

A6387

High voltage high- and low-side driver for automotive applications

Datasheet - production data



Features

- High voltage rail up to 550 V •
- dV/dt immunity ± 50 V/nsec in full temperature range
- Driver current capability
 - 400 mA source
 - 650 mA sink
- Switching times 50/30 nsec rise/fall with 1 nF load
- CMOS/TTL Schmitt-trigger inputs with hysteresis and pull down
- Internal bootstrap diode

- Outputs in phase with inputs
- Interlocking function
- AECQ100 automotive qualified

Applications

- Drive inverters for HEV and EV
- HID ballasts, power supply units
- Motion driver for home appliances, factory automation, industrial drives

Description

The A6387 is a high voltage device, manufactured with the BCD™ "offline" technology. It is a single chip half-bridge gate driver for N-channel Power MOSFETs or IGBTs. The high-side (floating) section is designed to stand a voltage rail of up to 550 V. The logic inputs are CMOS/TTL compatible for easy interfacing of the microcontroller or DSP.



Figure 1. Block diagram

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This is information on a product in full production.

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1 Electrical data

1.1 Absolute maximum ratings

Symbol	Parameter	Min.	Max.	Unit
V _{cc}	Supply voltage	- 0.3	18	V
V _{out}	Output voltage	V _{boot} - 18	V _{boot} + 0.3	V
V _{boot}	Bootstrap voltage	- 0.3	568	V
V _{hvg}	High-side gate output voltage	V _{out} - 0.3	V_{boot} + 0.3	V
V _{lvg}	Low-side gate output voltage	- 0.3	V _{cc} + 0.3	V
Vi	Logic input voltage	- 0.3	V _{cc} + 0.3	V
dV _{out} /dt	Allowed output slew rate		50	V/ns
P _{tot}	Total power dissipation ($T_A = 85 \degree$ C)		750	mW
Tj	Junction temperature		150	°C
T _{stg}	Storage temperature	-50	150	°C
ESD	Human Body Model	2		kV

Table 1. Absolute maximum ratings

1.2 Thermal data

Table 2. Thermal data

Symbol	Parameter	Value	Unit
R _{th(JA)}	Thermal resistance junction to ambient	150	°C/W

1.3 Recommended operating conditions

Table 3. Recommended operating conditions

Symbol	Pin	Parameter	Test condition	Min.	Max.	Unit
V _{cc}	3	Supply voltage		6.3	17	V
V _{BO} ⁽¹⁾	8 - 6	Floating supply voltage			17	V
V _{out}	7	Output voltage		-6 ⁽²⁾	530	V
f _{sw}		Switching frequency	HVG, LVG load C _L = 1 nF		400	kHz
Тj		Junction temperature		-40	125	°C

1. $V_{BO} = V_{boot} - V_{out}$.

2. LVG off. V_{CC} = 12 V.



2 Pin connection



Table 4. Pin descripti

No.	Pin	Туре	Function
1	LIN	I	Low-side driver logic input
2	HIN	I	High-side driver logic input
3	V _{cc}	Ρ	Low voltage power supply
4	GND	Р	Ground
5	LVG ⁽¹⁾	0	Low-side driver output
6	OUT	Р	High-side driver floating reference
7	HVG ⁽¹⁾	0	High-side driver output
8	V _{boot}	Р	Bootstrap supply voltage

 The circuit provides less than 1 V on the LVG and HVG pins (at I_{sink} = 10 mA). This allows the omitting of the "bleeder" resistor connected between the gate and the source of the external MOSFET normally used to hold the pin low.



3 Electrical characteristics

3.1 AC operation

 V_{CC} = 15 V; T_{J} = -40 °C \div 125 °C, unless otherwise specified.

Symbol	Pin	Parameter	Test condition	Min.	Тур.	Max.	Unit
t _{on}	1 vs. 5 2 vs. 7	High/low-side driver turn-on propagation delay	$V_{out} = 0 V$	40	120	240	ns
t _{off}	1 vs. 5 2 vs. 7	High/low-side driver turn-off propagation delay	V _{boot} = V _{CC} C _L = 1 nF	40	110	210	ns
t _r	5, 7	Rise time	C ₁ = 1 nF		50	100	ns
t _f	5, 7	Fall time	0L - 111		30	80	ns

 Table 5. AC operation electrical characteristics



Figure 3. Timing of input/output signals; turn-on/off propagation delays



3.2 DC operation

 V_{CC} = 15 V; T_{J} = -40 °C \div 125 °C, unless otherwise specified

Symbol	Pin	Parameter	Test condition	Min.	Тур.	Max.	Unit
Low supp	ly volt	age section					
$V_{cc_{thON}}$		V _{cc} UV turn-on threshold		5.5	6	6.3	V
V _{cc_thOFF}		V _{cc} UV turn-off threshold		5	5.5	6	V
V _{cc_hys}		V _{cc} UV hysteresis		0.3	0.5	0.7	V
I _{qccu}	3	Undervoltage quiescent supply current	$V_{cc} \le 5 V$		150	220	μA
I _{qcc}		Quiescent current			250	320	μA
R _{DSon}		Bootstrap driver on resistance ⁽¹⁾	LVG ON		125		Ω
Bootstrap	ped si	upply voltage section ⁽²⁾					
I _{QBO}		V _{BO} quiescent current	HVG ON			100	μA
I _{LK}	8	High voltage leakage current	$V_{hvg} = V_{out} = V_{boot} =$ 550 V			10	μA
High/low-	side d	river				-	
I _{so}	5 7	High/low-side source short- circuit current	V_{IN} = V_{ih} (t_p < 10 µs)	300	400		mA
I _{si}	5, 7	High/low-side sink short- circuit current	V_{IN} = V_{il} (t_p < 10 µs)	450	650		mA
Logic inpu	uts						
V _{il}		Low level logic threshold voltage				1.4	V
V _{ih}	1,2	High level logic threshold voltage		3.2			V
l _{ih}		High level logic input current	V _{IN} = 15 V	8	20	40	μA
l _{il}	1	Low level logic input current	V _{IN} = 0 V			1	μA

1. $R_{DS(on)}$ is tested in the following way:

$$\mathsf{R}_{\mathsf{DSON}} = \frac{(\mathsf{V}_{\mathsf{CC}} - \mathsf{V}_{\mathsf{BOOT}1}) - (\mathsf{V}_{\mathsf{CC}} - \mathsf{V}_{\mathsf{BOOT}2})}{\mathsf{I}_1(\mathsf{V}_{\mathsf{CC}}, \mathsf{V}_{\mathsf{BOOT}1}) - \mathsf{I}_2(\mathsf{V}_{\mathsf{CC}}, \mathsf{V}_{\mathsf{BOOT}2})}$$

where I_1 is pin 8 current when $V_{BOOT} = V_{BOOT1}$, I_2 when $V_{BOOT} = V_{BOOT2}$.

2. $V_{BO} = V_{boot} - V_{out}$.

4 Input logic

The A6387 input logic is V_{CC} (17 V) compatible. An interlocking feature is offered (see *Table 7*) to avoid undesired simultaneous turn-on of both power switches driven.

Ir	iput	Out	put
HIN	LIN	HVG	LVG
0	0	0	0
0	1	0	1
1	0	1	0
1	1	0	0

Table 7. Input logic







5 Bootstrap driver

A bootstrap circuitry is needed to supply the high voltage section. This function is normally accomplished by a high voltage fast recovery diode (*Figure 5* a). In the A6387 device a patented integrated structure replaces the external diode. It is realized by a high voltage DMOS, driven synchronously with the low-side driver (LVG), with a diode in series, as shown in *Figure 5* b. An internal charge pump (*Figure 5* b) provides the DMOS driving voltage.

C_{BOOT} selection and charging

To choose the proper C_{BOOT} value the external MOS can be seen as an equivalent capacitor. This capacitor C_{EXT} is related to the MOS total gate charge:

Equation 1

$$C_{EXT} = \frac{Q_{gate}}{V_{gate}}$$

The ratio between the capacitors C_{EXT} and C_{BOOT} is proportional to the cyclical voltage loss. It must be:

C_{BOOT}>>>C_{EXT}

For example: if Q_{gate} is 30 nC and V_{gate} is 10 V, C_{EXT} is 3 nF. With C_{BOOT} = 100 nF the drop would be 300 mV.

If HVG must be supplied for a long period, the C_{BOOT} selection must take into account also the leakage and quiescent losses.

For example: HVG steady-state consumption is lower than 100 μ A, therefore, if HVG T_{ON} is 5 ms, C_{BOOT} must supply 0.5 μ C to C_{EXT}. This charge on a 1 μ F capacitor means a voltage drop of 0.5 V.

The internal bootstrap driver offers a big advantage: the external fast recovery diode can be avoided (it usually has very high leakage current).

This structure can work only if V_{OUT} is close to GND (or lower) and, in the meantime, the LVG is on. The charging time (T_{charge}) of the C_{BOOT} is the time in which both conditions are fulfilled and it must be long enough to charge the capacitor.

The bootstrap driver introduces a voltage drop due to the DMOS R_{DSon} (typical value: 125 Ω). This drop can be neglected at low switching frequency, but it should be taken into account when operating at high switching frequency.

Equation 2 is useful to compute the drop on the bootstrap DMOS:

Equation 2

$$V_{drop} = I_{charge}R_{dson} \rightarrow V_{drop} = \frac{Q_{gate}}{T_{charge}}R_{dson}$$

where Q_{gate} is the gate charge of the external power MOS, R_{DSon} is the ON-resistance of the bootstrap DMOS, and T_{charge} is the charging time of the bootstrap capacitor.



For example: using a power MOS with a total gate charge of 30 nC, the drop on the bootstrap DMOS is about 1 V, if the T_{charge} is 5 $\mu s.$ In fact:

Equation 3

$$V_{drop} = \frac{30nC}{5\mu s} \cdot 125\Omega \sim 0.8V$$

 V_{drop} should be taken into account when the voltage drop on C_{BOOT} is calculated: if this drop is too high, or the circuit topology doesn't allow a sufficient charging time, an external diode can be used.



Figure 5. Bootstrap driver



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6 Typical characteristic









Figure 9. Turn-off time vs. temperature



Figure 10. Output source current vs. temperature

Figure 11. Output sink current vs. temperature







7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: *www.st.com*. ECOPACK is an ST trademark.







Table 8. 50-8 package mechanical data					
Symbol —		Dimensions (mm)			
Symbol	Min.	Тур.	Max.		
А			1.75		
A1	0.10		0.25		
A2	1.25				
b	0.28		0.48		
С	0.17		0.23		
D	4.80	4.90	5.00		
E	5.80	6.00	6.20		
E1	3.80	3.90	4.00		
е		1.27			
h	0.25		0.50		
L	0.40		1.27		
L1		1.04			
k	0°		8°		
ccc			0.10		

Table 8. SO-8 package mechanical data



8 Ordering information

Table	9.	Ordering	information
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Order code	Package	Packaging
A6387D	SO-8	Tube
A6387DTR	SO-8	Tape and reel

9 Revision history

Table 10. Document revision history

Date	Revision	Changes	
05-Jul-2012	1	First release	
10-Oct-2013	2	Updated: Section : Features on page 1 (added "AECQ100 compliant"). Section : Applications on page 1 added: - Drive inverters for HEV and EV, - HID ballasts, power supply units, - Motion driver for home appliances, factory automation, industrial drives. Table 1 on page 3 (removed note below Table 1). Minor corrections throughout document.	
22-Oct-2013	3	Updated <i>Section : Features on page 1</i> ("replaced AECQ100 compliant" by "AECQ100 automotive qualified").	
14-Apr-2014	4	Updated Section 3.1: AC operation on page 5 (added Figure 3). Updated Section 4: Input logic on page 7 (added Figure 4).	
04-Feb-2015	5	Updated <i>Table 1</i> (added <i>Human Body Model</i> parameter). Updated minimum supply voltage in <i>Table 3</i> and maximum V _{cc} UV turn-on threshold voltage in <i>Table 6</i> . Corrected typo in R _{DS(on)} testing equation in footnote of <i>Table 6</i> . Updated <i>Figure 5: Bootstrap driver</i> .	



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