

Key Features

- 2.5W Output at 10% THD with a 4Ω Load and 5V Power Supply
- Filterless, Low Quiescent Current and Low EMI
- High Efficiency up to 88%
- Superior Low Noise
- Short Circuit Protection
- Thermal Shutdown
- Few External Components to Save Space and Cost
- MSOP-8, SOP-8, DFN3x3 and DFN2x2 Packages Available
- Pb-Free Package

Applications

- PMP/MP4
- GPS
- Portable Speakers
- Walkie Talkie
- Handsfree phones/Speaker Phones
- Cellular Phones

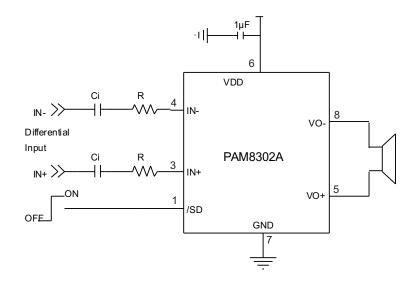
General Description

The PAM8302A is a 2.5W class-D mono audio amplifier. Its low THD+N feature offers high-quality sound reproduction. The new filterless architecture allows the device to drive speaker directly instead of using low-pass output filters, therefore save system cost and PCB area.

With the same number of external components, the efficiency of the PAM8302A is much better than that of class-AB cousins. It can optimize battery life thus is ideal for portable applications.

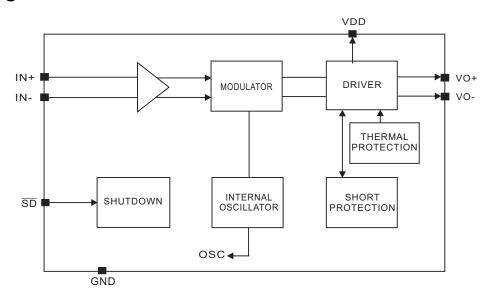
The PAM8302A is available in MSOP-8, SOP-8, DFN3x3 and DFN2x2 packages.

Typical Application

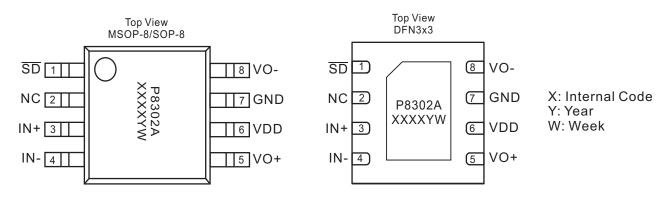


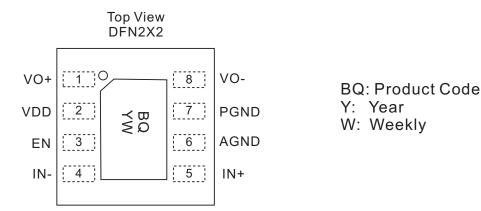


Block Diagram



Pin Configuration & Marking Information







2.5W Filterless Class-D Mono Audio Amplifier

Absolute Maximum Ratings

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Supply Voltage at no Input Signal6.0V	Maximum Junction Temperature150°C
Input Voltage0.3V to V _{DD} +0.3V	Storage Temperature65°C to 150°C
	Soldering Temperature300°C,5sec

Recommended Operating Conditions

Supply voltage Range2.5V to 5.5V	Operation Temperature Range40°Cto 85°C
	Junction Temperature Range40°C to 125°C

Thermal Information

Parameter	Symbol	Package	Maximum	Unit
Thermal Resistance (Junction to ambient)	0	SOP-8	115	°C/W
		MSOP-8	180	°C/W
	θ_{JA}	DFN 3x3-8	47.9	°C/W
		DFN 2x2-8	80	°C/W



2.5W Filterless Class-D Mono Audio Amplifier

Electrical Characteristic

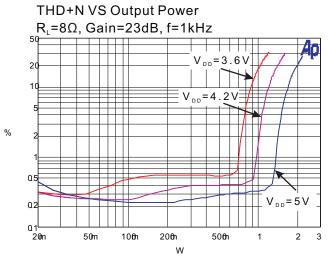
 V_{DD} =5V, Gain = 24dB, R_L =4 Ω , T_A =25°C, unless otherwise noted.

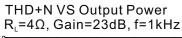
PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Supply Voltage Range	V_{DD}			2.5		5.5	V	
Quiescent Current	IQ	No Load			4	8	mA	
Shutdown Current	I _{SHDN}	V _{SHDN} =0V				1	μΑ	
		f=1kHz, R_L =4 Ω	V _{DD} =5V	2.25	2.50			
		THD+N=10%	V _{DD} =3.6V	1.10	1.25			
		f=1kHz, R _L =4Ω	V _{DD} =5V	1.80	2.00			
Output Dawar		THD+N=1%	V _{DD} =3.6V	0.86	0.95		W	
Output Power	Po	f=1kHz, R_L =8 Ω	V _{DD} =5V	1.35	1.50		VV	
		THD+N=10%	V _{DD} =3.6V	0.72	0.80			
		F=1kHz, R_L =8 Ω	V _{DD} =5V	1.15	1.30			
		THD+N=1%	V _{DD} =3.6V	0.6	0.65			
Peak Efficiency	η	f=1k	Hz		85	88	%	
		$R_L = 8\Omega$, $P_O = 0.1$ W, $f = 1$ kHz			0.30	0.35		
Total Harmonic Distortion Plus	THD+N	$R_L = 8\Omega$, $P_O = 0.5W$, $f = 1kHz$			0.45	0.50	%	
Noise	I HD+N	$R_L = 4\Omega$, $P_O = 0.1$ W, $f = 1$ kHz			0.35	0.40		
		$R_L = 4\Omega, P_O = 0.5$	W, f=1kHz		0.40	0.45		
Gain	Gv			22.5	24	25.5	dB	
Power Supply Ripple Rejection	PSRR	No input, f=1kHz,	Vpp=200mV	45	50		dB	
Dynamic Range	Dyn	f=20 to 20kHz		85	90		dB	
Signal to Noise Ratio	SNR	f =20 to 20kHz		75	80		dB	
Nicion	\ /	No A-weighting			180	300	/	
Noise	vn	A-weighting		Vn A-weighting		120	200	μV
Oscillator Frequency	f _{OSC}			200	250	300	kHz	
Drain-Source On-State		1 100 1	P MOSFET		0.45	0.50		
Resistance	R _{DS(ON)}	I _{DS} =100mA	N MOSFET		0.20	0.25	Ω	
SHDN Input High	V _{SH}			1.2			V	
SHDN Input Low	V _{SL}					0.4	V	
Over Temperature Protection	OTP	junction temperature		120	135		°C	
Over Temperature Hysteresis	OTH				30		°C	

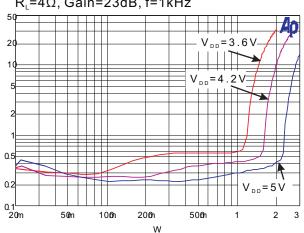


Typical Performance Characteristics

 $T_A = 25$ °C, unless otherwise noted.

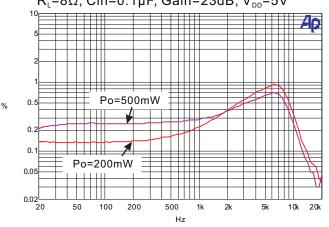






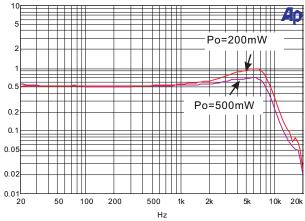
THD+N VS Frequency



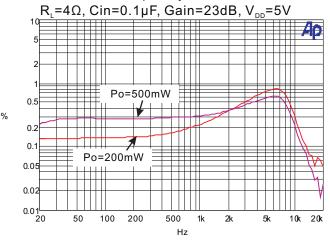


THD+N VS Frequency

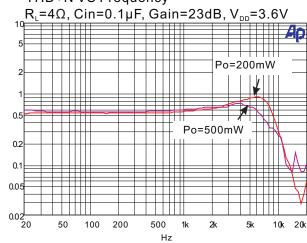




THD+N VS Frequency



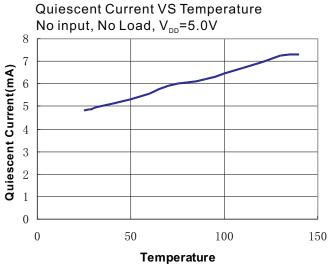
THD+N VS Frequency

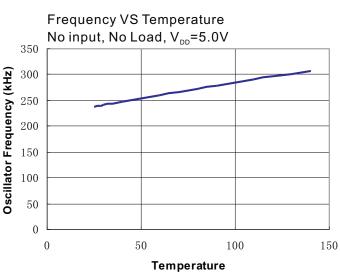


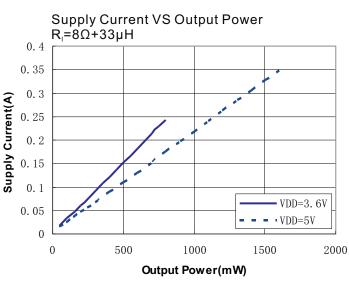


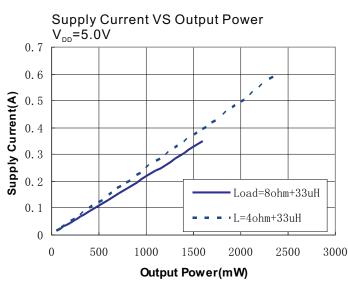
Typical Performance Characteristics

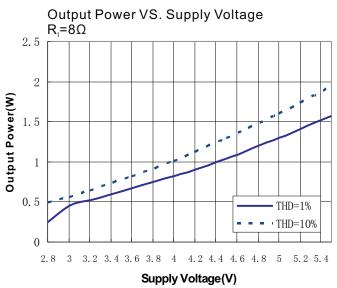
T_A=25°C, unless otherwise noted.

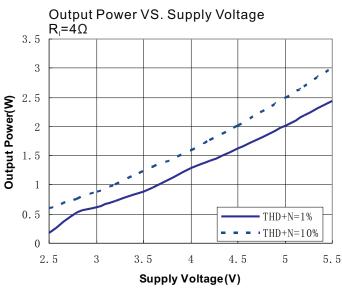








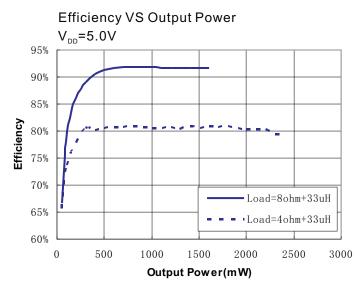


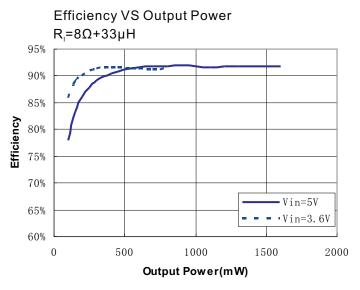


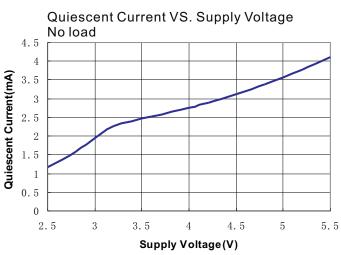


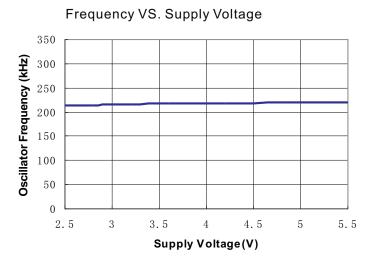
Typical Performance Characteristics

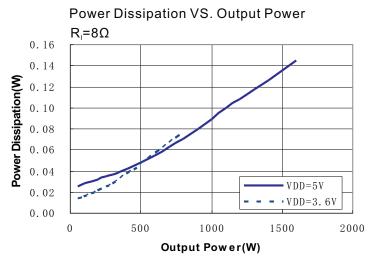
T_A=25°C, unless otherwise noted.





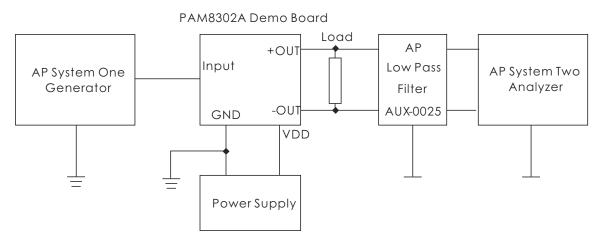








Test Setup for Performance Testing



Notes

- 1. The AP AUX-0025 low pass filter is necessary for every class-D amplifier measurement with AP analyzer.
- 2. Two $22\mu H$ inductors are used in series with load resistor to emulate the small speaker for efficiency measurement.



Application Information

Maximum Gain

As shown in block diagram (page 2), the PAM8302A has two internal amplifier stages. The first stage's gain is externally configurable, while the second stage's is internally fixed. The closed-loop gain of the first stage is set by selecting the ratio of $R_{\rm f}$ to $R_{\rm i}$ while the second stage's gain is fixed at 2x. The output of amplifier 1 serves as the input to amplifier 2, thus the two amplifiers produce signals identical in magnitude, but different in phase by 180°. Consequently, the differential gain for the IC is

$$A_{VD} = 20 \log [2 (R_f/R_i)]$$

The PAM8302A sets maximum R_i =80k Ω , minimum R_i =10k Ω , so the maximum closed-gain is 24dB.

Input Capacitors (Ci)

In typical application, an input capacitor, Ci, is required to allow the amplifier to bias input signals to a proper DC level for optimum operation. In this case, Ci and the minimum input impedance Ri (10k internal) form a high pass filter with a corner frequency determined by the following equation:

$$fc = \frac{1}{(2pRiCi)}$$

It is important to choose the value of Ci as it directly affects low frequency performance of the circuit, for example, when an application requires a flat bass response as low as 100Hz. Equation is reconfigured as follows:

$$Ci = \frac{1}{(2pR_if_c)}$$

As the input resistance is variable, for the Ci value of $0.16\mu F$, one should actually choose the Ci within the range of $0.1\mu F$ to $0.22\mu F$. A further consideration for this capacitor is the leakage path from the input source through the input network (Ri, RF,Ci) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain application. For this reason, a low leakage tantalum or ceramic

capacitor is the best choice. When a polarized capacitor is used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at VDD/2, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.

Power Supply Decoupling (Cs)

The PAM8302A is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output THD and PSRR as low as possible. Power supply decoupling affects low frequency response. Optimum decoupling is achieved by using two capacitors of different types that target different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-seriesresistance (ESR) ceramic capacitor, typically 1.0µF is good, placing it as close as possible to the device VDD terminal. For filtering lower-frequency noise signals, a capacitor of 10µF or

larger, closely located to near the audio power amplifier is recommended.

Shutdown Operation

In order to reduce shutdown power consumption, the PAM8302A contains shutdown circuitry for turn off the amplifier. This shutdown feature turns the amplifier off when a logic low is applied on the \overline{SD} pin. By switching the shutdown pin over to GND, the PAM8302A supply current draw will be minimized in idle mode.

For the best power on/off pop performance, the amplifier should be set in the shutdown mode prior to power on/off operation.

Under Voltage Lock-out (UVLO)

The PAM8302A incorporates circuitry to detect low on or off voltage. When the supply voltage drops to 2.1V or below, the PAM8302A goes into a state of shutdown, and the device comes out of its shutdown state and starts to normal operation by reset the power supply or SD pin.

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How to Reduce EMI (Electro Magnetic Interference)

A simple solution is to put an additional capacitor $1000\mu F$ at power supply terminal for power line coupling if the traces from amplifier to speakers are short (<20CM).

Most applications require a ferrite bead filter as shown at Figure 1. The ferrite filter depresses EMI of around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies and low impedance at low frequencies.

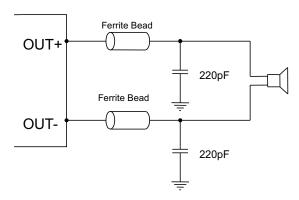
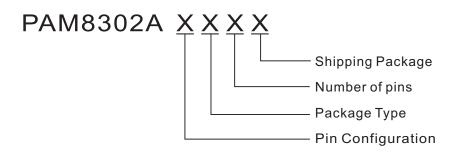


Figure 1: Ferrite Bead Filter to reduce EMI

2.5W Filterless Class-D Mono Audio Amplifier

Ordering Information



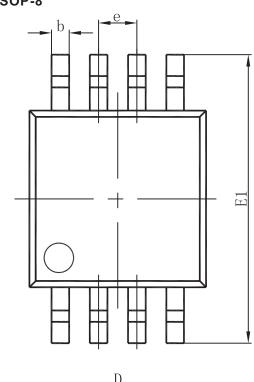
Pin C	onfiguration	Package Type	Number of pins
A:	B:	S: MSOP	C: 8
1: SD	1: VO+	D: SOP	
2: NC	2: VDD	Y: DFN3X3	
3: IN+	3: EN	G: DFN2X2	
4: IN-	4: IN-		
5: VO+	5: IN+		
6: VDD	6: AGND		
7: GND	7: PGND		
8: VO-	8: VO-		

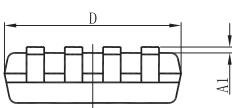
Part Number	Package Type	MOQ/Shipping Package
PAM8302AASCR	MSOP-8	2,500 Units/Tape & Reel
PAM8302AADCR	SOP-8	2,500 Units/Tape & Reel
PAM8302AAYCR	DFN3x3-8	3,000 Units/Tape & Reel
PAM8302ABGCR	DFN2x2-8	3,000 Units/Tape & Reel

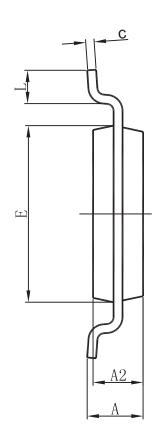


Outline Dimensions







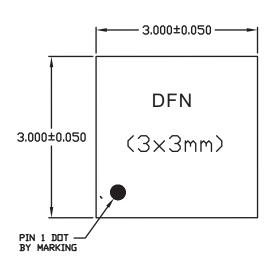


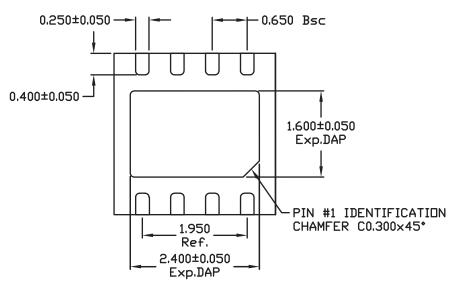
REF	Millimeter		
	Min	Max	
Α		1.10	
A1	0.05	0.15	
A2	0.78	0.94	
b	0.22	0.38	
С	0.08	0.23	
D	2.90	3.10	
E	2.90	3.10	
E1	4.75	5.05	
е	0.65BSC		
L	0.40 0.70		

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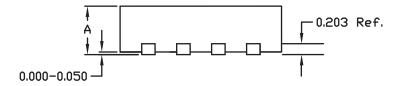
Outline Dimensions

DFN 3x3





Α	MAX.	0.800
	NOM.	0.750
	MIN.	0.700

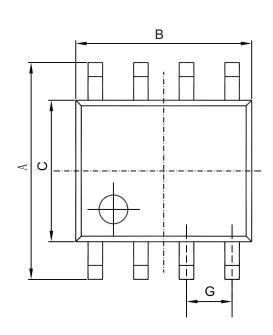


Note: All dimensions are in Millimeters.

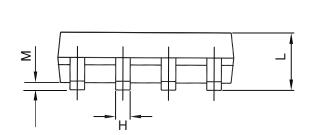


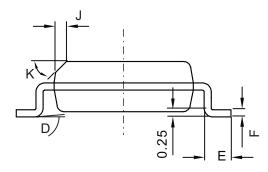
Outline Dimensions

SOP-8



REF	DIMENSIONS		
	Millimeters		
	Min	Max	
A	5.80	6.20	
В	4.80	5.00	
С	3.80	4.00	
D	0°	8°	
Е	0.40 0.90		
F	0.19 0.25		
M	0.10 0.25		
Н	0.35	0.49	
L	1.35	1.75	
J	0.375 REF		
K	45°		
G	1.27 TYP		







Outline Dimensions

DFN 2x2

