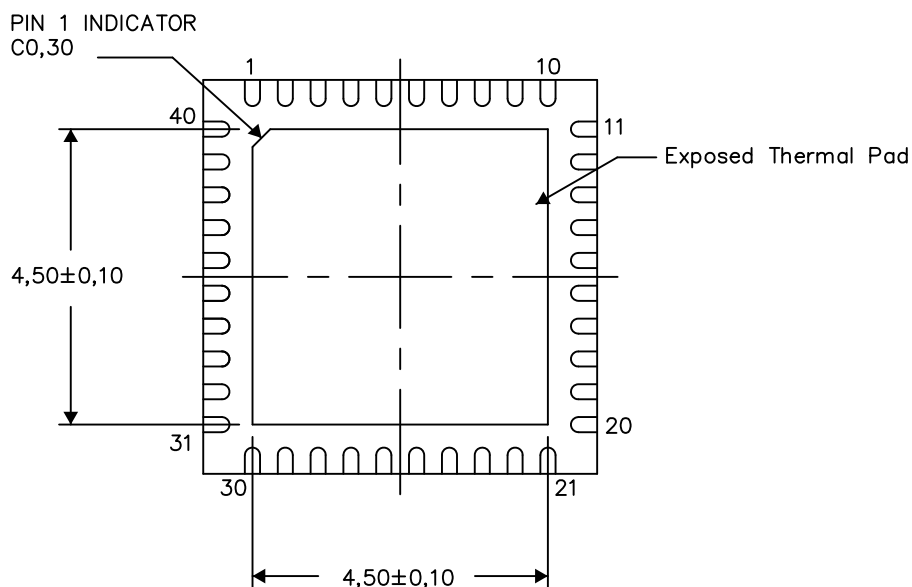
Bottom View

Exposed Thermal Pad Dimensions

4206355−4/X 08/14

NOTES: A. All linear dimensions are in millimeters

RHA (S–PVQFN–N40)   PLASTIC QUAD FLATPACK NO–LEAD

Example Board Layout

Note D

36x0,5

5,1 6,8

5,1

6,8

Non Solder Mask Defined Pad

R0,14

0,08

0,85

0,07
All Around

0,28

Example
Solder Mask Opening
(Note F)

Pad Geometry
(Note C)

Example Stencil Design
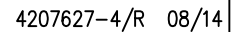0,125 Thick Stencil
(Note E)

0,8

0,230

36x0,5

0,3

0,3

5,15 6,75

9x1,2

9x1,2

5,15

6,75

(64% Printed Solder Coverage by Area)

Example Via Layout Design
Via layout may vary depending
on layout constraints
(Note D, F)

16xØ0,3

4,5

4,5

12x1,0

12x1,0

4207627–4/R   08/14

NOTES: A.   All linear dimensions are in millimeters.
         B.   This drawing is subject to change without notice.
         C.   Publication IPC–7351 is recommended for alternate designs.
         D.   This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat–Pack
              Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets
              for specific thermal information, via requirements, and recommended board layout. These documents are available at
              www.ti.com <http://www.ti.com>.
         E.   Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
              contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
         F.   Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting
              recommendations for vias placed in the thermal pad.

TEXAS
INSTRUMENTS
www.ti.com

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have *not* been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products		Applications	
Audio	www.ti.com/audio	Automotive and Transportation	www.ti.com/automotive
Amplifiers	amplifier.ti.com	Communications and Telecom	www.ti.com/communications
Data Converters	dataconverter.ti.com	Computers and Peripherals	www.ti.com/computers
DLP® Products	www.dlp.com	Consumer Electronics	www.ti.com/consumer-apps
DSP	dsp.ti.com	Energy and Lighting	www.ti.com/energy
Clocks and Timers	www.ti.com/clocks	Industrial	www.ti.com/industrial
Interface	interface.ti.com	Medical	www.ti.com/medical
Logic	logic.ti.com	Security	www.ti.com/security
Power Mgmt	power.ti.com	Space, Avionics and Defense	www.ti.com/space-avionics-defense
Microcontrollers	microcontroller.ti.com	Video and Imaging	www.ti.com/video
RFID	www.ti-rfid.com		
OMAP Applications Processors	www.ti.com/omap	**TI E2E Community**	e2e.ti.com
Wireless Connectivity	www.ti.com/wirelessconnectivity		

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265
Copyright © 2014, Texas Instruments Incorporated